## INTEGRATED CIRCUITS

# DATA SHEET



P83C660X2, P87C660X2 P83C661X2, P87C661X2

80C51 8-bit microcontroller family 16KB OTP/ROM, 512B RAM low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

Product data
Supersedes data of 2003 Jun 19

2003 Oct 02





80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

## P8xC660X2/661X2



#### **DESCRIPTION**

The devices are Single-Chip 8-Bit Microcontrollers manufactured in an advanced CMOS process and are derivatives of the 80C51 microcontroller family. The instruction set is 100% compatible with the 80C51 instruction set.

The devices support 6-clock/12-clock mode selection by programming an OTP bit (OX2) using parallel programming. In addition, an SFR bit (X2) in the clock control register (CKCON) also selects between 6-clock/12-clock mode.

These devices have either one or two I<sup>2</sup>C interfaces, capable of handling speeds up to 400 kbits/s (Fast I<sup>2</sup>C). They also have four 8-bit I/O ports, three 16-bit timer/event counters, a multi-source, four-priority-level, nested interrupt structure, an enhanced UART and on-chip oscillator and timing circuits.

The added features of the P8xC66xX2 make it a powerful microcontroller for applications that require pulse width modulation, high-speed I/O, I<sup>2</sup>C communication, and up/down counting capabilities such as motor control.

#### **FEATURES**

- 80C51 Central Processing Unit
  - 16 kbytes OTP (87C660X2, 87C661X2)
  - 16 kbytes ROM (83C660X2, 83C661X2)
  - 512 byte RAM
  - Boolean processor
  - Fully static operation
  - Low voltage (2.7 V to 5.5 V at 16 MHz) operation
- 12-clock operation with selectable 6-clock operation (via software or via parallel programmer)
- Memory addressing capability
  - Up to 64 kbytes ROM and 64 kbytes RAM
- Power control modes:
  - Clock can be stopped and resumed
  - Idle mode
  - Power-down mode

- CMOS and TTL compatible
- Two speed ranges at V<sub>CC</sub> = 5 V
  - 0 to 30 MHz with 6-clock operation
  - 0 to 33 MHz with 12-clock operation
- Parallel programming with 87C51 compatible hardware interface to programmer
- RAM expandable externally to 64 kbytes
- Programmable Counter Array (PCA)
  - PWM
  - Capture/compare
- PLCC and LQFP packages
- Extended temperature ranges
- Dual Data Pointers
- Security bits (3 bits)
- Encryption array 64 bytes
- 8/9 interrupt sources
- Four interrupt priority levels
- Four 8-bit I/O ports
- One I<sup>2</sup>C serial port interface has a selectable data transfer mode, either 400 kB/sec Fast-mode or 100 kB/sec Standard-mode (8xC660X2 and 8xC661X2)
- A second I<sup>2</sup>C serial port interface has the 400 kB/sec Fast data-transfer mode only and selectable slew rate control of the output pins (8xC661X2)
- Full-duplex enhanced UART
  - Framing error detection
  - Automatic address recognition
- Three 16-bit timers/counters T0, T1 (standard 80C51) and additional T2 (capture and compare)
- Programmable clock-out pin
- Asynchronous port reset
- Low EMI (inhibit ALE, slew rate controlled outputs, and 6-clock mode)
- Wake-up from Power Down by an external interrupt

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB  $\rm I^2C$  interfaces

## P8xC660X2/661X2

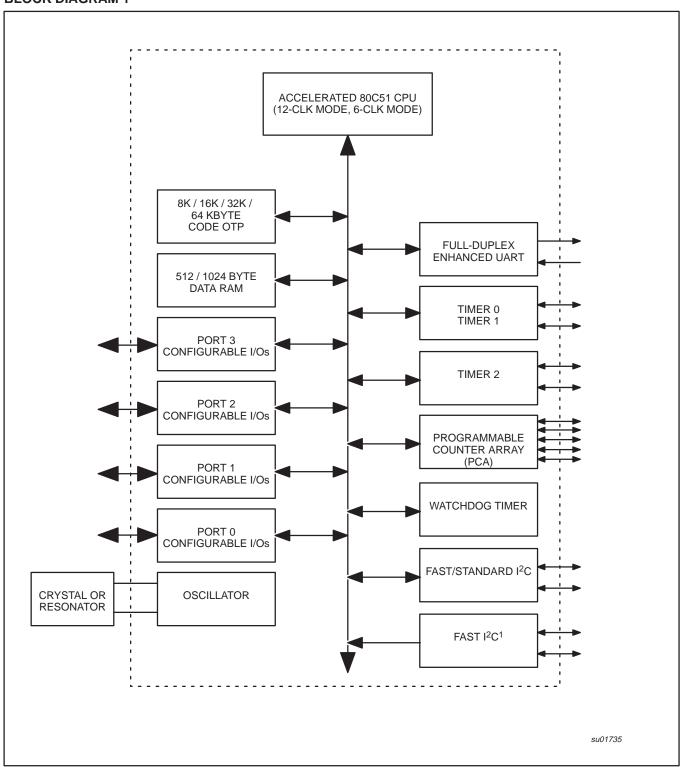
#### **SELECTION TABLE**

Туре		Memo	ry			Tim	ers		lı	Ser nterf	ial aces	5										
	RAM	ROM	ОТР	Flash	# of Timers	PWM	PCA	WD	UART	12C	CAN	SPI	ADC bits/ch.	I/O Pins	Interrupts (Ext.)/Levels	Program Security	Default Clock Rate	Optional Clock Rate	Reset active low/high?	Max. Freq. at 6-clk / 12-clk (MHz)	Freq. Range at 3V (MHz) (6-clk / 12-clk)	Freq. Range at 5V (MHz) (6-clk / 12-clk)
P87C660X2	512B	-	16K	-	4	√	√	√	√	1	-	-	-	32	7(2)/4	√	12-clk	6-clk	Н	30/33	0-16	0-30/33
P83C660X2	512B	16K	-	-	4	√	√	√	√	1	-	-	-	32	7(2)/4	√	12-clk	6-clk	Н	30/33	0-16	0-30/33
P87C661X2	512B	-	16K	-	4	√	√	√	1	2	-	-	-	32	7(2)/4	1	12-clk	6-clk	Н	30/33	0-16	0-30/33
P83C661X2	512B	16K	-	-	4	√	√	√	1	2	_	-	_	32	7(2)/4	V	12-clk	6-clk	Н	30/33	0-16	0-30/33

## **ORDERING INFORMATION**

Type number	Type number						Temp Range	
	ОТР	ROM	RAM	Name	Description		(°C)	
P83C660X2FA	-	16 KB	512B	PLCC44	plastic leaded chip carrier; 44 leads	SOT187-2	-40 to +85	
P83C660X2BBD	_	16 KB	512B	LQFP44	plastic low profile quad flat package; 44 leads; body 10 $\times$ 10 $\times$ 1.4 mm	SOT389-1	0 to +70	
P87C660X2FA	16 KB	_	512B	PLCC44	plastic leaded chip carrier; 44 leads	SOT187-2	-40 to +85	
P87C660X2BBD	16 KB	-	512B	LQFP44	plastic low profile quad flat package; 44 leads; body 10 $\times$ 10 $\times$ 1.4 mm	SOT389-1	0 to +70	
P83C661X2FA	_	16 KB	512B	PLCC44	plastic leaded chip carrier; 44 leads	SOT187-2	-40 to +85	
P83C661X2BBD	-	16 KB	512B	LQFP44	plastic low profile quad flat package; 44 leads; body 10 × 10 × 1.4 mm	SOT389-1	0 to +70	
P87C661X2FA	16 KB	-	512B	PLCC44	plastic leaded chip carrier; 44 leads	SOT187-2	-40 to +85	
P87C661X2BBD	16 KB	_	512B	LQFP44	plastic low profile quad flat package; 44 leads; body 10 × 10 × 1.4 mm	SOT389-1	0 to +70	

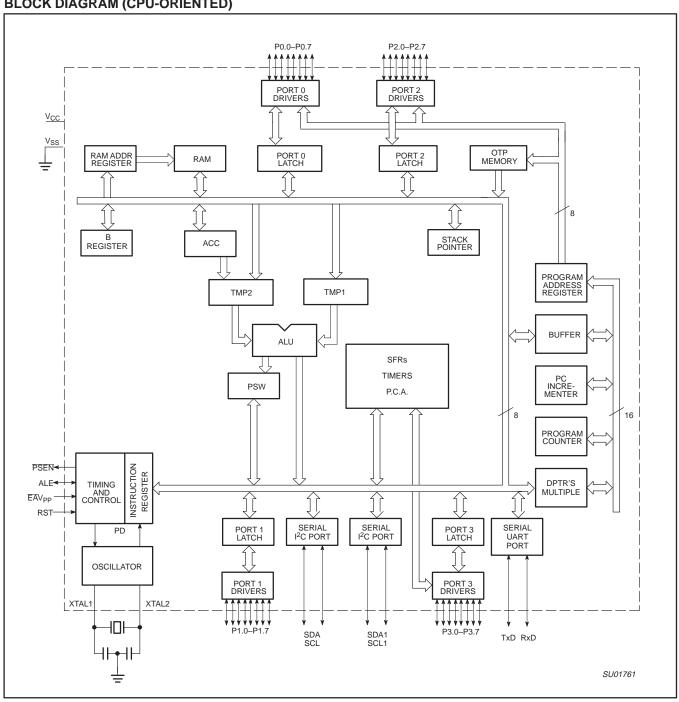
#### **BLOCK DIAGRAM 1**



1. 2nd I<sup>2</sup>C on P8xC661X2 only.

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## **BLOCK DIAGRAM (CPU-ORIENTED)**

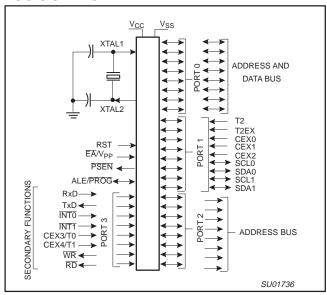


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80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

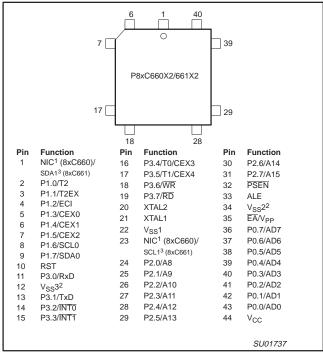
## P8xC660X2/661X2

#### LOGIC SYMBOL



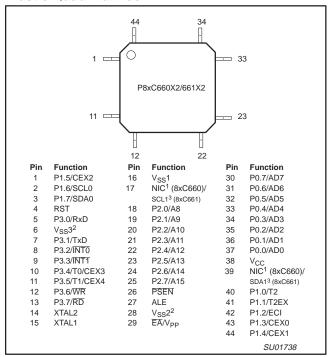
#### **PINNING**

#### **Plastic Leaded Chip Carrier**



- 1. No internal connection
- 2. May be left open, but it is recommended to connect  $V_{SS}2$  and  $V_{SS}3$  to GND to improve EMC performance
- 3. P8xC661X2 devices only, these pins are open-drain and have the same electrical characteristics as P1.6 and P1.7

#### **Plastic Quad Flat Pack**



- 1. No internal connection
- 2. May be left open, but it is recommended to connect  $\rm V_{SS}2$  and  $\rm V_{SS}3$  to GND to improve EMC performance
- 3. P8xC661X2 devices only

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## **PIN DESCRIPTIONS**

	PIN NU	IMBER			
MNEMONIC	PLCC	LQFP	TYPE	NAME AND FUNCTION	
V <sub>SS1</sub>	22	16	ı	Ground: 0 V reference.	
$V_{SS2}$	34	28	- 1	Ground: Additional ground pin (may be left open).	
$V_{SS3}$	12	6	ı	Ground: Additional ground pin (may be left open).	
V <sub>CC</sub>	44	38	I	<b>Power Supply:</b> This is the power supply voltage for normal, idle, and power-down operation.	
P0.0-0.7 <sup>2</sup>	43–36	37–30	I/O	<b>Port 0:</b> Port 0 is an open-drain, bidirectional I/O port. Port 0 pins that have 1s written to them float and can be used as high-impedance inputs. Port 0 is also the multiplexed low-order address and data bus during accesses to external program and data memory. In this application, it uses strong internal pull-ups when emitting 1s.	
P1.0–P1.7 <sup>2</sup>	2–9	40–44, 1–3	I/O	Port 1: Port 1 is an 8-bit bidirectional I/O port with internal pull-ups on all pins. Port 1 pins that have 1s written to them are pulled HIGH by the internal pull-ups and can be used as inputs. As inputs, port 1 pins that are externally pulled LOW will source current because of the internal pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ).	
				Alternate functions for P8xC660X2/661X2 Port 1 include:	
	2	40	I/O	T2 (P1.0): Timer/Counter 2 external count input/Clockout (see Programmable Clock-Out)	
	3	41	1	T2EX (P1.1): Timer/Counter 2 Reload/Capture/Direction Control	
	4	42	ı	ECI (P1.2): External Clock Input to the PCA	
	5	43	I/O	CEX0 (P1.3): Capture/Compare External I/O for PCA module 0	
	6	44	I/O	CEX1 (P1.4): Capture/Compare External I/O for PCA module 1	
	7	1	I/O	CEX2 (P1.5): Capture/Compare External I/O for PCA module 2	
	8	2	I/O	SCL (P1.6): I <sup>2</sup> C bus clock line (open drain)	
	9	3	I/O	SCL (P1.7): I <sup>2</sup> C bus data line (open drain)	
P2.0-P2.7 <sup>2</sup>	24–31	18–25	I/O	Port 2: Port 2 is an 8-bit bidirectional I/O port with internal pull-ups. Port 2 pins that have 1s written to them are pulled HIGH by the internal pull-ups and can be used as inputs. As inputs, port 2 pins that are externally being pulled LOW will source current because of the internal pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ). Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @DPTR). In this application, it uses strong internal pull-ups when emitting 1s. During accesses to external data memory that use 8-bit addresses (MOV @Ri), port 2 emits the contents of the P2 special function register.	
P3.0-P3.7 <sup>2</sup>	11, 13–19	5, 7–13	I/O	<b>Port 3:</b> Port 3 is an 8-bit bidirectional I/O port with internal pull-ups. Port 3 pins that have 1s written to them are pulled HIGH by the internal pull-ups and can be used as inputs. As inputs, port 3 pins that are externally being pulled LOW will source current because of the pull-ups. (See DC Electrical Characteristics: I <sub>IL</sub> ). Port 3 also serves the special features of the P8xC660X2/661X2, as listed below:	
	11	5	I	RxD (P3.0): Serial input port	
	13	7	0	TxD (P3.1): Serial output port	
	14	8	- 1	INTO (P3.2): External interrupt 0	
	15	9	I	INT1 (P3.3): External interrupt 1	
	16	10	I	CEX3/T0 (P3.4): Timer 0 external input; capture/compare external I/O for PCA module 3	
	17	11	I	CEX4/T1 (P3.5): Timer 1 external input; capture/compare external I/O for PC/module 4	
	18	12	0	WR (P3.6): External data memory write strobe	
	19	13	0	RD (P3.7): External data memory read strobe	
RST <sup>2</sup>	10	4	I	<b>Reset:</b> A HIGH on this pin for two machine cycles while the oscillator is running, resets the device. An internal resistor to V <sub>SS</sub> permits a power-on reset using only an external capacitor to V <sub>CC</sub> .	
SCL1	23	17	I/O	Second I <sup>2</sup> C bus clock line (open drain) (P8xC661X2)	
SDA1	1	39	I/O	Second I <sup>2</sup> C bus data line (open drain) (P8xC661X2)	

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

MNEMONIC	PIN NU	IMBER	TYPE	NAME AND FUNCTION	
MINEMONIC	PLCC	LQFP	IIFE	NAME AND FUNCTION	
ALE <sup>2</sup>	33	27	0	Address Latch Enable: Output pulse for latching the LOW byte of the address during an access to external memory. In normal operation, ALE is emitted twice every machine cycle, and can be used for external timing or clocking. Note that one ALE pulse is skipped during each access to external data memory. ALE can be disabled by setting SFR auxiliary.0. With this bit set, ALE will be active only during a MOVX instruction.	
PSEN <sup>2</sup>	32	26	0	Program Store Enable: The read strobe to external program memory. When executing code from the external program memory, PSEN is activated twice each machine cycle, except that two PSEN activations are skipped during each access to external data memory. PSEN is not activated during fetches from internal program memory.	
EA <sup>2</sup>	35	29	I	External Access Enable/Programming Supply Voltage: $\overline{EA}$ must be externally held LOW to enable the device to fetch code from external program memory locations. If $\overline{EA}$ is held HIGH, the device executes from internal program memory. The value on the $\overline{EA}$ pin is latched when RST is released and any subsequent changes have no effect. This pin also receives the programming supply voltage (V <sub>PP</sub> ) during programming.	
XTAL1	21	15	I	Crystal 1: Input to the inverting oscillator amplifier and input to the internal clock generator circuits.	
XTAL2	20	14	0	Crystal 2: Output from the inverting oscillator amplifier.	

#### NOTE:

To avoid "latch-up" effect at power-on, the voltage on any pin (other than EA) must not be higher than V<sub>CC</sub> + 0.5 V or less than V<sub>SS</sub> – 0.5 V.
 The pins are designed for test mode also.

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80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB  $\rm l^2C$  interfaces

#### **SPECIAL FUNCTION REGISTERS**

SYMBOL	DESCRIPTION	DIRECT	ВІТ	ADDRES	S, SYMB	OL, OR A	LTERNAT	TIVE POR	T FUNCT	ION	RESET
STWIBOL	DESCRIPTION	ADDRESS	MSB	,						LSB	VALUE
ACC*	Accumulator	E0H	E7	E6	E5	E4	E3	E2	E1	E0	00H
AUXR#	Auxiliary	8EH	_	_	SRD	-	FME	_	EXTRAM	AO	xxxx0x10E
AUXR1#	Auxiliary 1	A2H	-	-	-	LPEP	GPS	0	-	DPS	xxxx00x0E
B*	B register	F0H	F7	F6	F5	F4	F3	F2	F1	F0	00H
CCAP0H#	Module 0 Capture High	FAH									xxxxxxxxB
CCAP1H#	Module 1 Capture High	FBH									xxxxxxxxB
CCAP2H#	Module 2 Capture High	FCH									xxxxxxxxE
CCAP3H#	Module 3 Capture High	FDH									xxxxxxxx
CCAP4H#	Module 4 Capture High	FEH									xxxxxxxxE
CCAP0L#	Module 0 Capture Low	EAH									xxxxxxxxE
CCAP1L#	Module 1 Capture Low	EBH									xxxxxxxxE
CCAP2L#	Module 2 Capture Low	ECH									xxxxxxxxB
CCAP3L#	Module 3 Capture Low	EDH									xxxxxxxxB
CCAP4L#	Module 4 Capture Low	EEH									xxxxxxxxB
CCAPM0#	Module 0 Mode	C2H	_	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	x00000000
CCAPM1#	Module 1 Mode	СЗН	_	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	x00000000
CCAPM2#	Module 2 Mode	C4H	-	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	x00000000E
CCAPM3#	Module 3 Mode	C5H	_	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	x0000000E
CCAPM4#	Module 4 Mode	C6H	_	ECOM	CAPP	CAPN	MAT	TOG	PWM	ECCF	x0000000E
			C7	C6	C5	C4	C3	C2	C1	C0	]
CCON*#	PCA Counter Control	C0H	CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0	00x00000l
CH#	PCA Counter High	F9H									00H
CL#	PCA Counter Low	E9H									00H
CMOD#	PCA Counter Mode	C1H	CIDL	WDTE	_	_	_	CPS1	CPS0	ECF	00xxx000E
CKCON	Clock control	8FH	_	_	-	_	_	_	_	X2	хххххххх1В
DPTR:	Data Pointer (2 bytes)										
DPH	Data Pointer High	83H									00H
DPL	Data Pointer Low	82H									00H
			AF	AE	AD	AC	AB	AA	A9	A8	
IEN0*	Interrupt Enable 0	A8H	EA	EC	ES1	ES0	ET1	EX1	ET0	EX0	0000000E
IEN1*	Interrupt Enable 1	E8H	_	_	_	_	_	_	ES2	ET2	xxxxxx00E
			BF	BE	BD	BC	BB	BA	B9	B8	
IP*#	Interrupt Priority	B8H	PT2	PPC	PS1	PS0	PT1	PX1	PT0	PX0	0000000E
			B7	В6	B5	B4	В3	B2	B1	В0	
IPH#	Interrupt Priority High	В7Н	PT2H	PPCH	PS1H	PS0H	PT1H	PX1H	PT0H	PX0H	0000000E
			87	86	85	84	83	82	81	80	
P0*	Port 0	80H	AD7	AD6	AD5	AD4	AD3	AD2	AD1	AD0	FFH
			97	96	95	94	93	92	91	90	1
P1*#	Port 1	90H	SDA	SCL	CEX2	CEX1	CEX0	ECI	T2EX	T2	FFH
			A7	A6	A5	A4	А3	A2	A1	A0	]
P2*	Port 2	A0H	AD15	AD14	AD13	AD12	AD11	AD10	AD9	AD8	FFH
			B7	B6	B5	B4	B3	B2	B1	B0	1

<sup>\*</sup> SFRs are bit addressable.

<sup>#</sup> SFRs are modified from or added to the 80C51 SFRs.

Reserved bits.

<sup>1.</sup> Reset value depends on reset source.

<sup>2. 8</sup>xC661X2 only.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

P8xC660X2/661X2

## SPECIAL FUNCTION REGISTERS (Continued)

SYMBOL	DESCRIPTION	DIRECT ADDRESS	BIT MSB	ADDRES	S, SYMB	OL, OR A	LTERNATI	VE POR	T FUNCT	ION LSB	RESET VALUE
P3*	Port 3	вон	RD	WR	T1/ CEX4	T0/ CEX3	ĪNT1	ĪNT0	TxD	RxD	FFH
PCON#1	Power Control	87H	SMOD1	SMOD0	_	POF	GF1	GF0	PD	IDL	00xx0000B
			D7	D6	D5	D4	D3	D2	D1	D0	
PSW	Program Status Word	D0H	CY	AC	F0	RS1	RS0	OV	F1	Р	000000x0B
RCAP2H# RCAP2L# SADDR# SADEN#	Timer 2 Capture High Timer 2 Capture Low Slave Address Slave Address Mask	CBH CAH A9H B9H									00H 00H 00H 00H
SBUF	Serial Data Buffer	99H									xxxxxxxxB
			9F	9E	9D	9C	9B	9A	99	98	
SCON* SP	Serial Control Stack Pointer	98H 81H	SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI	00H 07H
SF	Stack Follitel	ОІП	8F	8E	8D	8C	8B	8A	89	88	0/11
TCON*	Timer Control	88H	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00H
			CF	CE	CD	СС	СВ	CA	C9	C8	
T2CON*	Timer 2 Control	C8H	TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	00H
T2MOD#	Timer 2 Mode Control	C9H	-	-	-	-	-	_	T2OE	DCEN	xxxxxx00B
TH0 TH1 TH2# TL0 TL1 TL2#	Timer High 0 Timer High 1 Timer High 2 Timer Low 0 Timer Low 1 Timer Low 2	8CH 8DH CDH 8AH 8BH CCH									00H 00H 00H 00H 00H 00H
TMOD	Timer Mode	89H	GATE	C/T	M1	M0	GATE	C/T	M1	M0	00H
S1CON	I2C Control	D8H	CR2	ENA1	STA	STO	SI	AA	CR1	CR0	00H
S1STA	I2C STATUS	D9H	SC4	SC3	SC2	SC1	SC0	0	0	0	F8H
S1DAT	I2C DATA	DAH									00H
S1ADR	I2C ADDRESS	DBH								GC	00H
S2CON <sup>2</sup>	Second I <sup>2</sup> C control	F8H	CR2	ENA1	STA	STO	SI	AA	CR1	CR0	00H
S2STA <sup>2</sup>	Second I <sup>2</sup> C	F1H	SC4	SC3	SC2	SC1	SC0	0	0	0	F8H
S2DAT <sup>2</sup>	Second I <sup>2</sup> C	F2H									00H
S2ADR <sup>2</sup>	Second I <sup>2</sup> C	F3H								GC	00H
S2IST <sup>2</sup>	Second I <sup>2</sup> C	F4H									
IP1 <sup>2</sup>	Interrupt priority 1	E7H								PS2	00H
IP1H <sup>2</sup>		F7H								PS2H	00H
WDTRST	Watchdog Timer Reset	A6H									

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<sup>\*</sup> SFRs are bit addressable.
# SFRs are modified from or added to the 80C51 SFRs.

<sup>-</sup> Reserved bits.

Reset value depends on reset source.

<sup>2. 8</sup>xC661X2 only.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

## P8xC660X2/661X2

#### **CLOCK CONTROL REGISTER (CKCON)**

This device allows control of the 6-clock/12-clock mode by means of both an SFR bit (X2) and an OTP bit. The OTP clock control bit

OX2, when programmed by a parallel programmer (6-clock mode), supersedes the X2 bit (CKCON.0). The CKCON register is shown below in Figure 1.

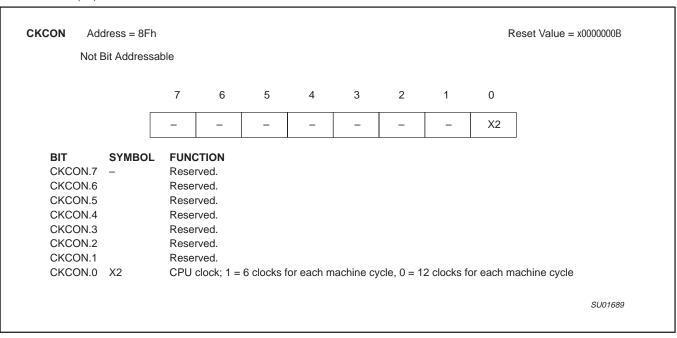


Figure 1. Clock control (CKCON) register

Also please note that the clock divider applies to the serial port for modes 0 & 2 (fixed baud rate modes). This is because modes 1 & 3 (variable baud rate modes) use either Timer 1 or Timer 2.

Below is the truth table for the CPU clock mode.

Table 1.

OX2 clock mode bit (can only be set by parallel programmer)	X2 bit (CKCON.0)	CPU clock mode
erased	0	12-clock mode (default)
erased	1	6-clock mode
programmed	Х	6-clock mode

#### **RESET**

A reset is accomplished by holding the RST pin HIGH for at least two machine cycles (12 oscillator periods in 6-clock mode, or 24 oscillator periods in 12-clock mode), while the oscillator is running. To ensure a good power-on reset, the RST pin must be HIGH long enough to allow the oscillator time to start up (normally a few milliseconds) plus two machine cycles. At power-on, the voltage on  $V_{CC}$  and RST must come up at the same time for a proper start-up. Ports 1, 2, and 3 will asynchronously be driven to their reset condition when a voltage above  $V_{IH1}$  (min.) is applied to RST.

The value on the  $\overline{\text{EA}}$  pin is latched when RST is deasserted and has no further effect.

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## P8xC660X2/661X2

## LOW POWER MODES Stop Clock Mode

The static design enables the clock speed to be reduced down to 0 MHz (stopped). When the oscillator is stopped, the RAM and Special Function Registers retain their values. This mode allows step-by-step utilization and permits reduced system power consumption by lowering the clock frequency down to any value. For lowest power consumption the Power Down mode is suggested.

#### **Idle Mode**

In the idle mode (see Table 2), the CPU puts itself to sleep while all of the on-chip peripherals stay active. The instruction to invoke the idle mode is the last instruction executed in the normal operating mode before the idle mode is activated. The CPU contents, the on-chip RAM, and all of the special function registers remain intact during this mode. The idle mode can be terminated either by any enabled interrupt (at which time the process is picked up at the interrupt service routine and continued), or by a hardware reset which starts the processor in the same manner as a power-on reset.

#### **Power-Down Mode**

To save even more power, a Power Down mode (see Table 2) can be invoked by software. In this mode, the oscillator is stopped and the instruction that invoked Power Down is the last instruction executed. The on-chip RAM and Special Function Registers retain their values down to 2 V and care must be taken to return  $V_{CC}$  to the minimum specified operating voltages before the Power Down Mode is terminated.

Either a hardware reset or external interrupt can be used to exit from Power Down. Reset redefines all the SFRs but does not change the on-chip RAM. An external interrupt allows both the SFRs and the on-chip RAM to retain their values.

To properly terminate Power Down, the reset or external interrupt should not be executed before  $V_{CC}$  is restored to its normal operating level and must be held active long enough for the oscillator to restart and stabilize (normally less than 10 ms).

With an external interrupt, INT0 and INT1 must be enabled and configured as level-sensitive. Holding the pin LOW restarts the oscillator but bringing the pin back HIGH completes the exit. Once the interrupt is serviced, the next instruction to be executed after RETI will be the one following the instruction that put the device into Power Down.

#### **LPEP**

The EPROM array contains some analog circuits that are not required when  $V_{CC}$  is less than 3.6 V but are required for a  $V_{CC}$  greater than 3.6 V. The LPEP bit (AUXR.4), when set, will powerdown these analog circuits resulting in a reduced supply current. This bit should be set ONLY for applications that operate at a  $V_{CC}$  less than 4 V.

#### **POWER-ON FLAG**

The Power-On Flag (POF) is set by on-chip circuitry when the  $\rm V_{CC}$  level on the P8xC66xX2 rises from 0 to 5 V. The POF bit can be set or cleared by software allowing a user to determine if the reset is the result of a power-on or a warm start after powerdown. The  $\rm V_{CC}$  level must remain above 3 V for the POF to remain unaffected by the  $\rm V_{CC}$  level.

#### **Design Consideration**

When the idle mode is terminated by a hardware reset, the device normally resumes program execution, from where it left off, up to two machine cycles before the internal reset algorithm takes control. On-chip hardware inhibits access to internal RAM in this event, but access to the port pins is not inhibited. To eliminate the possibility of an unexpected write when Idle is terminated by reset, the instruction following the one that invokes Idle should not be one that writes to a port pin or to external memory.

#### ONCE™ Mode

The ONCE ("On-Circuit Emulation") Mode facilitates testing and debugging of systems without the device having to be removed from the circuit. The ONCE Mode is invoked by:

- 1. Pull ALE LOW while the device is in reset and PSEN is HIGH;
- 2. Hold ALE LOW as RST is deactivated.

While the device is in ONCE Mode, the Port 0 pins go into a float state, and the other port pins and ALE and PSEN are weakly pulled HIGH. The oscillator circuit remains active. While the device is in this mode, an emulator or test CPU can be used to drive the circuit. Normal operation is restored when a normal reset is applied.

#### **Programmable Clock-Out**

A 50% duty cycle clock can be programmed to come out on P1.0. This pin, besides being a regular I/O pin, has two alternate functions. It can be programmed:

- 1. to input the external clock for Timer/Counter 2, or
- to output a 50% duty cycle clock ranging from 61 Hz to 4 MHz at a 16 MHz operating frequency in 12-clock mode (122 Hz to 8 MHz in 6-clock mode).

To configure the Timer/Counter 2 as a clock generator, bit C/T2 (in T2CON) must be cleared and bit T20E in T2MOD must be set. Bit TR2 (T2CON.2) also must be set to start the timer.

The Clock-Out frequency depends on the oscillator frequency and the reload value of Timer 2 capture registers (RCAP2H, RCAP2L) as shown in this equation:

 $\frac{\text{Oscillator Frequency}}{\text{n} \times (65536 - \text{RCAP2H}, \text{RCAP2L})}$ 

n = 2 in 6-clock mode 4 in 12-clock mode

Where (RCAP2H,RCAP2L) = the content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

In the Clock-Out mode Timer 2 roll-overs will not generate an interrupt. This is similar to when it is used as a baud-rate generator. It is possible to use Timer 2 as a baud-rate generator and a clock generator simultaneously. Note, however, that the baud-rate and the Clock-Out frequency will be the same.

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Table 2. External Pin Status During Idle and Power-Down Mode

MODE	PROGRAM MEMORY	ALE	PSEN	PORT 0	PORT 1	PORT 2	PORT 3
Idle	Internal	1	1	Data	Data	Data	Data
Idle	External	1	1	Float	Data	Address	Data
Power-down	Internal	0	0	Data	Data	Data	Data
Power-down	External	0	0	Float	Data	Data	Data

#### TIMER 0 AND TIMER 1 OPERATION

#### Timer 0 and Timer 1

The "Timer" or "Counter" function is selected by control bits C/T in the Special Function Register TMOD. These two Timer/Counters have four operating modes, which are selected by bit-pairs (M1, M0) in TMOD. Modes 0, 1, and 2 are the same for both Timers/Counters. Mode 3 is different. The four operating modes are described in the following text.

#### Mode 0

Putting either Timer into Mode 0 makes it look like an 8048 Timer, which is an 8-bit Counter with a divide-by-32 prescaler. Figure 3 shows the Mode 0 operation.

In this mode, the Timer register is configured as a 13-bit register. As the count rolls over from all 1s to all 0s, it sets the Timer interrupt flag TFn. The counted input is enabled to the Timer when TRn = 1 and either GATE = 0 or  $\overline{\text{INTn}}$  = 1. (Setting GATE = 1 allows the Timer to be controlled by external input  $\overline{\text{INTn}}$ , to facilitate pulse width measurements). TRn is a control bit in the Special Function Register TCON (Figure 4).

The 13-bit register consists of all 8 bits of THn and the lower 5 bits of TLn. The upper 3 bits of TLn are indeterminate and should be ignored. Setting the run flag (TRn) does not clear the registers.

Mode 0 operation is the same for Timer 0 as for Timer 1. There are two different GATE bits, one for Timer 1 (TMOD.7) and one for Timer 0 (TMOD.3).

#### Mode 1

Mode 1 is the same as Mode 0, except that the Timer register is being run with all 16 bits.

#### Mode 2

Mode 2 configures the Timer register as an 8-bit Counter (TLn) with automatic reload, as shown in Figure 5. Overflow from TLn not only sets TFn, but also reloads TLn with the contents of THn, which is preset by software. The reload leaves THn unchanged.

Mode 2 operation is the same for Timer 0 as for Timer 1.

#### Mode 3

Timer 1 in Mode 3 simply holds its count. The effect is the same as setting TR1 = 0.

Timer 0 in Mode 3 establishes TL0 and TH0 as two separate counters. The logic for Mode 3 on Timer 0 is shown in Figure 6. TL0 uses the Timer 0 control bits: C/T, GATE, TR0, and TF0 as well as pin  $\overline{\text{INT0}}$ . TH0 is locked into a timer function (counting machine cycles) and takes over the use of TR1 and TF1 from Timer 1. Thus, TH0 now controls the "Timer 1" interrupt.

Mode 3 is provided for applications requiring an extra 8-bit timer on the counter. With Timer 0 in Mode 3, an 80C51 can look like it has three Timer/Counters. When Timer 0 is in Mode 3, Timer 1 can be turned on and off by switching it out of and into its own Mode 3, or can still be used by the serial port as a baud rate generator, or in fact, in any application not requiring an interrupt.

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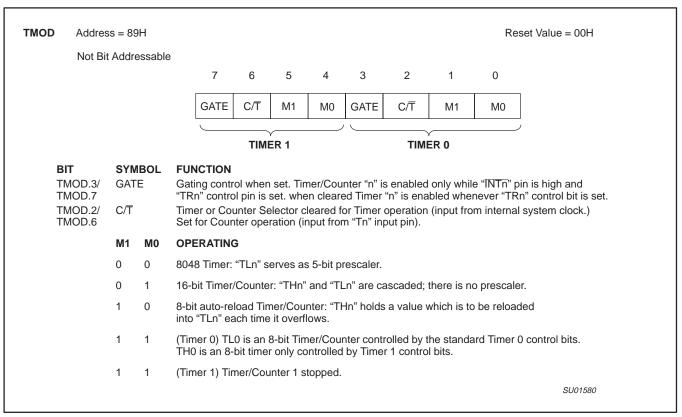


Figure 2. Timer/Counter 0/1 Mode Control (TMOD) Register

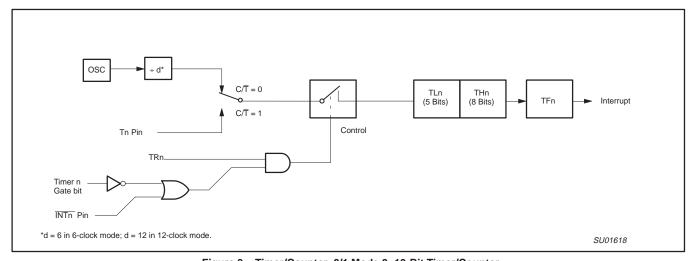


Figure 3. Timer/Counter 0/1 Mode 0: 13-Bit Timer/Counter

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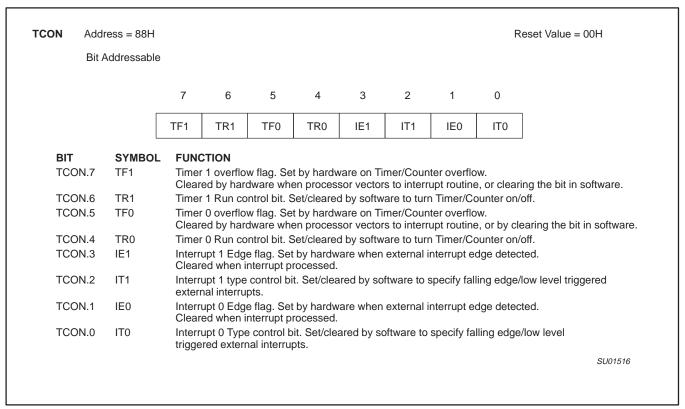


Figure 4. Timer/Counter 0/1 Control (TCON) Register

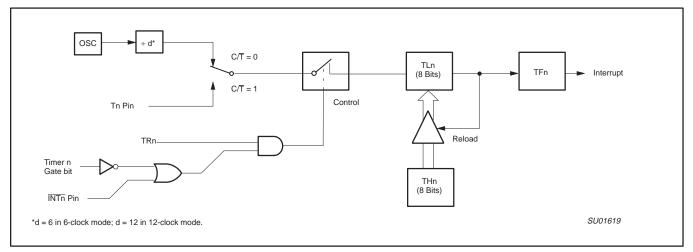


Figure 5. Timer/Counter 0/1 Mode 2: 8-Bit Auto-Reload

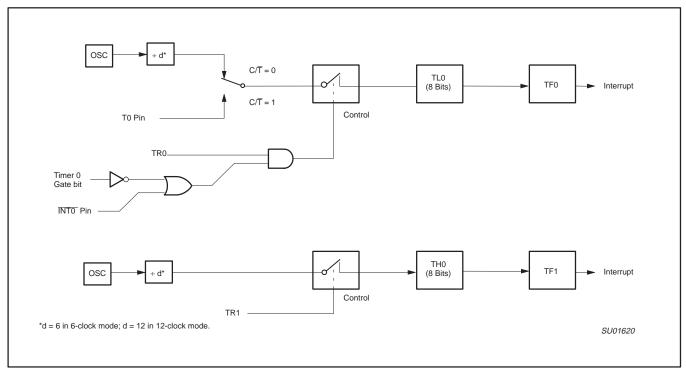


Figure 6. Timer/Counter 0 Mode 3: Two 8-Bit Counters

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#### **TIMER 2 OPERATION**

#### Timer 2

Timer 2 is a 16-bit Timer/Counter which can operate as either an event timer or an event counter, as selected by C/T2 in the special function register T2CON (see Figure 1). Timer 2 has three operating modes: Capture, Auto-reload (up or down counting), and Baud Rate Generator, which are selected by bits in the T2CON as shown in Table 3.

#### Capture Mode

In the capture mode there are two options which are selected by bit EXEN2 in T2CON. If EXEN2=0, then timer 2 is a 16-bit timer or counter (as selected by C/T2 in T2CON) which, upon overflowing sets bit TF2, the timer 2 overflow bit. This bit can be used to generate an interrupt (by enabling the Timer 2 interrupt bit in the IE register). If EXEN2= 1, Timer 2 operates as described above, but with the added feature that a 1- to -0 transition at external input T2EX causes the current value in the Timer 2 registers, TL2 and TH2, to be captured into registers RCAP2L and RCAP2H, respectively. In addition, the transition at T2EX causes bit EXF2 in T2CON to be set, and EXF2 like TF2 can generate an interrupt (which vectors to the same location as Timer 2 overflow interrupt. The Timer 2 interrupt service routine can interrogate TF2 and EXF2 to determine which event caused the interrupt). The capture mode is illustrated in Figure 2 (There is no reload value for TL2 and TH2 in this mode. Even when a capture event occurs from T2EX, the counter keeps on counting T2EX pin transitions or osc/6 pulses (osc/12 in 12-clock mode).).

#### Auto-Reload Mode (Up or Down Counter)

In the 16-bit auto-reload mode, Timer 2 can be configured (as either a timer or counter [C/ $\overline{T}2$  in T2CON]) then programmed to count up or down. The counting direction is determined by bit DCEN (Down

Counter Enable) which is located in the T2MOD register (see Figure 3). When reset is applied the DCEN=0 which means Timer 2 will default to counting up. If DCEN bit is set, Timer 2 can count up or down depending on the value of the T2EX pin.

Figure 4 shows Timer 2 which will count up automatically since DCEN=0. In this mode there are two options selected by bit EXEN2 in T2CON register. If EXEN2=0, then Timer 2 counts up to 0FFFFH and sets the TF2 (Overflow Flag) bit upon overflow. This causes the Timer 2 registers to be reloaded with the 16-bit value in RCAP2L and RCAP2H. The values in RCAP2L and RCAP2H are preset by software means.

If EXEN2=1, then a 16-bit reload can be triggered either by an overflow or by a 1-to-0 transition at input T2EX. This transition also sets the EXF2 bit. The Timer 2 interrupt, if enabled, can be generated when either TF2 or EXF2 are 1.

In Figure 5 DCEN=1 which enables Timer 2 to count up or down. This mode allows pin T2EX to control the direction of count. When a logic 1 is applied at pin T2EX Timer 2 will count up. Timer 2 will overflow at 0FFFFH and set the TF2 flag, which can then generate an interrupt, if the interrupt is enabled. This timer overflow also causes the 16-bit value in RCAP2L and RCAP2H to be reloaded into the timer registers TL2 and TH2.

When a logic 0 is applied at pin T2EX this causes Timer 2 to count down. The timer will underflow when TL2 and TH2 become equal to the value stored in RCAP2L and RCAP2H. Timer 2 underflow sets the TF2 flag and causes 0FFFFH to be reloaded into the timer registers TL2 and TH2.

The external flag EXF2 toggles when Timer 2 underflows or overflows. This EXF2 bit can be used as a 17th bit of resolution if needed. The EXF2 flag does not generate an interrupt in this mode of operation.

	(MSB	3)						(LSB)	
	ТІ	F2 EXF2	RCLK	TCLK	EXEN2	TR2	C/T2	CP/RL2	
Symbol	Position	Name and Sig	nificance	•	•		•	•	
TF2	T2CON.7		Timer 2 overflow flag set by a Timer 2 overflow and must be cleared by software. TF2 will not be set when either RCLK or TCLK = 1.						
EXF2	T2CON.6		nen Timer 2 e. EXF2 mu	interrupt is st be cleare	enabled, EX	(F2 = 1  will)	cause the C	PU to vecto	ition on T2EX and to the Timer 2 t in up/down
RCLK	T2CON.5		Receive clock flag. When set, causes the serial port to use Timer 2 overflow pulses for its receive clock in modes 1 and 3. RCLK = 0 causes Timer 1 overflow to be used for the receive clock.						
TCLK	T2CON.4	Transmit clock in modes 1 and							r its transmit clock k.
EXEN2	T2CON.3	Timer 2 extern transition on Ta ignore events a	2EX if Timer						
TR2	T2CON.2	Start/stop cont	ol for Time	2. A logic 1	starts the ti	mer.			
C/T2	T2CON.1		nternal time	r (OSĆ/6 in	6-clock mod		2 in 12-cloc	k mode)	
CP/RL2	T2CON.0	cleared, auto-r	eloads will onen either R	ccur either	with Timer 2	overflows of	or negative t	transitions at	EXEN2 = 1. When T2EX when ced to auto-reload SU01251

Figure 1. Timer/Counter 2 (T2CON) Control Register

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Table 3. Timer 2 Operating Modes

RCLK + TCLK	CP/RL2	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	Х	1	Baud rate generator
X	Х	0	(off)

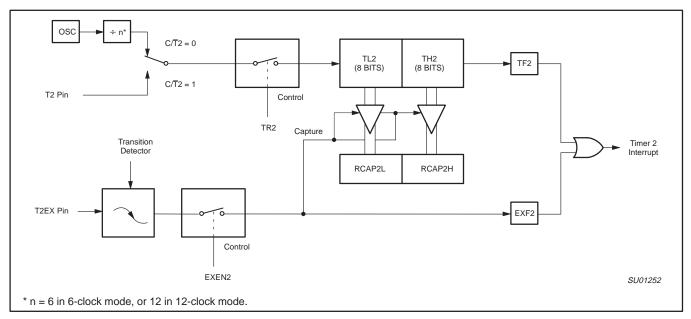


Figure 2. Timer 2 in Capture Mode

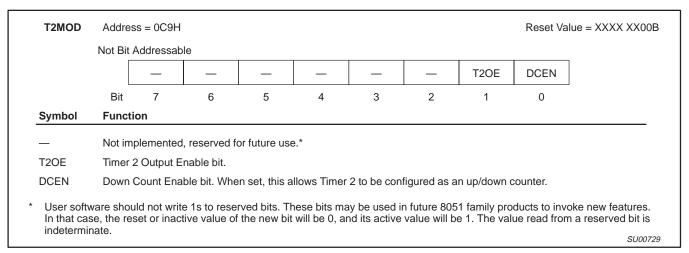


Figure 3. Timer 2 Mode (T2MOD) Control Register

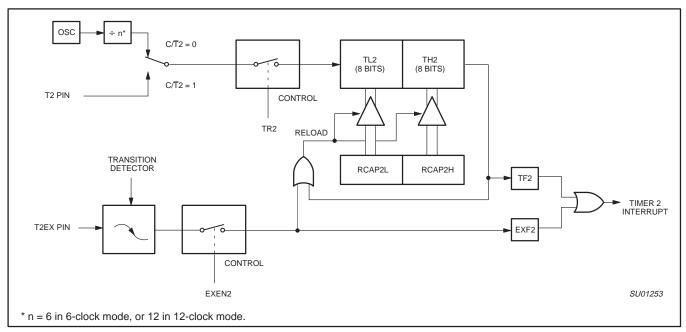


Figure 4. Timer 2 in Auto-Reload Mode (DCEN = 0)

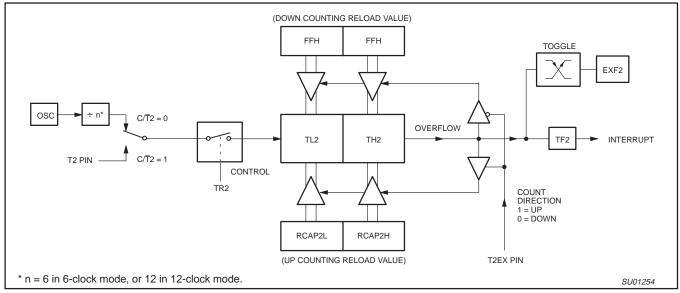


Figure 5. Timer 2 Auto Reload Mode (DCEN = 1)

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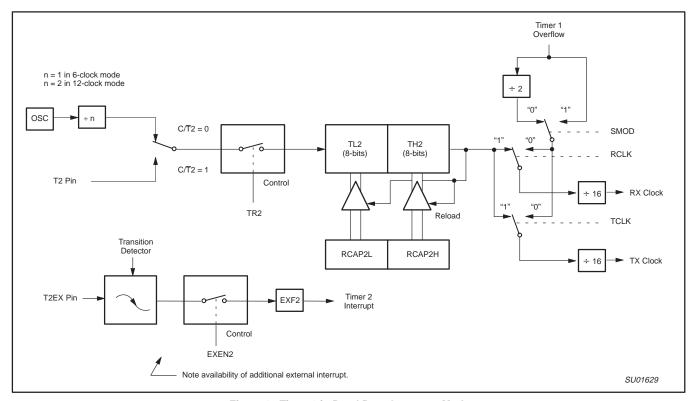


Figure 6. Timer 2 in Baud Rate Generator Mode

Table 4. Timer 2 Generated Commonly Used Baud Rates

Baud	Baud Rate		Tim	er 2
12-clock mode	6-clock mode	Osc Freq	RCAP2H	RCAP2L
375 k	750 k	12 MHz	FF	FF
9.6 k	19.2 k	12 MHz	FF	D9
4.8 k	9.6 k	12 MHz	FF	B2
2.4 k	4.8 k	12 MHz	FF	64
1.2 k	2.4 k	12 MHz	FE	C8
300	600	12 MHz	FB	1E
110	220	12 MHz	F2	AF
300	600	6 MHz	FD	8F
110	220	6 MHz	F9	57

#### **Baud Rate Generator Mode**

Bits TCLK and/or RCLK in T2CON (Table 4) allow the serial port transmit and receive baud rates to be derived from either Timer 1 or Timer 2. When TCLK= 0, Timer 1 is used as the serial port transmit baud rate generator. When TCLK= 1, Timer 2 is used as the serial port transmit baud rate generator. RCLK has the same effect for the serial port receive baud rate. With these two bits, the serial port can have different receive and transmit baud rates – one generated by Timer 1, the other by Timer 2.

Figure 6 shows the Timer 2 in baud rate generation mode. The baud rate generation mode is like the auto-reload mode,in that a rollover in TH2 causes the Timer 2 registers to be reloaded with the 16-bit value in registers RCAP2H and RCAP2L, which are preset by software.

The baud rates in modes 1 and 3 are determined by Timer 2's overflow rate given below:

Modes 1 and 3 Baud Rates = 
$$\frac{\text{Timer 2 Overflow Rate}}{16}$$

The timer can be configured for either "timer" or "counter" operation. In many applications, it is configured for "timer" operation (C/T2=0). Timer operation is different for Timer 2 when it is being used as a baud rate generator.

Usually, as a timer it would increment every machine cycle (i.e.,  $^{1}$ / $_{6}$  the oscillator frequency in 6-clock mode,  $^{1}$ / $_{12}$  the oscillator frequency in 12-clock mode). As a baud rate generator, it increments at the oscillator frequency in 6-clock mode ( $^{OSC}$ / $_{2}$  in 12-clock mode). Thus the baud rate formula is as follows:

Modes 1 and 3 Baud Rates =

\* n = 16 in 6-clock mode 32 in 12-clock mode

Where: (RCAP2H, RCAP2L)= The content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

The Timer 2 as a baud rate generator mode shown in Figure 6, is valid only if RCLK and/or TCLK = 1 in T2CON register. Note that a rollover in TH2 does not set TF2, and will not generate an interrupt. Thus, the Timer 2 interrupt does not have to be disabled when Timer 2 is in the baud rate generator mode. Also if the EXEN2 (T2 external enable flag) is set, a 1-to-0 transition in T2EX (Timer/counter 2 trigger input) will set EXF2 (T2 external flag) but will not cause a reload from (RCAP2H, RCAP2L) to (TH2,TL2). Therefore when Timer 2 is in use as a baud rate generator, T2EX can be used as an additional external interrupt, if needed.

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When Timer 2 is in the baud rate generator mode, one should not try to read or write TH2 and TL2. As a baud rate generator, Timer 2 is incremented every state time (osc/2) or asynchronously from pin T2; under these conditions, a read or write of TH2 or TL2 may not be accurate. The RCAP2 registers may be read, but should not be written to, because a write might overlap a reload and cause write and/or reload errors. The timer should be turned off (clear TR2) before accessing the Timer 2 or RCAP2 registers.

Table 4 shows commonly used baud rates and how they can be obtained from Timer 2.

## **Summary of Baud Rate Equations**

Timer 2 is in baud rate generating mode. If Timer 2 is being clocked through pin T2 (P1.0) the baud rate is:

Baud Rate = 
$$\frac{\text{Timer 2 Overflow Rate}}{16}$$

If Timer 2 is being clocked internally, the baud rate is:

$$\label{eq:Baud_Rate} \begin{aligned} \text{Baud} \;\; & \text{Rate} = \frac{f_{OSC}}{\left[ \; n \;^* \times \left[ 65536 \; - \; \left( \text{RCAP2H}, \text{RCAP2L} \right) \right] \right]} \\ ^* n = & \quad \begin{array}{c} \text{16 in 6-clock mode} \\ \text{32 in 12-clock mode} \end{array} \end{aligned}$$

Where f<sub>OSC</sub>= Oscillator Frequency

To obtain the reload value for RCAP2H and RCAP2L, the above equation can be rewritten as:

RCAP2H, RCAP2L = 
$$65536 - \left(\frac{f_{OSC}}{n^* \times Baud \ Rate}\right)$$

#### **Timer/Counter 2 Set-up**

Except for the baud rate generator mode, the values given for T2CON do not include the setting of the TR2 bit. Therefore, bit TR2 must be set, separately, to turn the timer on. see Table 5 for set-up of Timer 2 as a timer. Also see Table 6 for set-up of Timer 2 as a counter.

Table 5. Timer 2 as a Timer

	T20	CON
MODE	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)
16-bit Auto-Reload	00H	08H
16-bit Capture	01H	09H
Baud rate generator receive and transmit same baud rate	34H	36H
Receive only	24H	26H
Transmit only	14H	16H

### Table 6. Timer 2 as a Counter

	TMOD					
MODE	INTERNAL CONTROL (Note 1)	EXTERNAL CONTROL (Note 2)				
16-bit	02H	0AH				
Auto-Reload	03H	0BH				

#### NOTES:

- Capture/reload occurs only on timer/counter overflow.
- 2. Capture/reload occurs on timer/counter overflow and a 1-to-0 transition on T2EX (P1.1) pin except when Timer 2 is used in the baud rate generator mode.

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## P8xC660X2/661X2

#### FULL-DUPLEX ENHANCED UART

#### Standard UART operation

The serial port is full duplex, meaning it can transmit and receive simultaneously. It is also receive-buffered, meaning it can commence reception of a second byte before a previously received byte has been read from the register. (However, if the first byte still hasn't been read by the time reception of the second byte is complete, one of the bytes will be lost.) The serial port receive and transmit registers are both accessed at Special Function Register SBUF. Writing to SBUF loads the transmit register, and reading SBUF accesses a physically separate receive register.

The serial port can operate in 4 modes:

Mode 0: Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received (LSB first). The baud rate is fixed at 1/12 the oscillator frequency in 12-clock mode or 1/6 the oscillator frequency in 6-clock mode.

Mode 1: 10 bits are transmitted (through TxD) or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in Special Function Register SCON. The baud rate is variable.

Mode 2: 11 bits are transmitted (through TxD) or received (through RxD): start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). On Transmit, the 9th data bit (TB8 in SCON) can be assigned the value of 0 or 1. Or, for example, the parity bit (P, in the PSW) could be moved into TB8. On receive, the 9th data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/32 or 1/64 the oscillator frequency in 12-clock mode or 1/16 or 1/32 the oscillator frequency in 6-clock mode.

Mode 3: 11 bits are transmitted (through TxD) or received (through RxD): a start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable.

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

#### **Multiprocessor Communications**

Modes 2 and 3 have a special provision for multiprocessor communications. In these modes, 9 data bits are received. The 9th one goes into RB8. Then comes a stop bit. The port can be programmed such that when the stop bit is received, the serial port interrupt will be activated only if RB8 = 1. This feature is enabled by setting bit SM2 in SCON. A way to use this feature in multiprocessor systems is as follows:

When the master processor wants to transmit a block of data to one of several slaves, it first sends out an address byte which identifies the target slave. An address byte differs from a data byte in that the 9th bit is 1 in an address byte and 0 in a data byte. With SM2 = 1, no slave will be interrupted by a data byte. An address byte, however, will interrupt all slaves, so that each slave can examine the received byte and see if it is being addressed. The addressed slave will clear its SM2 bit and prepare to receive the data bytes that will be coming.

The slaves that weren't being addressed leave their SM2s set and go on about their business, ignoring the coming data bytes.

SM2 has no effect in Mode 0, and in Mode 1 can be used to check the validity of the stop bit. In a Mode 1 reception, if SM2 = 1, the receive interrupt will not be activated unless a valid stop bit is received.

#### **Serial Port Control Register**

The serial port control and status register is the Special Function Register SCON, shown in Figure 7. This register contains not only the mode selection bits, but also the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

#### **Baud Rates**

The baud rate in Mode 0 is fixed: Mode 0 Baud Rate = Oscillator Frequency / 12 (12-clock mode) or / 6 (6-clock mode). The baud rate in Mode 2 depends on the value of bit SMOD in Special Function Register PCON. If SMOD = 0 (which is the value on reset), and the port pins in 12-clock mode, the baud rate is 1/64 the oscillator frequency. If SMOD = 1, the baud rate is 1/32 the oscillator frequency. In 6-clock mode, the baud rate is 1/32 or 1/16 the oscillator frequency, respectively.

Mode 2 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times \text{(Oscillator Frequency)}$$

Where:

n = 64 in 12-clock mode, 32 in 6-clock mode

The baud rates in Modes 1 and 3 are determined by the Timer 1 or Timer 2 overflow rate.

#### **Using Timer 1 to Generate Baud Rates**

When Timer 1 is used as the baud rate generator (T2CON.RCLK = 0, T2CON.TCLK = 0), the baud rates in Modes 1 and 3 are determined by the Timer 1 overflow rate and the value of SMOD as follows:

Mode 1. 3 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times \text{(Timer 1 Overflow Rate)}$$

Where:

n = 32 in 12-clock mode, 16 in 6-clock mode

The Timer 1 interrupt should be disabled in this application. The Timer itself can be configured for either "timer" or "counter" operation, and in any of its 3 running modes. In the most typical applications, it is configured for "timer" operation, in the auto-reload mode (high nibble of TMOD = 0010B). In that case the baud rate is given by the formula:

Mode 1, 3 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times \frac{\text{Oscillator Frequency}}{12 \times [256-(\text{TH1})]}$$

Where:

n = 32 in 12-clock mode, 16 in 6-clock mode

One can achieve very low baud rates with Timer 1 by leaving the Timer 1 interrupt enabled, and configuring the Timer to run as a 16-bit timer (high nibble of TMOD = 0001B), and using the Timer 1 interrupt to do a 16-bit software reload. Figure 8 lists various commonly used baud rates and how they can be obtained from Timer 1.

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## P8xC660X2/661X2

S	CON	Addres	s = 98H									Reset Value = 00H
		Bit Add	ressable	7	6	5	4	3	2	1	0	_
				SM0	SM1	SM2	REN	TB8	RB8	TI	RI	
Where	e SM0,	SM1 spe	cify the serial po	ort mode	, as foll	ows:						
SM0	SM1	Mode	Description	E	Baud Ra	ate						
0	0	0 0 shift register f <sub>OSC</sub> /12 (12-clock mode) or f <sub>OSC</sub> /6 (6-clock mode)										
0	1	1 1 8-bit UART variable										
1	0	2	9-bit UART		f <sub>OSC</sub> /64	1 or f <sub>OS</sub>	<sub>C</sub> /32 (12	2-clock i	mode) o	r fosc/3	32 or f <sub>OS</sub>	<sub>SC</sub> /16 (6-clock mode)
1	1	3	9-bit UART		variable	Э						
SM2	Enables the multiprocessor communication feature in Modes 2 and 3. In Mode 2 or 3, if SM2 is set to 1, then RI will not be activated if the received 9th data bit (RB8) is 0. In Mode 1, if SM2=1 then RI will not be activated if a valid stop bit was not received. In Mode 0, SM2 should be 0.											
REN	Ena	ables seri	al reception. Set	by soft	ware to	enable	reception	n. Clea	r by sof	tware to	disable	e reception.
TB8	The	The 9th data bit that will be transmitted in Modes 2 and 3. Set or clear by software as desired.										
RB8	In Modes 2 and 3, is the 9th data bit that was received. In Mode 1, it SM2=0, RB8 is the stop bit that was received. In Mode 0, RB8 is not used.											
TI		Transmit interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or at the beginning of the stop bit in the other modes, in any serial transmission. Must be cleared by software.										
RI			, ,							,	halfway	through the stop bit time in the other
	mod	des, in an	y serial reception	n (exce	pt see S	SM2). N	lust be o	cleared	by softw	are.		SU01626

Figure 7. Serial Port Control (SCON) Register

	Baud Rate			SMOD	Timer 1			
Mode	12-clock mode	6-clock mode	fosc	SMOD	C/T	Mode	Reload Value	
Mode 0 Max	1.67 MHz	3.34 MHz	20 MHz	Х	Х	Х	Х	
Mode 2 Max	625 k	1250 k	20 MHz	1	Х	Х	X	
Mode 1, 3 Max	104.2 k	208.4 k	20 MHz	1	0	2	FFH	
Mode 1, 3	19.2 k	38.4 k	11.059 MHz	1	0	2	FDH	
	9.6 k	19.2 k	11.059 MHz	0	0	2	FDH	
	4.8 k	9.6 k	11.059 MHz	0	0	2	FAH	
	2.4 k	4.8 k	11.059 MHz	0	0	2	F4H	
	1.2 k	2.4 k	11.059 MHz	0	0	2	E8H	
	137.5	275	11.986 MHz	0	0	2	1DH	
	110	220	6 MHz	0	0	2	72H	
	110	220	12 MHz	0	0	1	FEEBH	

Figure 8. Timer 1 Generated Commonly Used Baud Rates

#### More About Mode 0

Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received: 8 data bits (LSB first). The baud rate is fixed a 1/12 the oscillator frequency (12-clock mode) or 1/6 the oscillator frequency (6-clock mode).

Figure 9 shows a simplified functional diagram of the serial port in Mode 0, and associated timing.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal at S6P2 also loads a 1 into the 9th position of the transmit shift register and tells the TX Control block to commence a transmission. The internal timing is such that one full machine cycle will elapse between "write to SBUF" and activation of SEND.

SEND enables the output of the shift register to the alternate output function line of P3.0 and also enable SHIFT CLOCK to the alternate output function line of P3.1. SHIFT CLOCK is LOW during S3, S4, and S5 of every machine cycle, and HIGH during S6, S1, and S2. At

S6P2 of every machine cycle in which SEND is active, the contents of the transmit shift are shifted to the right one position.

As data bits shift out to the right, zeros come in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position, is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control block to do one last shift and then deactivate SEND and set T1. Both of these actions occur at S1P1 of the 10th machine cycle after "write to SBUF."

Reception is initiated by the condition REN = 1 and R1 = 0. At S6P2 of the next machine cycle, the RX Control unit writes the bits 11111110 to the receive shift register, and in the next clock phase activates RECEIVE.

RECEIVE enable SHIFT CLOCK to the alternate output function line of P3.1. SHIFT CLOCK makes transitions at S3P1 and S6P1 of every machine cycle. At S6P2 of every machine cycle in which RECEIVE is active, the contents of the receive shift register are

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shifted to the left one position. The value that comes in from the right is the value that was sampled at the P3.0 pin at S5P2 of the same machine cycle.

As data bits come in from the right, 1s shift out to the left. When the 0 that was initially loaded into the rightmost position arrives at the leftmost position in the shift register, it flags the RX Control block to do one last shift and load SBUF. At S1P1 of the 10th machine cycle after the write to SCON that cleared RI, RECEIVE is cleared as RI is set.

#### More About Mode 1

Ten bits are transmitted (through TxD), or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in SCON. In the 80C51 the baud rate is determined by the Timer 1 or Timer 2 overflow rate.

Figure 10 shows a simplified functional diagram of the serial port in Mode 1, and associated timings for transmit receive.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal also loads a 1 into the 9th bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission actually commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. (Thus, the bit times are synchronized to the divide-by-16 counter, not to the "write to SBUF" signal.)

The transmission begins with activation of SEND which puts the start bit at TxD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TxD. The first shift pulse occurs one bit time after that.

As data bits shift out to the right, zeros are clocked in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control unit to do one last shift and then deactivate SEND and set TI. This occurs at the 10th divide-by-16 rollover after "write to SBUF."

Reception is initiated by a detected 1-to-0 transition at RxD. For this purpose RxD is sampled at a rate of 16 times whatever baud rate has been established. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written into the input shift register. Resetting the divide-by-16 counter aligns its rollovers with the boundaries of the incoming bit times.

The 16 states of the counter divide each bit time into 16ths. At the 7th, 8th, and 9th counter states of each bit time, the bit detector samples the value of RxD. The value accepted is the value that was seen in at least 2 of the 3 samples. This is done for noise rejection. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit goes back to looking for another 1-to-0 transition. This is to provide rejection of false start bits. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame will proceed.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in mode 1 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI. The signal to load SBUF and RB8, and to set RI, will be generated if, and only if, the following conditions are met at the time the final shift pulse is generated.:

- 1. R1 = 0, and
- 2. Either SM2 = 0, or the received stop bit = 1.

If either of these two conditions is not met, the received frame is irretrievably lost. If both conditions are met, the stop bit goes into RB8, the 8 data bits go into SBUF, and RI is activated. At this time,

whether the above conditions are met or not, the unit goes back to looking for a 1-to-0 transition in RxD.

#### More About Modes 2 and 3

Eleven bits are transmitted (through TxD), or received (through RxD): a start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). On transmit, the 9th data bit (TB8) can be assigned the value of 0 or 1. On receive, the 9the data bit goes into RB8 in SCON. The baud rate is programmable to either 1/32 or 1/64 (12-clock mode) or 1/16 or 1/32 the oscillator frequency (6-clock mode) the oscillator frequency in Mode 2. Mode 3 may have a variable baud rate generated from Timer 1 or Timer 2.

Figures 11 and 12 show a functional diagram of the serial port in Modes 2 and 3. The receive portion is exactly the same as in Mode 1. The transmit portion differs from Mode 1 only in the 9th bit of the transmit shift register.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal also loads TB8 into the 9th bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. (Thus, the bit times are synchronized to the divide-by-16 counter, not to the "write to SBUF" signal.)

The transmission begins with activation of SEND, which puts the start bit at TxD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TxD. The first shift pulse occurs one bit time after that. The first shift clocks a 1 (the stop bit) into the 9th bit position of the shift register. Thereafter, only zeros are clocked in. Thus, as data bits shift out to the right, zeros are clocked in from the left. When TB8 is at the output position of the shift register, then the stop bit is just to the left of TB8, and all positions to the left of that contain zeros. This condition flags the TX Control unit to do one last shift and then deactivate SEND and set TI. This occurs at the 11th divide-by-16 rollover after "write to SUBF."

Reception is initiated by a detected 1-to-0 transition at RxD. For this purpose RxD is sampled at a rate of 16 times whatever baud rate has been established. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written to the input shift register.

At the 7th, 8th, and 9th counter states of each bit time, the bit detector samples the value of R-D. The value accepted is the value that was seen in at least 2 of the 3 samples. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit goes back to looking for another 1-to-0 transition. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame will proceed.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in Modes 2 and 3 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI.

The signal to load SBUF and RB8, and to set RI, will be generated if, and only if, the following conditions are met at the time the final shift pulse is generated.

- 1. RI = 0, and
- 2. Either SM2 = 0, or the received 9th data bit = 1.

If either of these conditions is not met, the received frame is irretrievably lost, and RI is not set. If both conditions are met, the received 9th data bit goes into RB8, and the first 8 data bits go into SBUF. One bit time later, whether the above conditions were met or not, the unit goes back to looking for a 1-to-0 transition at the RxD input.

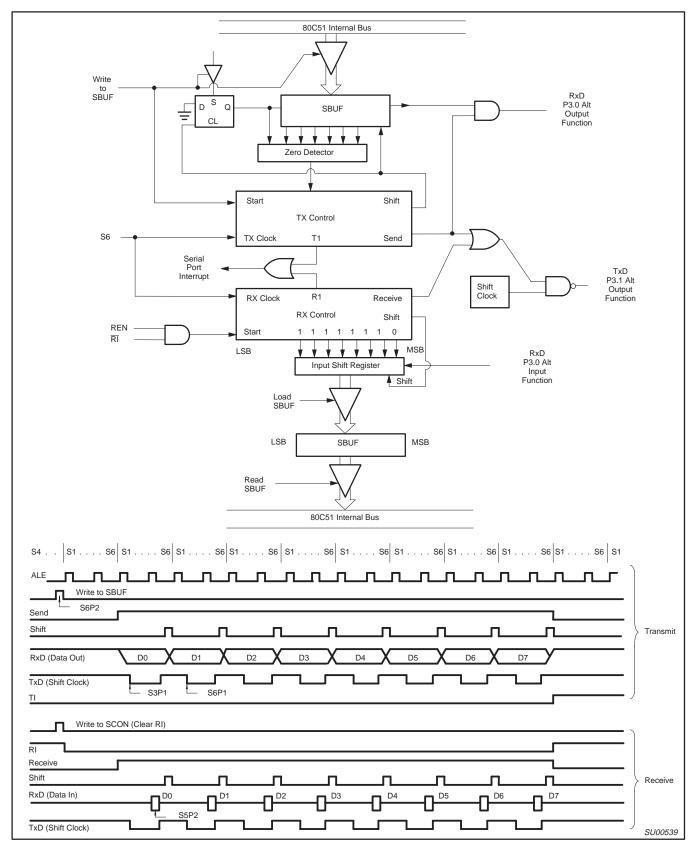


Figure 9. Serial Port Mode 0

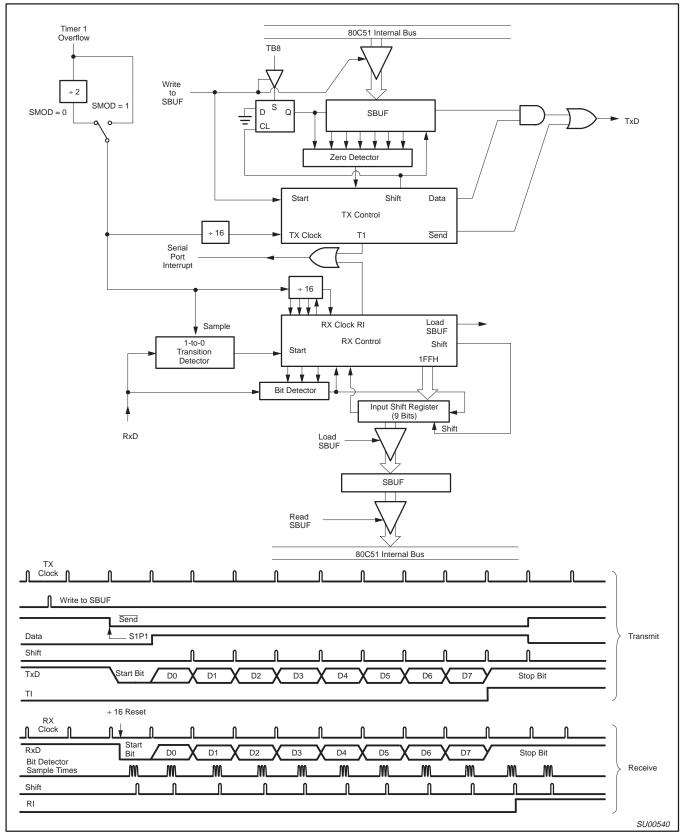


Figure 10. Serial Port Mode 1

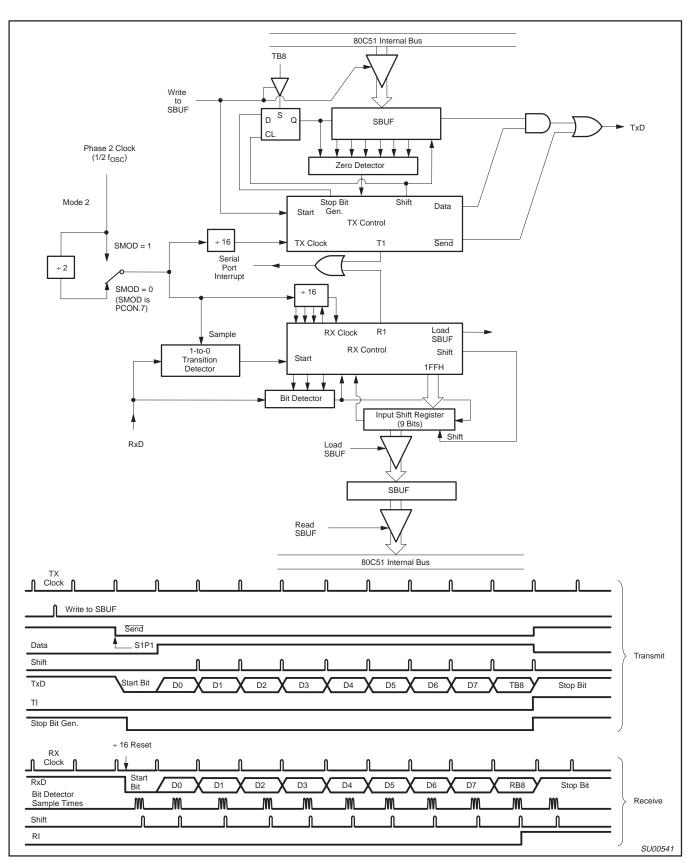


Figure 11. Serial Port Mode 2

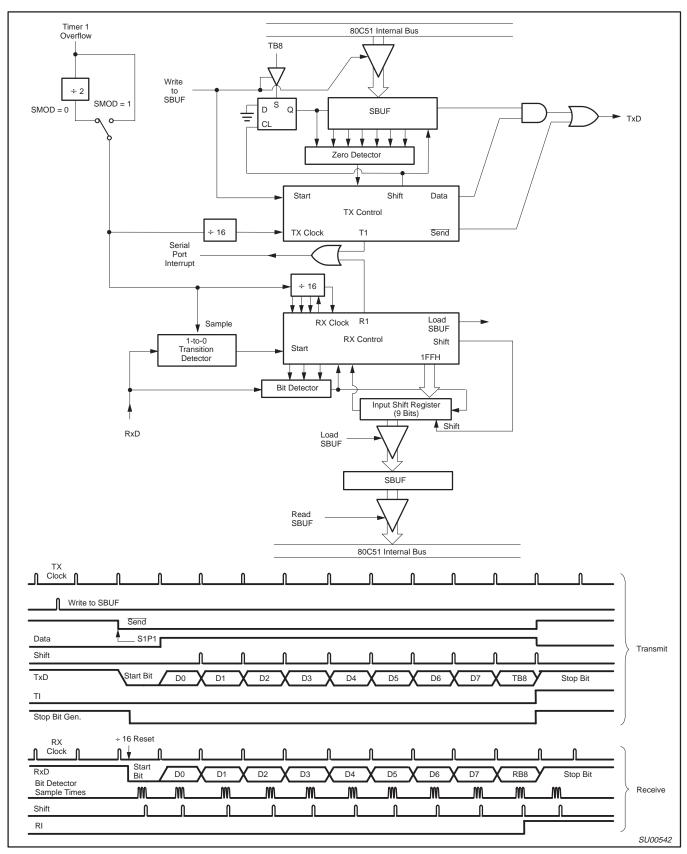


Figure 12. Serial Port Mode 3

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## P8xC660X2/661X2

#### **Enhanced Features**

The UART operates in all of the usual modes that are described in the first section of Data Handbook IC20, 80C51-Based 8-Bit Microcontrollers. In addition the UART can perform framing error detect by looking for missing stop bits, and automatic address recognition. The UART also fully supports multiprocessor communication as does the standard 80C51 UART.

When used for framing error detect the UART looks for missing stop bits in the communication. A missing bit will set the FE bit in the SCON register. The FE bit shares the SCON.7 bit with SM0 and the function of SCON.7 is determined by PCON.6 (SMOD0) (see Figure 7). If SMOD0 is set then SCON.7 functions as FE. SCON.7 functions as SM0 when SMOD0 is cleared. When used as FE SCON.7 can only be cleared by software. Refer to Figure 13.

#### **Automatic Address Recognition**

Automatic Address Recognition is a feature which allows the UART to recognize certain addresses in the serial bit stream by using hardware to make the comparisons. This feature saves a great deal of software overhead by eliminating the need for the software to examine every serial address which passes by the serial port. This feature is enabled by setting the SM2 bit in SCON. In the 9 bit UART modes, mode 2 and mode 3, the Receive Interrupt flag (RI) will be automatically set when the received byte contains either the "Given" address or the "Broadcast" address. The 9-bit mode requires that the 9th information bit is a 1 to indicate that the received information is an address and not data. Automatic address recognition is shown in Figure 14.

The 8 bit mode is called Mode 1. In this mode the RI flag will be set if SM2 is enabled and the information received has a valid stop bit following the 8 address bits and the information is either a Given or Broadcast address.

Mode 0 is the Shift Register mode and SM2 is ignored.

Using the Automatic Address Recognition feature allows a master to selectively communicate with one or more slaves by invoking the Given slave address or addresses. All of the slaves may be contacted by using the Broadcast address. Two special Function Registers are used to define the slave's address, SADDR, and the address mask, SADEN. SADEN is used to define which bits in the SADDR are to b used and which bits are "don't care". The SADEN mask can be logically ANDed with the SADDR to create the "Given" address which the master will use for addressing each of the slaves. Use of the Given address allows multiple slaves to be recognized while excluding others. The following examples will help to show the versatility of this scheme:

Slave 0 SADEN =

SADDR = 1100 0000 1111 1101 Given 1100 00X0 Slave 1 SADDR = 1100 0000 SADEN = 1111 1110 Given 1100 000X

In the above example SADDR is the same and the SADEN data is used to differentiate between the two slaves. Slave 0 requires a 0 in bit 0 and it ignores bit 1. Slave 1 requires a 0 in bit 1 and bit 0 is ignored. A unique address for Slave 0 would be 1100 0010 since slave 1 requires a 0 in bit 1. A unique address for slave 1 would be 1100 0001 since a 1 in bit 0 will exclude slave 0. Both slaves can be selected at the same time by an address which has bit 0 = 0 (for slave 0) and bit 1 = 0 (for slave 1). Thus, both could be addressed with 1100 0000.

In a more complex system the following could be used to select slaves 1 and 2 while excluding slave 0:

Slave 0	SADDR	=	1100	0000
	SADEN	=	<u>1111</u>	1001
	Given	=	1100	0XX0
Slave 1	SADDR	=	1110	0000
	SADEN	=	1111	1010
	Given	=	1110	0X0X
Slave 2	SADDR	=	1110	0000
	SADEN	=	<u>1111</u>	1100
	Given	=	1110	00XX

In the above example the differentiation among the 3 slaves is in the lower 3 address bits. Slave 0 requires that bit 0 = 0 and it can be uniquely addressed by 1110 0110. Slave 1 requires that bit 1 = 0 and it can be uniquely addressed by 1110 and 0101. Slave 2 requires that bit 2 = 0 and its unique address is 1110 0011. To select Slaves 0 and 1 and exclude Slave 2 use address 1110 0100, since it is necessary to make bit 2 = 1 to exclude slave 2.

The Broadcast Address for each slave is created by taking the logical OR of SADDR and SADEN. Zeros in this result are trended as don't-cares. In most cases, interpreting the don't-cares as ones, the broadcast address will be FF hexadecimal.

Upon reset SADDR (SFR address 0A9H) and SADEN (SFR address 0B9H) are leaded with 0s. This produces a given address of all "don't cares" as well as a Broadcast address of all "don't cares". This effectively disables the Automatic Addressing mode and allows the microcontroller to use standard 80C51 type UART drivers which do not make use of this feature.

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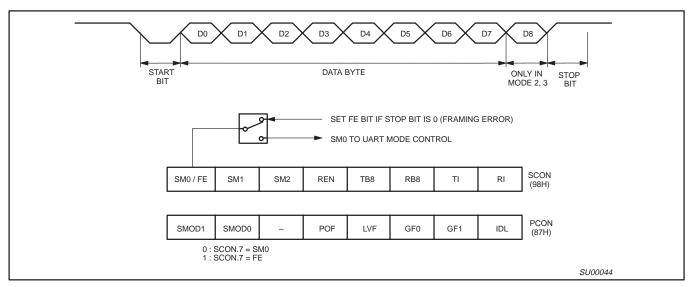


Figure 13. UART Framing Error Detection

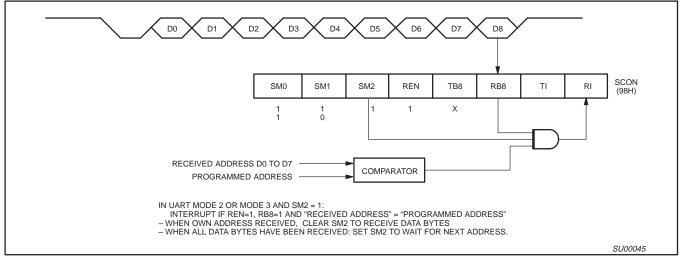


Figure 14. UART Multiprocessor Communication, Automatic Address Recognition

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## P8xC660X2/661X2

#### SIO1 and SIO2, I<sup>2</sup>C Serial I/O

The I<sup>2</sup>C-bus is a simple bi-directional 2-wire bus to transfer information between devices connected to the bus. The main features of the bus are:

- Only two bus lines are required: a serial clock line (SCL) and a serial data line (SDA).
- Bi-directional data transfer between masters and slaves.
- Each device connected to the bus is software addressable by a unique address.
- Masters can operate as Master-transmitter or as Master-receiver.
- It is a true multi-master bus (no central master) and includes collision detection and arbitration to prevent data corruption if two or more masters simultaneously initiate data transfer.
- Serial clock synchronization allows devices with different bit rates to communicate via the same serial bus.
- Serial clock synchronization can be used as a handshake mechanism to suspend and resume serial transfer.
- Devices can be added to or removed from an I<sup>2</sup>C-bus system without affecting any other device on the bus.
- Fault diagnostics and debugging are simple; malfunctions can be immediately traced.

For more information see the Philips publication "The I<sup>2</sup>C-Bus Specification", especially for detailed descriptions of the Fast and the Standard data-transfer modes. Also, refer to the data sheets for the 8xC552, the 8xC554, the 8xC557, and the 8xC65x.

The SIO1 I<sup>2</sup>C serial port interface has a selectable bi-directional data-transfer mode, either the 400Kbit/s Fast-mode or the 100Kbit/s Standard-mode. In the Fast-mode, the port performance and the register definitions are identical to those of the 8xC557 devices, and in the Standard-mode (the reset default), they are identical to those of the 8xC652, 8xC654, 8xC552, and 8xC554 devices.

The Fast-mode is functionally the same as the Standard-mode except for the bit rate selection (see Tables 7 and 8), the timing of the SCL and SDA signals (see the I<sup>2</sup>C electrical characteristics), and the output slew-rate control. The Fast-mode allows up to a four-fold bit-rate increase over that of the Standard-mode, and yet, it is downward compatible with the Standard-mode, i.e. it can be used in a 0 to 100Kbit/s bus system.

The SCL serial port for the clock line of the  $I^2C$  bus is an alternate function of the P1.6 port pin, and the SDA serial port for the data line of the  $I^2C$  bus is an alternate function of the P1.7 port pin. Consequently, these 2 port pins are open drain outputs (no pull-ups), and the output latches of P1.6 and P1.7 must be set to logic 1 in order to enable the SIO1 outputs.

The second I<sup>2</sup>C serial port of the 8xC661X2, SIO2, has the 400Kbit/s Fast data-transfer mode only and selectable slew-rate control of the output pins. It also has the same port performance and register definitions as those of the 8xC557. The SCL1 and SDA1 serial ports have dedicated pins with open-drain outputs and Schmitt-trigger inputs.

There is an analog circuit for controlling the turn-on and turn-off rates of the output pull-down (slew-rate control circuit) which is required to meet the electrical specifications of the Fast-mode under nominal conditions (5 V). To achieve the maximum slew-rates, the

circuit must be disabled. For the SIO1 serial port, the slew-rate control circuits for both the SCL and SDA pins are disabled in the Standard mode (maximum slew-rate), and they are enabled in the Fast-mode. For the SIO2 serial port, the slew-rate control circuits for both pins are enabled by reset, but the Slew-Rate Disable bit (SRD bit) in the AUXR Register disables the slew-rate circuits for both the SCL1 and SDA1 pins when set for maximum slew-rates. This feature of the SIO2 slew-rate control is very useful for higher bus loads, higher temperatures and lower voltages that cause additional decreases in slew-rates.

All of the functional descriptions discussed below apply to both the SIO1 and the SIO2  $I^2C$  serial ports although the text may refer to the SIO1 only. See page 10 for the corresponding SIO2 register addresses.

The  $I^2C$  on-chip logic performs a byte oriented data transfer, clock generation, address recognition and bus control arbitration, and interfaces to the external  $I^2C$ -bus via the two port pins SCL and SDA. It meets the  $I^2C$ -bus specification and supports all transfer modes (other than the low-speed mode) from-and-to the  $I^2C$ -bus. The logic handles byte transfers autonomously. It also keeps track of serial transfers, and a status register (SxSTA) reflects the status of the SIOx logic and the  $I^2C$ -bus.

The CPU interfaces to the logic of each of the two  $I^2$ Cs via the following four Special Function Registers (where x=1,2):

- SxCON: Control register, bit addressable by the CPU.
- SxSTA: Status register whose contents may be used as a vector to service routines.
- SxDAT: Data shift register; the data byte is stable as long as the SI bit = 1 (SxCON.3).
- SxADR: Slave address register; its LSB enables / disables general call address recognition.

A typical I<sup>2</sup>C-bus configuration is shown in Figure 15, and Figure 16 shows how a data transfer is accomplished on the bus. Depending on the state of the direction bit (R/W), two types of data transfers are possible on the I<sup>2</sup>C-bus:

- Data transfer from a master transmitter to a slave receiver. The first byte transmitted by the master is the slave address. Next follows a number of data bytes. The slave returns an acknowledge bit after each received byte.
- Data transfer from a slave transmitter to a master receiver. The
  first byte (the slave address) is transmitted by the master. The
  slave then returns an acknowledge bit. Next follows the data
  bytes transmitted by the slave to the master. The master returns
  an acknowledge bit after all received bytes other than the last
  byte. At the end of the last received byte, a "not acknowledge" is
  returned.

The master device generates all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the I<sup>2</sup>C bus will not be released.

**Modes of Operation:** The on-chip SIO1 logic may operate in the following four modes:

- 1. Master Transmitter Mode:
  - Serial data output through P1.7/SDA while P1.6/SCL outputs the serial clock. The first byte transmitted contains the slave address

of the receiving device (7 bytes) and the data direction bit. In this case the data direction bit (R/W) will be logic 0, and we say that a "W" is transmitted. Thus the first byte transmitted is SLA+W. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an acknowledge bit is received. START and STOP conditions are output to indicate the beginning and the end of a serial transfer.

#### 2. Master Receiver Mode:

The first byte transmitted contains the slave address of the transmitting device (7 bits) and the data direction bit. In this case, the data direction bit ( $R/\overline{W}$ ) will be logic 1, and we say that an "R" is transmitted. Thus the first byte transmitted is SLA+R. Serial data is received via P1.7/SDA while P1.6/SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received an acknowledge bit is transmitted. START and STOP conditions are output to indicate the beginning and end of a serial transfer.

#### 3. Slave Receiver Mode:

Serial data and the serial clock are received through P1.7/SDA and P1.6/SCL. After each byte is received, an acknowledge bit is

transmitted. START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the slave address and direction bit.

#### 4. Slave Transmitter Mode:

The first byte is received and handled as in the slave receiver mode. However, in this mode, the direction bit will indicate that the transfer direction is reversed. Serial data is transmitted via P1.7/SDA while the serial clock is input through P1.6/SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer.

In a given application, SIO1 may operate as a master and as a slave. In the slave mode, the SIO1 hardware looks for its own slave address and the general call address. If one of these addresses is detected, an interrupt is requested. When the microcontroller wishes to become the bus master, the hardware waits until the bus is free before the master mode is entered so that a possible slave action is not interrupted. If bus arbitration is lost in the master mode, SIO1 switches to the slave mode immediately and can detect its own slave address in the same serial transfer.

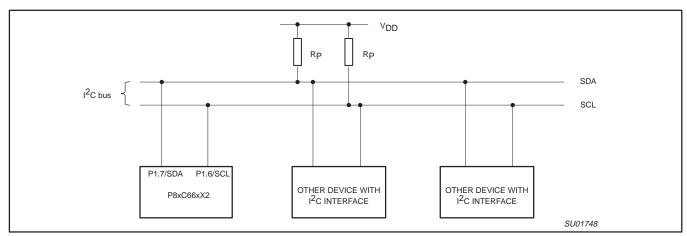


Figure 15. Typical I<sup>2</sup>C Bus Configuration

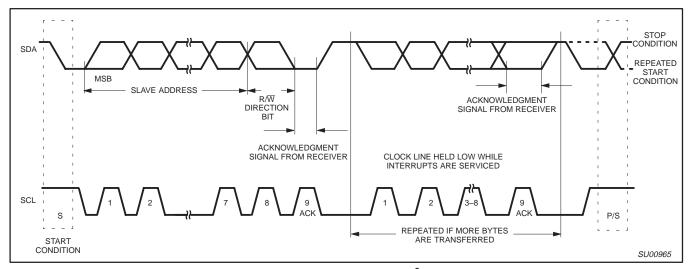


Figure 16. Data Transfer on the I<sup>2</sup>C Bus

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## P8xC660X2/661X2

**SIO1 Implementation and Operation:** Figure 17 shows how the on-chip I<sup>2</sup>C bus interface is implemented, and the following text describes the individual blocks.

#### INPUT FILTERS AND OUTPUT STAGES

The input filters have  $I^2C$  compatible input levels. If the input voltage is less than 1.5 V, the input logic level is interpreted as 0; if the input voltage is greater than 3.0 V, the input logic level is interpreted as 1. Input signals are synchronized with the internal clock ( $f_{OSC}/4$ ), and spikes shorter than three oscillator periods are filtered out.

The output stages consist of open drain transistors that can sink 3 mA at  $V_{OUT}$  < 0.4 V. These open drain outputs do not have clamping diodes to  $V_{DD}$ . Thus, if the device is connected to the I<sup>2</sup>C bus and  $V_{DD}$  is switched off, the I<sup>2</sup>C bus is not affected.

#### ADDRESS REGISTER, S1ADR

This 8-bit special function register may be loaded with the 7-bit slave address (7 most significant bits) to which SIO1 will respond when programmed as a slave transmitter or receiver. The LSB (GC) is used to enable general call address (00H) recognition.

#### **COMPARATOR**

The comparator compares the received 7-bit slave address with its own slave address (7 most significant bits in S1ADR). It also compares the first received 8-bit byte with the general call address (00H). If an equality is found, the appropriate status bits are set and an interrupt is requested.

#### SHIFT REGISTER, S1DAT

This 8-bit special function register contains a byte of serial data to be transmitted or a byte which has just been received. Data in S1DAT is always shifted from right to left; the first bit to be transmitted is the MSB (bit 7) and, after a byte has been received, the first bit of received data is located at the MSB of S1DAT. While data is being shifted out, data on the bus is simultaneously being shifted in; S1DAT always contains the last byte present on the bus. Thus, in the event of lost arbitration, the transition from master transmitter to slave receiver is made with the correct data in S1DAT.

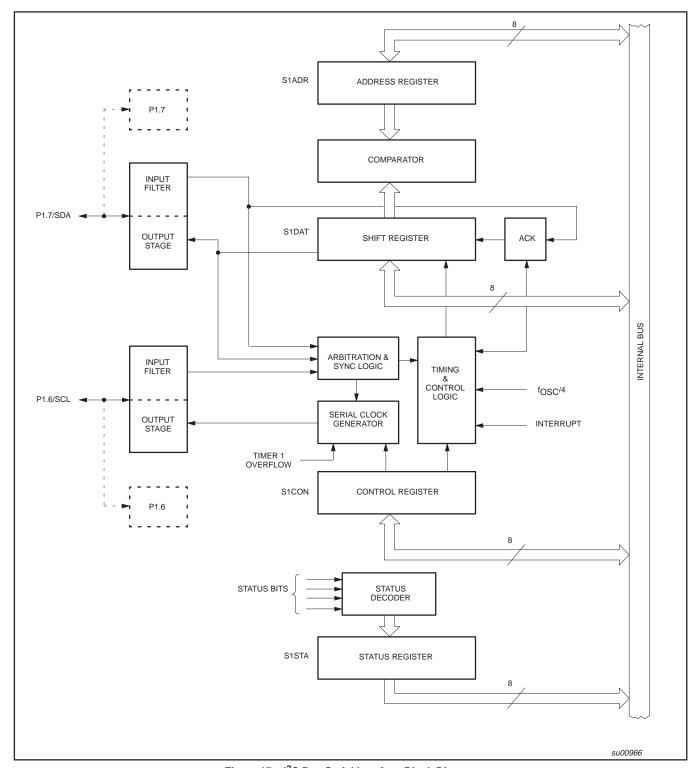


Figure 17.  $\,$  I $^2$ C Bus Serial Interface Block Diagram

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ARBITRATION AND SYNCHRONIZATION LOGIC

In the master transmitter mode, the arbitration logic checks that every transmitted logic 1 actually appears as a logic 1 on the I<sup>2</sup>C bus. If another device on the bus overrules a logic 1 and pulls the SDA line LOW, arbitration is lost, and SIO1 immediately changes from master transmitter to slave receiver. SIO1 will continue to output clock pulses (on SCL) until transmission of the current serial byte is complete.

Arbitration may also be lost in the master receiver mode. Loss of arbitration in this mode can only occur while SIO1 is returning a "not acknowledge: (logic 1) to the bus. Arbitration is lost when another device on the bus pulls this signal LOW. Since this can occur only at the end of a serial byte, SIO1 generates no further clock pulses. Figure 18 shows the arbitration procedure.

The synchronization logic will synchronize the serial clock generator with the clock pulses on the SCL line from another device. If two or more master devices generate clock pulses, the "mark" duration is determined by the device that generates the shortest "marks," and the "space" duration is determined by the device that generates the longest "spaces." Figure 19 shows the synchronization procedure.

A slave may stretch the space duration to slow down the bus master. The space duration may also be stretched for handshaking purposes. This can be done after each bit or after a complete byte transfer. SIO1 will stretch the SCL space duration after a byte has been transmitted or received and the acknowledge bit has been transferred. The serial interrupt flag (SI) is set, and the stretching continues until the serial interrupt flag is cleared.

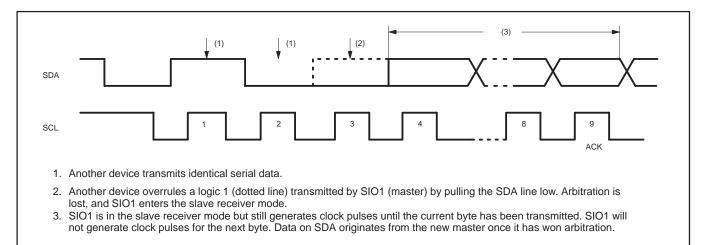
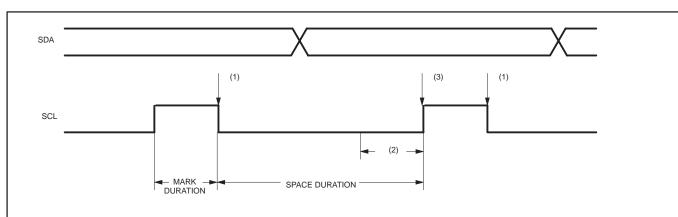


Figure 18. Arbitration Procedure



- 1. Another service pulls the SCL line low before the SIO1 "mark" duration is complete. The serial clock generator is immediately reset and commences with the "space" duration by pulling SCL low.
- Another device still pulls the SCL line low after SIO1 releases SCL. The serial clock generator is forced into the wait state until the SCL line is released.
- 3. The SCL line is released, and the serial clock generator commences with the mark duration.

SU00968

SU00967

Figure 19. Serial Clock Synchronization

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#### SERIAL CLOCK GENERATOR

This programmable clock pulse generator provides the SCL clock pulses when SIO1 is in the master transmitter or master receiver mode. It is switched off when SIO1 is in a slave mode. In standard speed mode, the programmable output clock frequencies are:  $f_{\rm OSC}/120,\,f_{\rm OSC}/9600,$  and the Timer 1 overflow rate divided by eight. The output clock pulses have a 50% duty cycle unless the clock generator is synchronized with other SCL clock sources as described above.

#### TIMING AND CONTROL

The timing and control logic generates the timing and control signals for serial byte handling. This logic block provides the shift pulses for S1DAT, enables the comparator, generates and detects start and stop conditions, receives and transmits acknowledge bits, controls the master and slave modes, contains interrupt request logic, and monitors the  $\rm l^2C$  bus status.

#### CONTROL REGISTER, S1CON

This 7-bit special function register is used by the microcontroller to control the following SIO1 functions: start and restart of a serial transfer, termination of a serial transfer, bit rate, address recognition, and acknowledgment.

#### STATUS DECODER AND STATUS REGISTER

The status decoder takes all of the internal status bits and compresses them into a 5-bit code. This code is unique for each I<sup>2</sup>C bus status. The 5-bit code may be used to generate vector addresses for fast processing of the various service routines. Each service routine processes a particular bus status. There are 26 possible bus states if all four modes of SIO1 are used. The 5-bit status code is latched into the five most significant bits of the status register when the serial interrupt flag is set (by hardware) and remains stable until the interrupt flag is cleared by software. The three least significant bits of the status register are always zero. If the status code is used as a vector to service routines, then the routines are displaced by eight address locations. Eight bytes of code is sufficient for most of the service routines (see the software example in this section).

The Four SIO1 Special Function Registers: The microcontroller interfaces to SIO1 via four special function registers. These four SFRs (S1ADR, S1DAT, S1CON, and S1STA) are described individually in the following sections.

The Address Register, S1ADR: The CPU can read from and write to this 8-bit, directly addressable SFR. S1ADR is not affected by the SIO1 hardware. The contents of this register are irrelevant when SIO1 is in a master mode. In the slave modes, the seven most significant bits must be loaded with the microcontroller's own slave address, and, if the least significant bit is set, the general call address (00H) is recognized; otherwise it is ignored.

	7	6	5	4	3	2	1	0		
S1ADR (DBH)	Х	Х	Х	Х	Х	Х	Х	GC		
	own slave address									

The most significant bit corresponds to the first bit received from the  $I^2C$  bus after a start condition. A logic 1 in S1ADR corresponds to a HIGH level on the  $I^2C$  bus, and a logic 0 corresponds to a LOW level on the bus.

**The Data Register, S1DAT:** S1DAT contains a byte of serial data to be transmitted or a byte which has just been received. The CPU can

read from and write to this 8-bit, directly addressable SFR while it is not in the process of shifting a byte. This occurs when SIO1 is in a defined state and the serial interrupt flag is set. Data in S1DAT remains stable as long as SI is set. Data in S1DAT is always shifted from right to left: the first bit to be transmitted is the MSB (bit 7), and, after a byte has been received, the first bit of received data is located at the MSB of S1DAT. While data is being shifted out, data on the bus is simultaneously being shifted in; S1DAT always contains the last data byte present on the bus. Thus, in the event of lost arbitration, the transition from master transmitter to slave receiver is made with the correct data in S1DAT.

	7	6	5	4	3	2	1	0		
S1DAT (DAH)	SD7	SD6	SD5	SD4	SD3	SD2	SD1	SD0		
shift direction										

SD7 - SD0:

Eight bits to be transmitted or just received. A logic 1 in S1DAT corresponds to a HIGH level on the I<sup>2</sup>C bus, and a logic 0 corresponds to a LOW level on the bus. Serial data shifts through S1DAT from right to left. Figure 20 shows how data in S1DAT is serially transferred to and from the SDA line.

S1DAT and the ACK flag form a 9-bit shift register which shifts in or shifts out an 8-bit byte, followed by an acknowledge bit. The ACK flag is controlled by the SIO1 hardware and cannot be accessed by the CPU. Serial data is shifted through the ACK flag into S1DAT on the rising edges of serial clock pulses on the SCL line. When a byte has been shifted into S1DAT, the serial data is available in S1DAT, and the acknowledge bit is returned by the control logic during the ninth clock pulse. Serial data is shifted out from S1DAT via a buffer (BSD7) on the falling edges of clock pulses on the SCL line.

When the CPU writes to S1DAT, BSD7 is loaded with the content of S1DAT.7, which is the first bit to be transmitted to the SDA line (see Figure 21). After nine serial clock pulses, the eight bits in S1DAT will have been transmitted to the SDA line, and the acknowledge bit will be present in ACK. Note that the eight transmitted bits are shifted back into S1DAT.

The Control Register, S1CON: The CPU can read from and write to this 8-bit, directly addressable SFR. Two bits are affected by the SIO1 hardware: the SI bit is set when a serial interrupt is requested, and the STO bit is cleared when a STOP condition is present on the  $I^2C$  bus. The STO bit is also cleared when ENS1 = "0".

	7	6	5	4	3	2	1	0
S1CON (D8H)	CR2	ENS1	STA	STO	SI	AA	CR1	CR0

ENS1, THE SIO1 ENABLE BIT

ENS1 = "0": When ENS1 is "0", the SDA and SCL outputs are in a high impedance state. SDA and SCL input signals are ignored, SIO1 is in the "not addressed" slave state, and the STO bit in S1CON is forced to "0". No other bits are affected. P1.6 and P1.7 may be used as open drain I/O ports.

ENS1 = "1": When ENS1 is "1", SIO1 is enabled. The P1.6 and P1.7 port latches must be set to logic 1.

ENS1 should not be used to temporarily release SIO1 from the I2C bus since, when ENS1 is reset, the I2C bus status is lost. The AA flag should be used instead (see description of the AA flag in the following text).

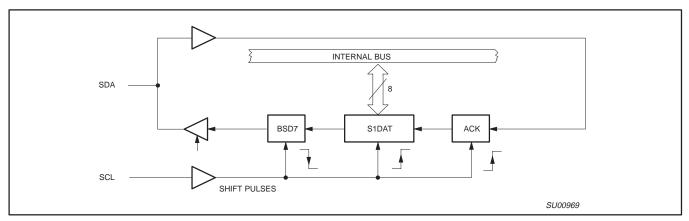


Figure 20. Serial Input/Output Configuration

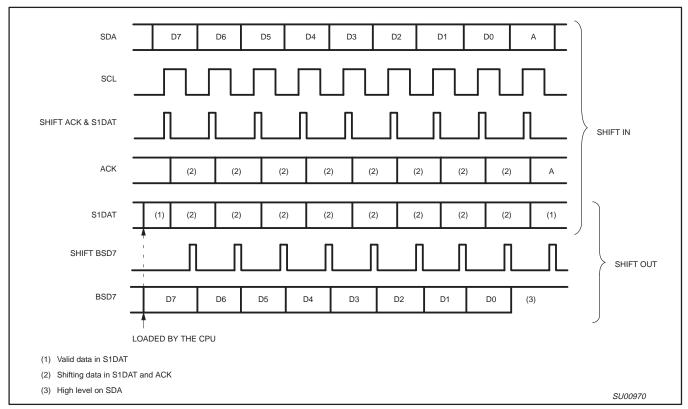


Figure 21. Shift-in and Shift-out Timing

In the following text, it is assumed that ENS1 = "1".

### STA, THE START FLAG

STA = "1": When the STA bit is set to enter a master mode, the SIO1 hardware checks the status of the I2C bus and generates a START condition if the bus is free. If the bus is not free, then SIO1 waits for a STOP condition (which will free the bus) and generates a START condition after a delay of a half clock period of the internal serial clock generator.

If STA is set while SIO1 is already in a master mode and one or more bytes are transmitted or received, SIO1 transmits a repeated START condition. STA may be set at any time. STA may also be set when SIO1 is an addressed slave.

STA = "0": When the STA bit is reset, no START condition or repeated START condition will be generated.

### STO, THE STOP FLAG

STO = "1": When the STO bit is set while SIO1 is in a master mode, a STOP condition is transmitted to the  $I^2C$  bus. When the STOP condition is detected on the bus, the SIO1 hardware clears the STO flag. In a slave mode, the STO flag may be set to recover from an error condition. In this case, no STOP condition is transmitted to the  $I^2C$  bus. However, the SIO1 hardware behaves as if a STOP condition has been received and switches to the defined "not addressed" slave receiver mode. The STO flag is automatically cleared by hardware.

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If the STA and STO bits are both set, the a STOP condition is transmitted to the I<sup>2</sup>C bus if SIO1 is in a master mode (in a slave mode, SIO1 generates an internal STOP condition which is not transmitted). SIO1 then transmits a START condition.

STO = "0": When the STO bit is reset, no STOP condition will be generated.

SI, THE SERIAL INTERRUPT FLAG

SI = "1": When the SI flag is set, then, if the EA and ES1 (interrupt enable register) bits are also set, a serial interrupt is requested. SI is set by hardware when one of 25 of the 26 possible SIO1 states is entered. The only state that does not cause SI to be set is state F8H, which indicates that no relevant state information is available.

While SI is set, the LOW period of the serial clock on the SCL line is stretched, and the serial transfer is suspended. A HIGH level on the SCL line is unaffected by the serial interrupt flag. SI must be reset by software.

SI = "0": When the SI flag is reset, no serial interrupt is requested, and there is no stretching of the serial clock on the SCL line.

AA, THE ASSERT ACKNOWLEDGE FLAG

AA = "1": If the AA flag is set, an acknowledge (LOW level to SDA) will be returned during the acknowledge clock pulse on the SCL line when:

- The "own slave address" has been received
- The general call address has been received while the general call bit (GC) in S1ADR is set
- A data byte has been received while SIO1 is in the master receiver mode
- A data byte has been received while SIO1 is in the addressed slave receiver mode

AA = "0": if the AA flag is reset, a not acknowledge (HIGH level to SDA) will be returned during the acknowledge clock pulse on SCL when:

- A data has been received while SIO1 is in the master receiver mode
- A data byte has been received while SIO1 is in the addressed slave receiver mode

When SIO1 is in the addressed slave transmitter mode, state C8H will be entered after the last serial is transmitted (see Figure 25). When SI is cleared, SIO1 leaves state C8H, enters the not addressed slave receiver mode, and the SDA line remains at a HIGH level. In state C8H, the AA flag can be set again for future address recognition.

When SIO1 is in the not addressed slave mode, its own slave address and the general call address are ignored. Consequently, no acknowledge is returned, and a serial interrupt is not requested. Thus, SIO1 can be temporarily released from the I<sup>2</sup>C bus while the

bus status is monitored. While SIO1 is released from the bus, START and STOP conditions are detected, and serial data is shifted in. Address recognition can be resumed at any time by setting the AA flag. If the AA flag is set when the part's own slave address or the general call address has been partly received, the address will be recognized at the end of the byte transmission.

CR0, CR1, AND CR2, THE CLOCK RATE BITS These three bits determine the serial clock frequency when SIO1 is in a master mode. The various serial rates are shown in Table 7.

For the SIO1 serial port, the Standard data transfer mode is the default mode after reset. To change the data transfer mode to the Fast-mode, the Fast Mode Enable bit (FME bit) of the AUXR Register (AUXR.3 bit) must be set. **After setting the FME bit you cannot clear it** (a one-time set bit), and it can only be **cleared with a reset**.

For the SIO2 serial port, the analog circuits for controlling the slew–rates of the output pull-downs may be disabled with the Slew–Rate Disable bit (AUXR.5 bit). For maximum slew rates, setting this bit disables the slew–rate control circuits of the SCL1 and SDA1 pins. This bit is cleared by reset (reset default), and it can be set/cleared by software. This feature of the SIO2 slew–rate control is very useful for higher bus loads, higher temperatures and lower voltages that cause additional decreases in slew–rates.

	7	6	5	4	3	2	1	0
AUXR (8EH)	_	_	SRD	_	FME	-	EX- TRAM	A0

A 12.5kHz bit rate may be used by devices that interface to the I<sup>2</sup>C bus via standard I/O port lines which are software driven and slow. 100kHz is usually the maximum bit rate and can be derived from a 16 MHz, 12 MHz, or a 6 MHz oscillator. A variable bit rate (0.5kHz to 62.5kHz) may also be used if Timer 1 is not required for any other purpose while SIO1 is in a master mode.

The frequencies shown in Table 7 are unimportant when SIO1 is in a slave mode. In the slave modes, SIO1 will automatically synchronize with any clock frequency up to 100kHz.

The Status Register, S1STA: S1STA is an 8-bit read-only special function register. The three least significant bits are always zero. The five most significant bits contain the status code. There are 26 possible status codes. When S1STA contains F8H, no relevant state information is available and no serial interrupt is requested. All other S1STA values correspond to defined SIO1 states. When each of these states is entered, a serial interrupt is requested (SI = "1"). A valid status code is present in S1STA one machine cycle after SI is set by hardware and is still present one machine cycle after SI has been reset by software.

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Table 7. 400 kbytes I<sup>2</sup>C interface serial clock rates

			BIT FREQUE	NCY (kHz) AT f	OSC in 6X MO	DE			f <sub>OSC</sub> DIVIDE BY
CR2	CR1	CR0	3 MHz	6 MHz	12 MHz	16 MHz	24 MHz	30 MHz	
1	0	0	25	50	100	133	200	250	120
1	0	1	2	4	8	10	15	19	1600
1	1	0	38	75	150	200	300	375	80
1	1	1	50	100	200	267	400	500	60
0	0	0	100	200	400	533	800	1000	30
0	0	1	4	8	15	20	30	38	800
0	1	0	150	300	600	800	1200	1500	20
0	1	1	200	400	800	1067	1600	2000	15
			BIT FREQUE	NCY (kHz) AT f	OSC in 12X MC	DDE			f <sub>OSC</sub> DIVIDE BY
CR2	CR1	CR0	BIT FREQUEI	NCY (kHz) AT f	OSC in 12X MC	DDE 16 MHz	24 MHz	33 MHz	f <sub>OSC</sub> DIVIDE BY
<b>CR2</b>	<b>CR1</b> 0	<b>CR0</b>		1	1	1	<b>24 MHz</b> 100	<b>33 MHz</b> 138	f <sub>OSC</sub> DIVIDE BY
CR2 1	+	1	3 MHz	6 MHz	12 MHz	16 MHz			
CR2 1 1	0	0	3 MHz 13	6 MHz 25	<b>12 MHz</b> 50	<b>16 MHz</b> 67	100	138	240
CR2 1 1 1 1	0	0	3 MHz 13	6 MHz 25 2	<b>12 MHz</b> 50 4	<b>16 MHz</b> 67 5	100	138	240 3200
1 1 1	0 0 1	0 1 0	3 MHz 13 1 19	6 MHz 25 2 38	12 MHz 50 4 75	16 MHz 67 5 100	100 8 150	138 10 206	240 3200 160
1 1 1	0 0 1 1 1	0 1 0 1	3 MHz 13 1 19 25	6 MHz 25 2 38 50	12 MHz 50 4 75 100	16 MHz 67 5 100 133	100 8 150 200	138 10 206 275	240 3200 160 120
1 1 1 1 0	0 0 1 1 1 0	0 1 0 1	3 MHz 13 1 19 25 50	6 MHz 25 2 38 50 100	12 MHz 50 4 75 100 200	16 MHz 67 5 100 133 267	100 8 150 200 400	138 10 206 275 550	240 3200 160 120 60

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Table 8. 100 kbytes I<sup>2</sup>C interface serial clock rates

			BIT FREQUE	NCY (kHz) AT f	OSC in 6X MO	DE			f <sub>OSC</sub> DIVIDE BY
CR2	CR1	CR0	3 MHz	6 MHz	12 MHz	16 MHz	24 MHz	30 MHz	1
0	0	0	23	47	94	125	188	234	128
0	0	1	27	54	107	143	214	268	112
0	1	0	31	63	125	167	250	313	96
0	1	1	38	75	150	200	300	375	80
1	0	0	6	13	25	33	50	63	480
1	0	1	50	100	200	267	400	500	60
1	1	0	100	200	400	533	800	1000	30
1	1	1	0.2 to 31.2	0.5 to 62.5	1.0 to 125	1.3 to 167	2.0 to 250	2.4 to 313	48x(256–(reload value Timer1))
									Mode 2 value range: 0 to 254
			BIT FREQUE	REQUENCY (kHz) AT f <sub>OSC</sub> in 12X MODE					f <sub>OSC</sub> DIVIDE BY
CR2	CR1	CR0	3 MHz	6 MHz	12 MHz	16 MHz	24 MHz	33 MHz	_
0	0	0	12	23	47	63	94	129	256
0	0	1	13	27	54	71	107	147	224
0	1	0	16	31	63	83	125	172	192
0	1	1	19	38	75	100	150	206	160
1	0	0	3	6	13	17	25	34	960
1	0	1	25	50	100	133	200	275	120
1	1	0	50	100	200	267	400	550	60
1	1	1	0.1 to 15.6	0.2 to 31.3	0.5 to 62.5	0.7 to 83.3	1.0 to 125	1.3 to 172	96x(256–(reload value Timer1))
									Mode 2 value range: 0 to 254

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**More Information on SIO1 Operating Modes:** The four operating modes are:

Master Transmitter

- Master Receiver

- Slave Receiver

Slave Transmitter

Data transfers in each mode of operation are shown in Figures 22–25. These figures contain the following abbreviations:

**Explanation** Abbreviation Start condition SLA 7-bit slave address Read bit (HIGH level at SDA) R W Write bit (LOW level at SDA) Acknowledge bit (LOW level at SDA) Α Ā Not acknowledge bit (HIGH level at SDA) Data 8-bit data byte Stop condition

In Figures 22-25, circles are used to indicate when the serial interrupt flag is set. The numbers in the circles show the status code held in the S1STA register. At these points, a service routine must be executed to continue or complete the serial transfer. These service routines are not critical since the serial transfer is suspended

until the serial interrupt flag is cleared by software.

When a serial interrupt routine is entered, the status code in S1STA is used to branch to the appropriate service routine. For each status code, the required software action and details of the following serial transfer are given in Tables 9-13.

Master Transmitter Mode: In the master transmitter mode, a number of data bytes are transmitted to a slave receiver (see Figure 22). Before the master transmitter mode can be entered, S1CON must be initialized as follows:

	7	6	5	4	3	2	1	0
S1CON (D8H)	CR2	ENS1	STA	STO	SI	AA	CR1	CR0
	bit rate	1	0	0	0	Х	— bit r	ate —

CR0, CR1, and CR2 define the serial bit rate. ENS1 must be set to logic 1 to enable SIO1. If the AA bit is reset, SIO1 will not acknowledge its own slave address or the general call address in the event of another device becoming master of the bus. In other words, if AA is reset, SIO0 cannot enter a slave mode. STA, STO, and SI must be reset.

The master transmitter mode may now be entered by setting the STA bit using the SETB instruction. The SIO1 logic will now test the I<sup>2</sup>C bus and generate a start condition as soon as the bus becomes free. When a START condition is transmitted, the serial interrupt flag (SI) is set, and the status code in the status register (S1STA) will be 08H. This status code must be used to vector to an interrupt service routine that loads S1DAT with the slave address and the data direction bit (SLA+W). The SI bit in S1CON must then be reset before the serial transfer can continue.

When the slave address and the direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set again, and a number of status codes in S1STA are possible. There are 18H, 20H, or 38H for the master mode and also 68H, 78H, or B0H if the slave mode was enabled (AA = logic 1). The appropriate action to be taken for each of these status codes is detailed in Table 9. After a repeated start condition (state 10H). SIO1

may switch to the master receiver mode by loading S1DAT with SLA+R).

Master Receiver Mode: In the master receiver mode, a number of data bytes are received from a slave transmitter (see Figure 23). The transfer is initialized as in the master transmitter mode. When the start condition has been transmitted, the interrupt service routine must load S1DAT with the 7-bit slave address and the data direction bit (SLA+R). The SI bit in S1CON must then be cleared before the serial transfer can continue.

When the slave address and the data direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set again, and a number of status codes in S1STA are possible. These are 40H, 48H, or 38H for the master mode and also 68H, 78H, or B0H if the slave mode was enabled (AA = logic 1). The appropriate action to be taken for each of these status codes is detailed in Table 10. ENS1, CR1, and CR0 are not affected by the serial transfer and are not referred to in Table 10. After a repeated start condition (state 10H), SIO1 may switch to the master transmitter mode by loading S1DAT with SLA+W.

Slave Receiver Mode: In the slave receiver mode, a number of data bytes are received from a master transmitter (see Figure 24). To initiate the slave receiver mode, S1ADR and S1CON must be loaded as follows:

	7	6	5	4	3	2	1	0
S1ADR (DBH)	Х	Х	Х	Х	Х	Х	Х	GC
			ow	n slave ad	dress			

The upper 7 bits are the address to which SIO1 will respond when addressed by a master. If the LSB (GC) is set, SIO1 will respond to the general call address (00H); otherwise it ignores the general call address.

	7	6	5	4	3	2	1	0	
S1CON (D8H)	CR2	ENS1	STA	STO	SI	AA	CR1	CR0	
	X	1	0	0	0	1	X	X	

CR0, CR1, and CR2 do not affect SIO1 in the slave mode. ENS1 must be set to logic 1 to enable SIO1. The AA bit must be set to enable SIO1 to acknowledge its own slave address or the general call address. STA, STO, and SI must be reset.

When S1ADR and S1CON have been initialized, SIO1 waits until it is addressed by its own slave address followed by the data direction bit which must be "0" (W) for SIO1 to operate in the slave receiver mode. After its own slave address and the W bit have been received, the serial interrupt flag (I) is set and a valid status code can be read from S1STA. This status code is used to vector to an interrupt service routine, and the appropriate action to be taken for each of these status codes is detailed in Table 11. The slave receiver mode may also be entered if arbitration is lost while SIO1 is in the master mode (see status 68H and 78H).

If the AA bit is reset during a transfer, SIO1 will return a not acknowledge (logic 1) to SDA after the next received data byte. While AA is reset, SIO1 does not respond to its own slave address or a general call address. However, the I²C bus is still monitored and address recognition may be resumed at any time by setting AA. This means that the AA bit may be used to temporarily isolate SIO1 from the I²C bus.

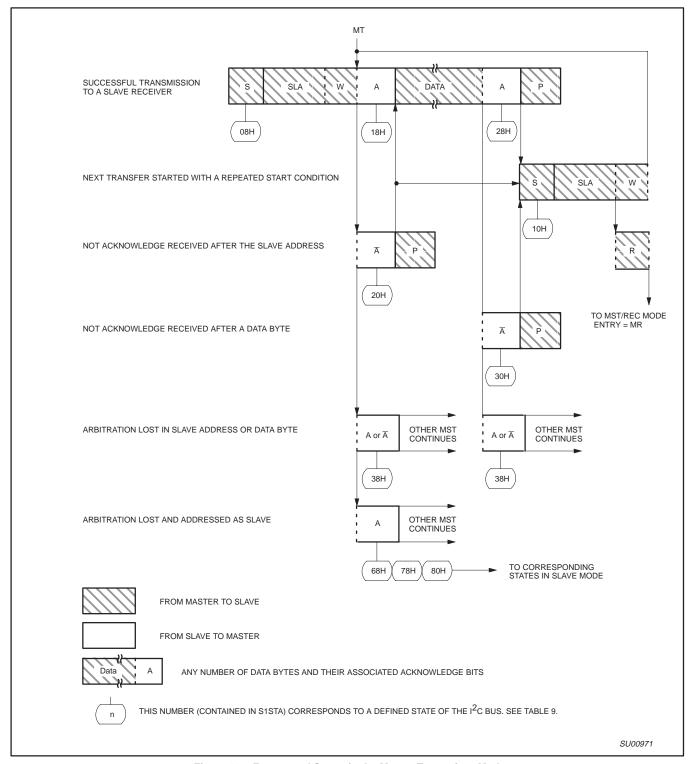


Figure 22. Format and States in the Master Transmitter Mode

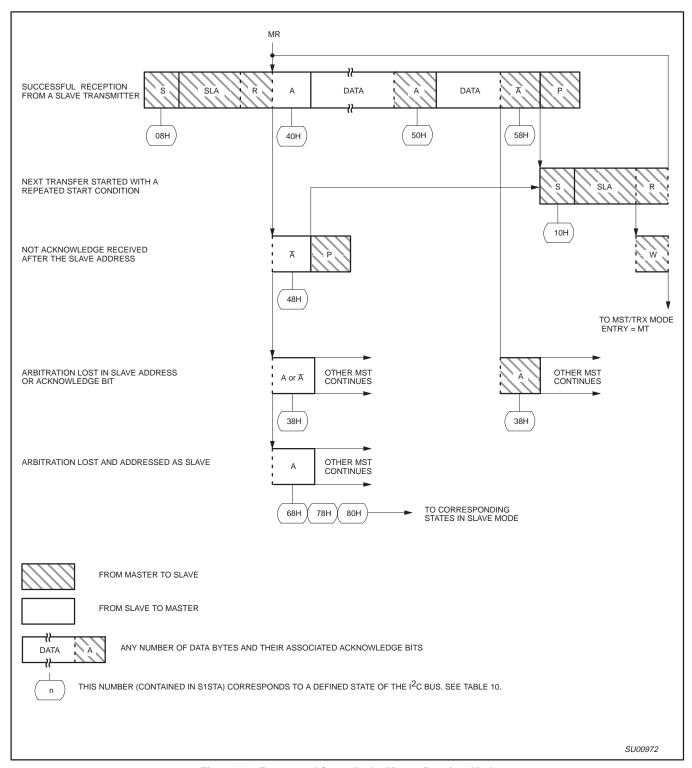


Figure 23. Format and States in the Master Receiver Mode

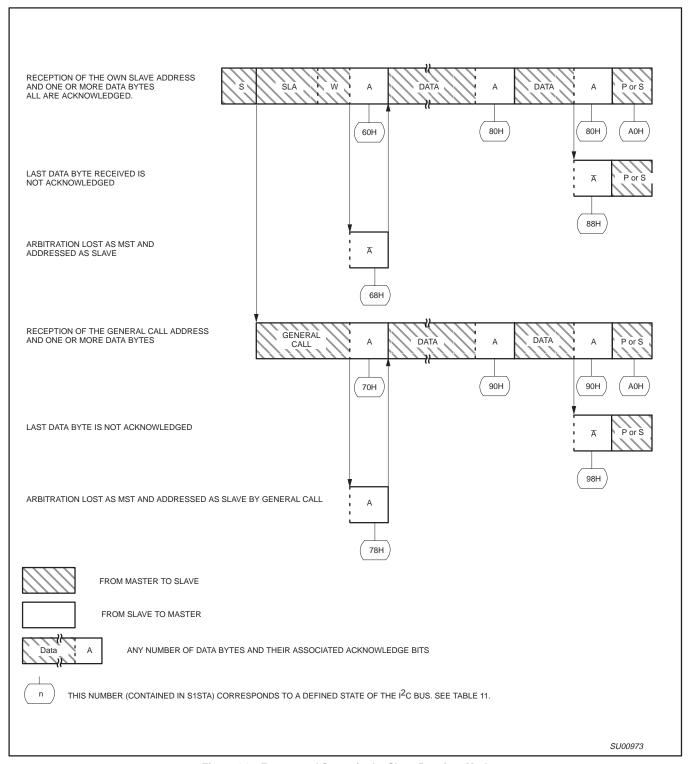


Figure 24. Format and States in the Slave Receiver Mode

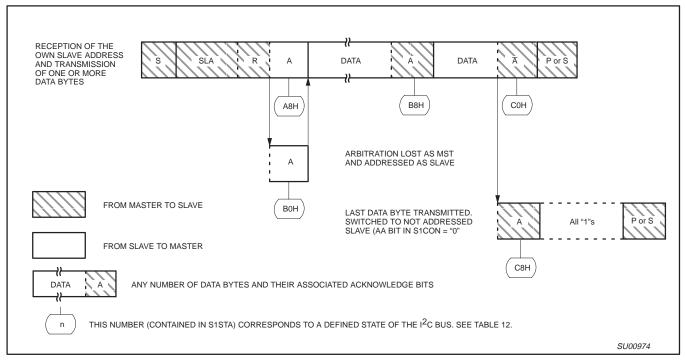


Figure 25. Format and States of the Slave Transmitter Mode

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Table 9. Master Transmitter Mode

STATUS	STATUS OF THE	APPLICATION SOFTWARE RESPONSE					
CODE	I <sup>2</sup> C BUS AND	TO/FROM S1DAT		TO S1	CON		NEXT ACTION TAKEN BY SIO1 HARDWARE
(S1STA)	SIO1 HARDWARE	TO/TROW STDAT	STA	STO	SI	AA	
08H	A START condition has been transmitted	Load SLA+W	Х	0	0	Х	SLA+W will be transmitted; ACK bit will be received
10H	A repeated START	Load SLA+W or	Х	0	0	Х	As above
	condition has been transmitted	Load SLA+R	Х	0	0	Х	SLA+W will be transmitted; SIO1 will be switched to MST/REC mode
18H	SLA+W has been transmitted; ACK has	Load data byte or	0	0	0	X	Data byte will be transmitted; ACK bit will be received
	been received	no S1DAT action or	1	0	0	X	Repeated START will be transmitted;
		no S1DAT action or	0	1	0	X	STOP condition will be transmitted; STO flag will be reset
		no S1DAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted; STO flag will be reset
20H	SLA+W has been transmitted; NOT ACK	Load data byte or	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
	has been received	no S1DAT action or	1	0	0	Х	Repeated START will be transmitted;
		no S1DAT action or	0	1	0	X	STOP condition will be transmitted; STO flag will be reset
		no S1DAT action	1	1	0	X	STOP condition followed by a START condition will be transmitted; STO flag will be reset
28H	Data byte in S1DAT has been transmitted; ACK	Load data byte or	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
	has been received	no S1DAT action or	1	0	0	Х	Repeated START will be transmitted;
		no S1DAT action or	0	1	0	X	STOP condition will be transmitted; STO flag will be reset
		no S1DAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted; STO flag will be reset
30H	Data byte in S1DAT has been transmitted; NOT	Load data byte or	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
	ACK has been received	no S1DAT action or	1	0	0	X	Repeated START will be transmitted;
		no S1DAT action or	0	1	0	X	STOP condition will be transmitted; STO flag will be reset
		no S1DAT action	1	1	0	X	STOP condition followed by a START condition will be transmitted; STO flag will be reset
38H	Arbitration lost in SLA+R/W or	No S1DAT action or	0	0	0	Х	I <sup>2</sup> C bus will be released; not addressed slave will be entered
	Data bytes	No S1DAT action	1	0	0	X	A START condition will be transmitted when the bus becomes free

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Table 10. Master Receiver Mode

STATUS	STATUS OF THE I <sup>2</sup> C	APPLICATION S	OFTWA	RE RE	SPONS		
CODE	BUS AND	TO/FROM S1DAT		TO S	ICON		NEXT ACTION TAKEN BY SIO1 HARDWARE
(S1STA)	SIO1 HARDWARE	10/FROM STDAT	STA STO SI AA		AA		
08H	A START condition has been transmitted	Load SLA+R	Х	0	0	Х	SLA+R will be transmitted; ACK bit will be received
10H	A repeated START condition has been transmitted	Load SLA+R or Load SLA+W	X	0	0	X	As above SLA+W will be transmitted; SIO1 will be switched to MST/TRX mode
38H	Arbitration lost in NOT ACK bit	No S1DAT action or No S1DAT action	0	0	0	X	I <sup>2</sup> C bus will be released; SIO1 will enter a slave mode A START condition will be transmitted when the bus becomes free
40H	SLA+R has been transmitted; ACK has been received	No S1DAT action or no S1DAT action	0	0	0	0	Data byte will be received; NOT ACK bit will be returned Data byte will be received; ACK bit will be returned
48H	SLA+R has been	No S1DAT action or	1	0	0	Х	Repeated START condition will be transmitted
	transmitted; NOT ACK has been received	no S1DAT action or no S1DAT action	0	1	0	X	STOP condition will be transmitted; STO flag will be reset STOP condition followed by a START condition will be transmitted; STO flag will be reset
50H	Data byte has been received; ACK has been returned	Read data byte or read data byte	0	0	0	0	Data byte will be received; NOT ACK bit will be returned Data byte will be received; ACK bit will be returned
58H	Data byte has been	Read data byte or	1	0	0	Х	Repeated START condition will be transmitted
	received; NOT ACK has been returned	read data byte or read data byte	0	1	0	X	STOP condition will be transmitted; STO flag will be reset STOP condition followed by a START condition will be transmitted; STO flag will be reset

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Table 11. Slave Receiver Mode

STATUS	STATUS OF THE	APPLICATION S	OFTWA	RE RE	SPONS	SE.	
CODE	I <sup>2</sup> C BUS AND	TO/FROM S1DAT		TO S1	CON		NEXT ACTION TAKEN BY SIO1 HARDWARE
(S1STA)	SIO1 HARDWARE	10,1110 012,11	STA	STO	SI	AA	
60H	Own SLA+W has been received; ACK	No S1DAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
	has been returned	no S1DAT action	Х	0	0	1	Data byte will be received and ACK will be returned
68H	Arbitration lost in SLA+R/W as master; Own SLA+W has been received, ACK	No S1DAT action or no S1DAT action	X	0	0	0	Data byte will be received and NOT ACK will be returned  Data byte will be received and ACK will be returned
	returned						
70H	General call address (00H) has been received; ACK has	No S1DAT action or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
	been returned	no S1DAT action	Х	0	0	1	Data byte will be received and ACK will be returned
78H	Arbitration lost in SLA+R/W as master; General call address has been received.	No S1DAT action or	X	0	0	0	Data byte will be received and NOT ACK will be returned
	ACK has been returned	no S1DAT action	X	0	0	1	Data byte will be received and ACK will be returned
80H	Previously addressed with own SLV address; DATA has	Read data byte or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
	been received; ACK has been returned	read data byte	Х	0	0	1	Data byte will be received and ACK will be returned
88H	Previously addressed with own SLA; DATA	Read data byte or	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
	byte has been received; NOT ACK has been returned	read data byte or	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1
		read data byte or	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
		read data byte	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1. A START condition will be transmitted when the bus becomes free.
90H	Previously addressed with General Call;	Read data byte or	Х	0	0	0	Data byte will be received and NOT ACK will be returned
	DATA byte has been received; ACK has been returned	read data byte	Х	0	0	1	Data byte will be received and ACK will be returned
98H	Previously addressed with General Call;	Read data byte or	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
	DATA byte has been received; NOT ACK has been returned	read data byte or	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1
		read data byte or	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
		read data byte	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1. A START condition will be transmitted when the bus becomes free.

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Table 11. Slave Receiver Mode (Continued)

SILITATS	TATUS STATUS OF THE APPLICATION SOFTWARE RES				SPONS	SE	
CODE	I <sup>2</sup> C BUS AND	TO/FROM S1DAT		TO S1CON			NEXT ACTION TAKEN BY SIO1 HARDWARE
(S1STA)			AA				
A0H	A STOP condition or repeated START	No STDAT action or	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
	condition has been received while still addressed as	No STDAT action or	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1
	SLV/REC or SLV/TRX	No STDAT action or	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
		No STDAT action	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1. A START condition will be transmitted when the bus becomes free.

Table 12. Slave Transmitter Mode

STATUS	STATUS OF THE	APPLICATION S	OFTWA	RE RE	SPONS	SE		
CODE	I <sup>2</sup> C BUS AND	TO/FROM S1DAT		TO S1	CON		NEXT ACTION TAKEN BY SIO1 HARDWARE	
(S1STA)	SIO1 HARDWARE	10/FROM STDAT	STA	STO	SI	AA		
A8H	Own SLA+R has been received; ACK	Load data byte or	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received	
	has been returned	load data byte	Х	0	0	1	Data byte will be transmitted; ACK will be received	
ВОН	Arbitration lost in SLA+R/W as master; Own SLA+R has	Load data byte or	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received	
	been received, ACK has been returned	load data byte	Х	0	0	1	Data byte will be transmitted; ACK bit will be received	
В8Н	Data byte in S1DAT has been transmitted;	Load data byte or	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received	
	ACK has been received	load data byte	Х	0	0	1	Data byte will be transmitted; ACK bit will be received	
C0H	Data byte in S1DAT has been transmitted;	No S1DAT action or	0	0	0	01	Switched to not addressed SLV mode; no recognition of own SLA or General call address	
	NOT ACK has been received	no S1DAT action or	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1	
		no S1DAT action or	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free	
		no S1DAT action	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1. A START condition will be transmitted when the bus becomes free.	
C8H	Last data byte in S1DAT has been	No S1DAT action or	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address	
	transmitted (AA = 0); ACK has been received	no S1DAT action or	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1	
		no S1DAT action or	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free	
		no S1DAT action	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if S1ADR.0 = logic 1. A START condition will be transmitted when the bus becomes free.	

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Table 13. Miscellaneous States

STATUS	STATUS OF THE	APPLICATION SO	OFTWA	RE RES	SPONS	SE	
CODE	I <sup>2</sup> C BUS AND	TO/FROM S1DAT	TO S1CON				NEXT ACTION TAKEN BY SIO1 HARDWARE
(S1STA)	SIO1 HARDWARE	TO/FROW STDAT	STA	STO	SI	AA	
F8H	No relevant state information available; SI = 0	No S1DAT action	No	S1CO	N actio	n	Wait or proceed current transfer
00H	Bus error during MST or selected slave modes, due to an illegal START or STOP condition. State 00H can also occur when interference causes SIO1 to enter an undefined state.	No S1DAT action	0	1	0	Х	Only the internal hardware is affected in the MST or addressed SLV modes. In all cases, the bus is released and SIO1 is switched to the not addressed SLV mode. STO is reset.

Slave Transmitter Mode: In the slave transmitter mode, a number of data bytes are transmitted to a master receiver (see Figure 25). Data transfer is initialized as in the slave receiver mode. When S1ADR and S1CON have been initialized, SIO1 waits until it is addressed by its own slave address followed by the data direction bit which must be "1" (R) for SIO1 to operate in the slave transmitter mode. After its own slave address and the R bit have been received, the serial interrupt flag (SI) is set and a valid status code can be read from S1STA. This status code is used to vector to an interrupt service routine, and the appropriate action to be taken for each of these status codes is detailed in Table 12. The slave transmitter mode may also be entered if arbitration is lost while SIO1 is in the master mode (see state B0H).

If the AA bit is reset during a transfer, SIO1 will transmit the last byte of the transfer and enter state C0H or C8H. SIO1 is switched to the not addressed slave mode and will ignore the master receiver if it continues the transfer. Thus the master receiver receives all 1s as serial data. While AA is reset, SIO1 does not respond to its own slave address or a general call address. However, the I<sup>2</sup>C bus is still monitored, and address recognition may be resumed at any time by setting AA. This means that the AA bit may be used to temporarily isolate SIO1 from the I<sup>2</sup>C bus.

**Miscellaneous States:** There are two S1STA codes that do not correspond to a defined SIO1 hardware state (see Table 13). These are discussed below.

### S1STA = F8H:

This status code indicates that no relevant information is available because the serial interrupt flag, SI, is not yet set. This occurs between other states and when SIO1 is not involved in a serial transfer.

### S1STA = 00H:

This status code indicates that a bus error has occurred during an SIO1 serial transfer. A bus error is caused when a START or STOP condition occurs at an illegal position in the format frame. Examples of such illegal positions are during the serial transfer of an address byte, a data byte, or an acknowledge bit. A bus error may also be caused when external interference disturbs the internal SIO1 signals. When a bus error occurs, SI is set. To recover from a bus error, the STO flag must be set and SI must be cleared. This causes SIO1 to enter the "not addressed" slave mode (a defined state) and to clear the STO flag (no other bits in S1CON are affected). The

SDA and SCL lines are released (a STOP condition is not transmitted).

**Some Special Cases:** The SIO1 hardware has facilities to handle the following special cases that may occur during a serial transfer:

Simultaneous Repeated START Conditions from Two Masters

A repeated START condition may be generated in the master transmitter or master receiver modes. A special case occurs if another master simultaneously generates a repeated START condition (see Figure 26). Until this occurs, arbitration is not lost by either master since they were both transmitting the same data.

If the SIO1 hardware detects a repeated START condition on the I<sup>2</sup>C bus before generating a repeated START condition itself, it will release the bus, and no interrupt request is generated. If another master frees the bus by generating a STOP condition, SIO1 will transmit a normal START condition (state 08H), and a retry of the total serial data transfer can commence.

DATA TRANSFER AFTER LOSS OF ARBITRATION

Arbitration may be lost in the master transmitter and master receiver modes (see Figure 18). Loss of arbitration is indicated by the following states in S1STA; 38H, 68H, 78H, and B0H (see Figures 22 and 23).

If the STA flag in S1CON is set by the routines which service these states, then, if the bus is free again, a START condition (state 08H) is transmitted without intervention by the CPU, and a retry of the total serial transfer can commence.

FORCED ACCESS TO THE I<sup>2</sup>C Bus

In some applications, it may be possible for an uncontrolled source to cause a bus hang-up. In such situations, the problem may be caused by interference, temporary interruption of the bus or a temporary short-circuit between SDA and SCL.

If an uncontrolled source generates a superfluous START or masks a STOP condition, then the  $\rm I^2C$  bus stays busy indefinitely. If the STA flag is set and bus access is not obtained within a reasonable amount of time, then a forced access to the  $\rm I^2C$  bus is possible. This is achieved by setting the STO flag while the STA flag is still set. No STOP condition is transmitted. The SIO1 hardware behaves as if a STOP condition was received and is able to transmit a START condition. The STO flag is cleared by hardware (see Figure 27).

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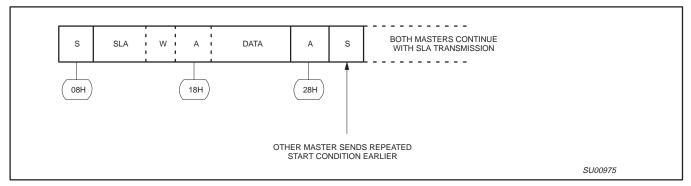


Figure 26. Simultaneous Repeated START Conditions from 2 Masters

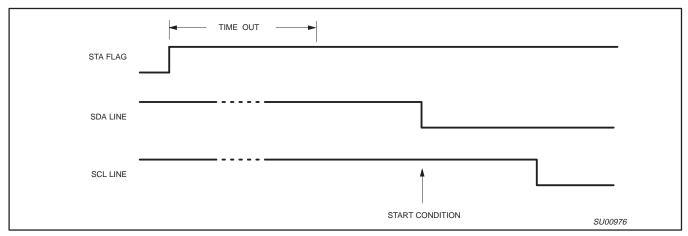


Figure 27. Forced Access to a Busy I<sup>2</sup>C Bus

I<sup>2</sup>C Bus Obstructed by a Low Level on SCL or SDA An I<sup>2</sup>C bus hang-up occurs if SDA or SCL is pulled LOW by an uncontrolled source. If the SCL line is obstructed (pulled LOW) by a device on the bus, no further serial transfer is possible, and the SIO1 hardware cannot resolve this type of problem. When this occurs, the problem must be resolved by the device that is pulling the SCL bus line LOW.

If the SDA line is obstructed by another device on the bus (e.g., a slave device out of bit synchronization), the problem can be solved by transmitting additional clock pulses on the SCL line (see Figure 28). The SIO1 hardware transmits additional clock pulses when the STA flag is set, but no START condition can be generated because the SDA line is pulled LOW while the I<sup>2</sup>C bus is considered free. The SIO1 hardware attempts to generate a START condition after every two additional clock pulses on the SCL line. When the SDA line is eventually released, a normal START condition is transmitted, state 08H is entered, and the serial transfer continues.

If a forced bus access occurs or a repeated START condition is transmitted while SDA is obstructed (pulled LOW), the SIO1

hardware performs the same action as described above. In each case, state 08H is entered after a successful START condition is transmitted and normal serial transfer continues. Note that the CPU is not involved in solving these bus hang-up problems.

### Bus Error

A bus error occurs when a START or STOP condition is present at an illegal position in the format frame. Examples of illegal positions are during the serial transfer of an address byte, a data or an acknowledge bit.

The SIO1 hardware only reacts to a bus error when it is involved in a serial transfer either as a master or an addressed slave. When a bus error is detected, SIO1 immediately switches to the not addressed slave mode, releases the SDA and SCL lines, sets the interrupt flag, and loads the status register with 00H. This status code may be used to vector to a service routine which either attempts the aborted serial transfer again or simply recovers from the error condition as shown in Table 13.

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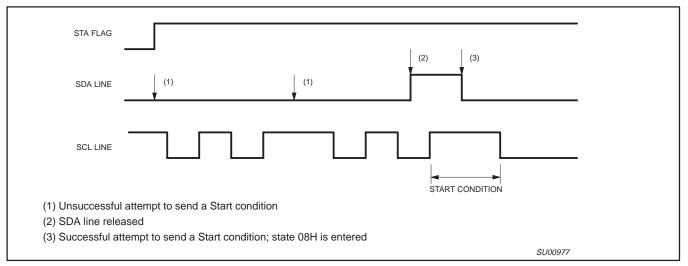


Figure 28. Recovering from a Bus Obstruction Caused by a Low Level on SDA

# **Software Examples of SIO1 Service Routines:** This section consists of a software example for:

- Initialization of SIO1 after a RESET
- Entering the SIO1 interrupt routine
- The 26 state service routines for the
  - Master transmitter mode
  - Master receiver mode
  - Slave receiver mode
  - Slave transmitter mode

### INITIALIZATION

In the initialization routine, SIO1 is enabled for both master and slave modes. For each mode, a number of bytes of internal data RAM are allocated to the SIO to act as either a transmission or reception buffer. In this example, 8 bytes of internal data RAM are reserved for different purposes. The data memory map is shown in Figure 29. The initialization routine performs the following functions:

- S1ADR is loaded with the part's own slave address and the general call bit (GC)
- P1.6 and P1.7 bit latches are loaded with logic 1s
- RAM location HADD is loaded with the high-order address byte of the service routines
- The SIO1 interrupt enable and interrupt priority bits are set
- The slave mode is enabled by simultaneously setting the ENS1 and AA bits in S1CON and the serial clock frequency (for master modes) is defined by loading CR0 and CR1 in S1CON. The master routines must be started in the main program.

The SIO1 hardware now begins checking the  $I^2C$  bus for its own slave address and general call. If the general call or the own slave address is detected, an interrupt is requested and S1STA is loaded with the appropriate state information. The following text describes a fast method of branching to the appropriate service routine.

#### SIO<sub>1</sub> INTERRUPT ROUTINE

When the SIO1 interrupt is entered, the PSW is first pushed on the stack. Then S1STA and HADD (loaded with the high-order address byte of the 26 service routines by the initialization routine) are pushed on to the stack. S1STA contains a status code which is the lower byte of one of the 26 service routines. The next instruction is RET, which is the return from subroutine instruction. When this instruction is executed, the HIGH and LOW order address bytes are popped from stack and loaded into the program counter.

The next instruction to be executed is the first instruction of the state service routine. Seven bytes of program code (which execute in eight machine cycles) are required to branch to one of the 26 state service routines.

SI	PUSH PSW	Save PSW
	PUSH S1STA	Push status code
		(LOW order address byte)
	PUSH HADD	Push HIGH order address byte
	RET	Jump to state service routine

The state service routines are located in a 256-byte page of program memory. The location of this page is defined in the initialization routine. The page can be located anywhere in program memory by loading data RAM register HADD with the page number. Page 01 is chosen in this example, and the service routines are located between addresses 0100H and 01FFH.

### THE STATE SERVICE ROUTINES

The state service routines are located 8 bytes from each other. Eight bytes of code are sufficient for most of the service routines. A few of the routines require more than 8 bytes and have to jump to other locations to obtain more bytes of code. Each state routine is part of the SIO1 interrupt routine and handles one of the 26 states. It ends with a RETI instruction which causes a return to the main program.

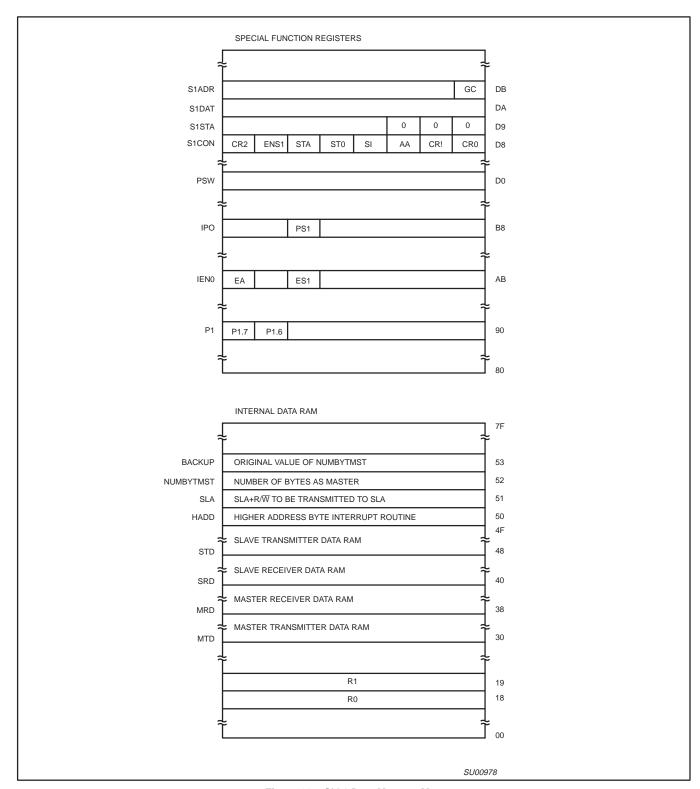


Figure 29. SIO1 Data Memory Map

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MASTER TRANSMITTER AND MASTER RECEIVER MODES
The master mode is entered in the main program. To enter the master transmitter mode, the main program must first load the internal data RAM with the slave address, data bytes, and the number of data bytes to be transmitted. To enter the master receiver mode, the main program must first load the internal data RAM with the slave address and the number of data bytes to be received. The R/W bit determines whether SIO1 operates in the master transmitter or master receiver mode.

Master mode operation commences when the STA bit in S1CION is set by the SETB instruction and data transfer is controlled by the master state service routines in accordance with Table 9, Table 10, Figure 22, and Figure 23. In the example below, 4 bytes are transferred. There is no repeated START condition. In the event of lost arbitration, the transfer is restarted when the bus becomes free. If a bus error occurs, the I<sup>2</sup>C bus is released and SIO1 enters the not selected slave receiver mode. If a slave device returns a not acknowledge, a STOP condition is generated.

A repeated START condition can be included in the serial transfer if the STA flag is set instead of the STO flag in the state service routines vectored to by status codes 28H and 58H. Additional software must be written to determine which data is transferred after a repeated START condition.

SLAVE TRANSMITTER AND SLAVE RECEIVER MODES
After initialization, SIO1 continually tests the I<sup>2</sup>C bus and branches
to one of the slave state service routines if it detects its own slave
address or the general call address (see Table 11, Table 12, Figure
24, and Figure 25). If arbitration was lost while in the master mode,
the master mode is restarted after the current transfer. If a bus error

occurs, the  $I^2C$  bus is released and SIO1 enters the not selected slave receiver mode.

In the slave receiver mode, a maximum of 8 received data bytes can be stored in the internal data RAM. A maximum of 8 bytes ensures that other RAM locations are not overwritten if a master sends more bytes. If more than 8 bytes are transmitted, a not acknowledge is returned, and SIO1 enters the not addressed slave receiver mode. A maximum of one received data byte can be stored in the internal data RAM after a general call address is detected. If more than one byte is transmitted, a not acknowledge is returned and SIO1 enters the not addressed slave receiver mode.

In the slave transmitter mode, data to be transmitted is obtained from the same locations in the internal data RAM that were previously loaded by the main program. After a not acknowledge has been returned by a master receiver device, SIO1 enters the not addressed slave mode.

ADAPTING THE SOFTWARE FOR DIFFERENT APPLICATIONS
The following software example shows the typical structure of the interrupt routine including the 26 state service routines and may be used as a base for user applications. If one or more of the four modes are not used, the associated state service routines may be removed but, care should be taken that a deleted routine can never be invoked.

This example does not include any time-out routines. In the slave modes, time-out routines are not very useful since, in these modes, SIO1 behaves essentially as a passive device. In the master modes, an internal timer may be used to cause a time-out if a serial transfer is not complete after a defined period of time. This time period is defined by the system connected to the I<sup>2</sup>C bus.

# P8xC660X2/661X2

	1*******	****************	*****	*******		
	: ! SIO1 EQUATE LIST					
	[*************************************					
	! LOCATIONS OF THE SIO1 SPECIAL FUNCTION REGISTERS					
_	!******	****************		*******		
00D8 00D9	S1CON S1STA	-0xd8 -0xd9				
00DA	S1DAT	–0xda				
00DB	S1ADR	-0xdb				
00A8	IEN0	-0xa8				
00B8	IP0	-02b8				
	•	***************	******	*********		
	! BIT LOC.	ATIONS ************************************	******	********		
00DD	STA	-0xdd	! STA bit i	n S1CON		
00BD	SIO1HP	-0xbd	! IP0, SIO	1 Priority bit		
	!******	**************	******	*******		
		ATE DATA TO WRITE INTO REGISTER S1		*****		
00D5	:	TSTA_STO_NOTSI_AA_CR0	-0xd5	! Generates STOP ! (CR0 = 100kHz)		
00C5	ENS1_NO	TSTA_NOTSTO_NOTSI_AA_CR0	-0xc5	! Releases BUS and		
00C1	ENS1_NO	TSTA_NOTSTO_NOTSI_NOTAA_CR0	-0xc1	! Releases BUS and ! NOT ACK		
00E5	ENS1_ST	A_NOTSTO_NOTSI_AA_CR0	-0xe5	! Releases BUS and ! set STA		
	******	**************	*****	*******		
		L IMMEDIATE DATA	******	******		
0031	OWNSLA	-0x31		A+General Call written into S1ADR		
00A0	ENSIO1	-0xa0	! EA+ES1	, enable SIO1 interrupt written into IEN0		
0001	PAG1	-0x01		AG1 as HADD		
00C0 00C1	SLAW SLAR	-0xc0 -0xc1	_	to be transmitted to be transmitted		
0018	SELRB3	-0x18		egister Bank 3		
	******	***************	******	*******		
		DNS IN DATA RAM	******	*******		
0030	MTD	-0x30		X/DATA base address		
0038	MRD	-0x38		C/DATA base address		
0040 0048	SRD STD	-0x40 -0x48		C/DATA base address K/DATA base address		
0053	BACKUP	-0x53		from NUMBYTMST		
			! To restor	re NUMBYTMST in case oitration Loss.		
0052	NUMBYTN	MST -0x52	! Number	of bytes to transmit		
0051	SLA	-0x51		SSLA+R/W to be		
0050	HADD	-0x50	! High Add	dress byte for STATE 0 E 25.		

# P8xC660X2/661X2

		! INITIALI ! Example	ZATION RO to initialize	UTINE IIC Inte	rface as slave receiver or slave transm	itter and
		!******** .sect	************ strt	*****	**********	******
0000	4100	.base	0x00	ajmp	INIT	! RESET
		.sect .base	initial 0x200			
0200	75DB31	INIT:	0,200	mov	,	! Load own SLA + enable
0203	D296			setb		! general call recognition ! P1.6 High level.
0205	D297				. ,	! P1.7 High level.
0207 020A	755001 43A8A0			mov orl	HADD,#PAG1 IEN0,#ENSIO1	! Enable SIO1 interrupt
020D	C2BD			clr		! SIO1 interrupt LOW priority
020F	75D8C5			mov	S1CON, #ENS1_NOTSTA_NOTSTO	_NOTSI_AA_CR0 ! Initialize SLV funct.
		!******	******	*****	**********	******
		•				
					T FUNCTION 	
0212	755204			mov	NUMBYTMST,#0x4	! Transmit 4 bytes.
0215	7551C0				SLA,#SLAW	! SLA+W, Transmit funct.
0218	D2DD			setb	STA	! set STA in S1CON
		! ! START N	ASTER RE	CEIVE	FUNCTION	
021A 021D	755204 7551C1	•		mov	NUMBYTMST,#0x4 SLA,#SLAR	! Receive 4 bytes. ! SLA+R, Receive funct.
0220	D2DD			seib	STA	! set STA in S1CON
		! SIO1 IN	ΓERRUPT R	ROUTIN	**************************************	
		.sect .base	intvec 0x00	*****		! SIO1 interrupt vector
		! They ser ! The RET	ve as return instruction	addres	ed onto the stack. ss for the RET instruction. e Program Counter to address HADD, nt subroutine.	
002B 002D	C0D0 C0D9			push push	psw S1STA	! save psw
002F	C050				HADD	
0031	22			ret		! JMP to address HADD,S1STA.
		! STATE ! ACTION	: 00, Bus : Enter no	error. ot addre	ssed SLV mode and release bus. STO	reset.
		.sect .base	st0 0x100			
0100	75D8D5			mov	S1CON,#ENS1_NOTSTA_STO_NOT	ΓSI_AA_CR0 ! clr SI ! set STO,AA
0103 0105	D0D0 32			pop reti	psw	. 301 010,774
2003 O	ct 02				56	

# P8xC660X2/661X2

		! MASTER ! MASTER ! State 08 : ! The R/W ! MST/TR> ! ***********************************	state 1 bit decides (mode or w	RVICE ********* 0 are be whether within M ********  TART c  W are ti	ROUTINES  which is received.	**********
0108 010B	8551DA 75D8C5				S1DAT,SLA S1CON,#ENS1_NOTSTA_NOTSTO	
010E	01A0			ajmp	INITBASE1	! clr SI
		!! ! STATE ! ! ACTION ! .sect .base	transmitt : SLA+R/\	ed.	START condition has been ransmitted, ACK bit is received.	
0110 0113	8551DA 75D8C5			mov mov	S1DAT,SLA S1CON,#ENS1_NOTSTA_NOTSTO	!Load SLA+R/W )_NOTSI_AA_CR0 !clr SI
010E	01A0			ajmp	INITBASE1	
00A0 00A3 00A5 00A7 00AA 00AC	75D018 7930 7838 855253 D0D0 32	.sect .base INITBASE	ibase1 0xa0 1:	mov mov	psw,#SELRB3 r1,#MTD r0,#MRD BACKUP,NUMBYTMST psw	! Save initial value
		*************************************				
0118 011B 011D	75D018 87DA 01B5			mov	psw,#SELRB3 S1DAT,@r1 CON	

# P8xC660X2/661X2

		!					
		! STATE ! ACTION !	: Transmit	STOP			
		.sect	mts20 0x120				
0120	75D8D5			mov	S1CON,#ENS1_NOTSTA_STO_NOTSI_AA_CR0 ! set STO, clr SI		
0123 0125	D0D0 32			pop reti	psw		
		! STATE	: 28, DATA : If Transn else tran	A of S1 nitted D smit ne	DAT have been transmitted, ACK received.  PATA is last DATA then transmit a STOP condition, ext DATA.		
		.sect	mts28 0x128				
0128 012B	D55285 75D8D5			djnz mov	NUMBYTMST,NOTLDAT1 ! JMP if NOT last DATA S1CON,#ENS1_NOTSTA_STO_NOTSI_AA_CR0 ! clr SI, set AA		
012E	01B9			ajmp	RETmt : Cli Si, set An		
			mts28sb 0x0b0				
00B0 00B3	75D018 87DA	NOTLDAT1	:	mov	psw,#SELRB3 S1DAT,@r1		
00B5	75D8C5	CON:		mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr SI, set AA		
00B8 00B9	09 D0D0	RETmt	:	inc pop	r1 psw		
00BB	32	· · · · · · · · · · · · · · · · · · ·		reti			
		! STATE ! ACTION	: 30, DATA	A of S1 a STO	DAT have been transmitted, NOT ACK received. P condition.		
		.sect	mts30 0x130				
0130	75D8D5			mov	S1CON,#ENS1_NOTSTA_STO_NOTSI_AA_CR0 ! set STO, clr SI		
0133 0135	D0D0 32			pop reti	psw		
		! ACTION ! !	: Bus is re	leased TART o	ost in SLA+W or DATA.  , not addressed SLV mode is entered. condition is transmitted when the IIC bus is free again.		
		.base	0x138				
0138 013B 013E	75D8E5 855352 01B9			mov mov ajmp	S1CON,#ENS1_STA_NOTSTO_NOTSI_AA_CR0 NUMBYTMST,BACKUP RETmt		

# P8xC660X2/661X2

		!*******	******	***********************
		! MASTER RE	CEIVER STA	ATE SERVICE ROUTINES
		:		***************************************
		! STATE : 4 ! S	0, Previous s SLA+R have b	state was STATE 08 or STATE 10, been transmitted, ACK received. received, ACK returned.
		.sect mts	- – – – – - s40 140	
0140	75D8C5		mov	v S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr STA, STO, SI set AA
0143	D0D0 32		pop reti	o psw
		! STATE : 4 ! ACTION : S	8, SLA+R ha	ave been transmitted, NOT ACK received. on will be generated.
		.sect mts	s48 148	
0148	75D8D5	STOP:	mov	v S1CON,#ENS1_NOTSTA_STO_NOTSI_AA_CR0 ! set STO, clr SI
014B 014D	D0D0 32		pop reti	•
		! STATE : 5 ! ACTION : R ! D	50, DATA hav Read DATA of DATA will be r hen NOT ACI	re been received, ACK returned. of S1DAT. received, if it is last DATA K will be returned else ACK will be returned.
		.sect mrs	s50 150	
0150 0153 0155	75D018 A6DA 01C0		mov	v psw,#SELRB3 v @r0,S1DAT ! Read received DATA
		.sect mrs	s50s c0	
00C0 00C3	D55205 75D8C1	REC1:	djnz mov	
00C6 00C8	8003 75D8C5	NOTLDAT2:	sjm mov	p RETmr
00CB 00CC 00CE	08 D0D0 32	RETmr:	inc pop reti	r0 psw
		!STATE : 5 !ACTION : R	8, DATA hav	re been received, NOT ACK returned. If S1DAT and generate a STOP condition.
		.sect mrs	s58 158	
0158 015B 015D	75D018 A6DA 80E9		mov	v psw,#SELRB3 v @R0,S1DAT pp STOP

# P8xC660X2/661X2

		!  ************** ! SLAVE RE !*************!	********** ECEIVER S ************************************	******* STATE : ******	**************************************
		! STATE ! ACTION	: 60, Own : DATA wi	SLA+V	V have been received, ACK returned. ceived and ACK returned.
		.sect	srs60 0x160		
0160	75D8C5			mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr SI, set AA
0163 0166	75D018 01D0	.sect .base	insrd 0xd0	mov ajmp	psw,#SELRB3 INITSRD
00D0 00D2 00D4 00D6	7840 7908 D0D0 32	INITSRD:		mov mov pop reti	r0,#SRD r1,#8 psw
		! ! ACTION !	: 68, Arbiti Own SLA : DATA wi STA is se	ration lo A+W ha II be rec et to res	ost in SLA and R/W as MST ave been received, ACK returned ceived and ACK returned.  Start MST mode after the bus is free again.
			srs68 0x168		
0168 016B 016E	75D8E5 75D018 01D0			, ,	S1CON,#ENS1_STA_NOTSTO_NOTSI_AA_CR0 psw,#SELRB3 INITSRD
		! STATE ! ACTION	: 70, Gene : DATA wi	eral call	I has been received, ACK returned. ceived and ACK returned.
		.sect	srs70 0x170		
0170	75D8C5	.5400	OXTTO	mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr SI, set AA
0173 0176	75D018 01D0			mov ajmp	
		!	General : DATA wi	call has	ost in SLA+R/W as MST. s been received, ACK returned. ceived and ACK returned. start MST mode after the bus is free again.
		.sect .base	srs78 0x178		
0178 017B 017E	75D8E5 75D018 01D0	.vasc	0.110	mov mov ajmp	S1CON,#ENS1_STA_NOTSTO_NOTSI_AA_CR0 psw,#SELRB3 ! Initialize SRD counter INITSRD

# P8xC660X2/661X2

		! ACTION : ! !	: Read DA IF receiv THEN su ELSE ne	ATA. ed DAT uperfluct ext DAT	addressed with own SLA. DATA receing was the last ous DATA will be received and NOT. A will be received and ACK returned.	ACK returned
			srs80 0x180			
0180 0183 0185	75D018 A6DA 01D8			mov mov ajmp	psw,#SELRB3 @r0,S1DAT REC2	! Read received DATA
			srs80s 0xd8			
00D8 00DA	D906 75D8C1	REC2: LDAT:		djnz mov	r1,NOTLDAT3 S1CON,#ENS1_NOTSTA_NOTST	O_NOTSI_NOTAA_CR0 ! clr SI,AA
00DD 00DF	D0D0 32	NOTI DATO		pop reti	psw	·
00E0 00E3	75D8C5 08	NOTLDAT3:		mov	s1CON,#ENS1_NOTSTA_NOTST0	O_NOTSI_AA_CR0 ! clr SI, set AA
00E4 00E6	D0D0 32	RETsr:		pop reti	psw	
		! ACTION :	: 88, Previ : No save Recognit	iously a of DAT tion of o	addressed with own SLA. DATA recei A, Enter NOT addressed SLV mode. own SLA. General call recognized, if	ved NOT ACK returned. S1ADR. 0-1.
		.sect s	srs88 0x188			
0188	75D8C5			mov	S1CON,#ENS1_NOTSTA_NOTST	O_NOTSI_AA_CR0 ! clr SI, set AA
018B	01E4	l		ajmp	RETsr	
		! STATE : ! ! ACTION :	DATA ha Read DA After Ger the second	s been TA. neral cand nd DAT	addressed with general call. I received, ACK has been returned.  all only one byte will be received with  A will be received with NOT ACK.  ceived and NOT ACK returned.	
			srs90 0x190			
0190 0193 0195	75D018 A6DA 01DA	<b>.</b>		mov mov ajmp	psw,#SELRB3 @r0,S1DAT LDAT	! Read received DATA
		!	DATA ha : No save Recognit	s been of DAT tion of o	addressed with general call. received, NOT ACK has been return A, Enter NOT addressed SLV mode. own SLA. General call recognized, if	ned. S1ADR. 0-1.
			srs98 0x198			<b></b>
0198	75D8C5			mov	S1CON,#ENS1_NOTSTA_NOTST	O_NOTSI_AA_CR0 ! clr SI, set AA
019B 019D	D0D0 32			pop reti	psw	•

# P8xC660X2/661X2

		! STATE !	: A0, A ST while still : No save	OP cor l addres of DAT	ndition or repeated START has been ressed as SLV/REC or SLV/TRX. A, Enter NOT addressed SLV mode. bwn SLA. General call recognized, if S	eceived, 1ADR. 0–1.
		.sect .base	srsA0 0x1a0			
01A0	75D8C5			mov	S1CON,#ENS1_NOTSTA_NOTSTO	_NOTSI_AA_CR0 ! clr SI, set AA
01A3 01A5	D0D0 32			pop reti	psw	: Ull Ul, GOL AA
		!************! ! SLAVE TI !*********	RANSMITTI	ER STA		*******
		! ! STATE ! ACTION	: A8, Own : DATA wil	SLA+F	R received, ACK returned. nsmitted, A bit received.	
		.sect .base	stsa8 0x1a8			
01A8 01AB	8548DA 75D8C5			mov mov	S1CON,#ENS1_NOTSTA_NOTSTO	! load DATA in S1DAT _NOTSI_AA_CR0 ! clr SI, set AA
01AE	01E8			ajmp	INITBASE2	
00E8 00EB 00ED 00EE 00F0	75D018 7948 09 D0D0 32	.sect .base INITBASE2	ibase2 0xe8 2:	mov mov inc pop reti	psw,#SELRB3 r1, #STD r1 psw	
			: B0, Arbit	ration lo	ost in SLA and R/W as MST. Own SL/ nsmitted, A bit received. start MST mode after the bus is free ag	A+R received, ACK returned.
		.sect .base	stsb0 0x1b0			
01B0 01B3 01B6	8548DA 75D8E5 01E8			mov mov ajmp	S1DAT,STD S1CON,#ENS1_STA_NOTSTO_NO INITBASE2	! load DATA in S1DAT TSI_AA_CR0

# P8xC660X2/661X2

		! ACTION : D	DATA will be to	been transmitted, ACK received. ansmitted, ACK bit is received.
01B8 01BB 01BD	75D018 87DA 01F8	.sect sts	sb8 :1b8 mov mov	psw,#SELRB3 S1DAT,@r1 > SCON
		.sect scr .base 0xf		
00F8	75D8C5	SCON:	mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr Sl. set AA
00FB 00FC 00FE	09 D0D0 32		inc pop reti	r1 psw
		! ACTION : E	C0, DATA has Enter not addr	been transmitted, NOT ACK received. essed SLV mode.
		.sect sts	 sc0 :1c0	
01C0	75D8C5	.base ox	mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr SI, set AA
01C3 01C5	D0D0 32		pop reti	psw
		! STATE : C ! ACTION : E	C8, Last DATA Enter not addr	has been transmitted (AA=0), ACK received.
		.sect sts		
01C8	75D8C5	.base 0x	mov	S1CON,#ENS1_NOTSTA_NOTSTO_NOTSI_AA_CR0 ! clr SI, set AA
01CB 01CD	D0D0 32		pop reti	psw
		:	******	***************************************
		:	******	T ROUTINE
		!		

Figure 30. Internal and External Data Memory Address Space with EXTRAM = 0

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

### P8xC660X2/661X2

### **Interrupt Priority Structure**

The P8xC660X2/661X2 has an 8/9 source four-level interrupt structure (see Table 15).

There are 4 SFRs associated with the four-level interrupt. They are the IE, IEN1, IP, and IPH. (See Figures 31, 32, and 33.) The IPH (Interrupt Priority High) register makes the four-level interrupt structure possible. The IPH is located at SFR address B7H.

The function of the IPH SFR, when combined with the IP SFR, determines the priority of each interrupt. The priority of each interrupt is determined as shown in the following table:

Table 14.

PRIORI	TY BITS	INTERRUPT PRIORITY LEVEL
IPH.x	IP.x	
0	0	Level 0 (lowest priority)
0	1	Level 1
1	0	Level 2
1	1	Level 3 (highest priority)

Table 15. Interrupt Table P8xC661X2

SOURCE	POLLING PRIORITY	REQUEST BITS	HARDWARE CLEAR?	VECTOR ADDRESS
X0	1	IE0	N (L) <sup>1</sup> Y (T) <sup>2</sup>	03H
SIO1 (I2C)	2	-	N	2BH
SIO2 (I2C)	3	-	N	43H
T0	4	TP0	Y	0BH
X1	5	IE1	N (L) Y (T)	13H
T1	6	TF1	Y	1BH
SP	7	RI, TI	N	23H
T2	8	TF2, EXF2	N	3BH
PCA	9	CF, CCFn n = 0–4	N	33H

### NOTES:

- 1. L = Level activated
- 2. T = Transition activated

Table 16. Interrupt Table P8xC662X2

SOURCE	POLLING PRIORITY	REQUEST BITS	HARDWARE CLEAR?	VECTOR ADDRESS
X0	1	IE0	N (L) <sup>1</sup> Y (T) <sup>2</sup>	03H
SIO1 (I2C)	2	-	N	2BH
T0	3	TP0	Υ	0BH
X1	4	IE1	N (L) Y (T)	13H
T1	5	TF1	Υ	1BH
SP	6	RI, TI	N	23H
T2	7	TF2, EXF2	N	3BH
PCA	8	CF, CCFn n = 0-4	N	33H

### NOTES:

- L = Level activated
- 2. T = Transition activated

The priority scheme for servicing the interrupts is the same as that for the 80C51, except there are four interrupt levels rather than two as on the 80C51. An interrupt will be serviced as long as an interrupt of equal or higher priority is not already being serviced. If an interrupt of equal or higher level priority is being serviced, the new

interrupt will wait until it is finished before being serviced. If a lower priority level interrupt is being serviced, it will be stopped and the new interrupt serviced. When the new interrupt is finished, the lower priority level interrupt that was stopped will be completed.

## P8xC660X2/661X2

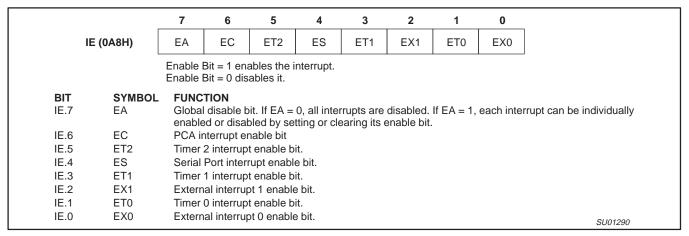


Figure 31. IE Registers

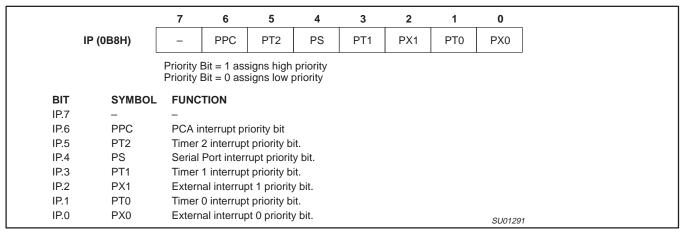


Figure 32. IP Registers

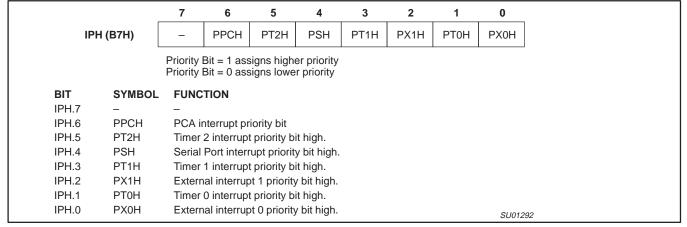


Figure 33. IPH Registers

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

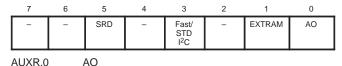
## P8xC660X2/661X2

### **Reduced EMI Mode**

The AO bit (AUXR.0) in the AUXR register when set disables the ALE output unless the CPU needs to perform an off-chip memory access.

### **Reduced EMI Mode**

### AUXR (8EH)



See more detailed description in Figure 48.

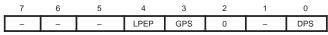
### **Dual DPTR**

The dual DPTR structure (see Figure 34) is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1/bit0 that allows the program code to switch between them.

New Register Name: AUXR1#

SFR Address: A2HReset Value: xxxxxxx0B

### AUXR1 (A2H)



Where:

DPS = AUXR1/bit0 = Switches between DPTR0 and DPTR1.

Select Reg	DPS
DPTR0	0
DPTR1	1

The DPS bit status should be saved by software when switching between DPTR0 and DPTR1.

The GF2 bit is a general purpose user-defined flag. Note that bit 2 is not writable and is always read as a zero. This allows the DPS bit to be quickly toggled simply by executing an INC AUXR1 instruction without affecting the GF2 bit.

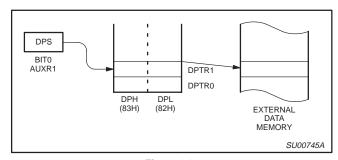


Figure 34.

#### **DPTR Instructions**

The instructions that refer to DPTR refer to the data pointer that is currently selected using the AUXR1/bit 0 register. The six instructions that use the DPTR are as follows:

INC DPTR	Increments the data pointer by 1
MOV DPTR, #data16	Loads the DPTR with a 16-bit constant
MOV A, @ A+DPTR	Move code byte relative to DPTR to ACC
MOVX A, @ DPTR	Move external RAM (16-bit address) to ACC
MOVX @ DPTR , A	Move ACC to external RAM (16-bit address)
JMP @ A + DPTR	Jump indirect relative to DPTR

The data pointer can be accessed on a byte-by-byte basis by specifying the LOW or HIGH byte in an instruction which accesses the SFRs. See *Application Note AN458* for more details.

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## P8xC660X2/661X2

### **Programmable Counter Array (PCA)**

The Programmable Counter Array available on the P8xC66xX2 is a special 16-bit Timer that has five 16-bit capture/compare modules associated with it. Each of the modules can be programmed to operate in one of four modes: rising and/or falling edge capture, software timer, high-speed output, or pulse width modulator. Each module has a pin associated with it in port 1. Module 0 is connected to P1.3 (CEX0), module 1 to P1.4 (CEX1), etc. The basic PCA configuration is shown in Figure 35.

The PCA timer is a common time base for all five modules and can be programmed to run at: 1/6 the oscillator frequency, 1/2 the oscillator frequency, the Timer 0 overflow, or the input on the ECI pin (P1.2). The timer count source is determined from the CPS1 and CPS0 bits in the CMOD SFR as follows (see Figure 38):

### **CPS1 CPS0 PCA Timer Count Source**

- 0 1/6 oscillator frequency (6-clock mode); 1/12 oscillator frequency (12-clock mode)
- 0 1 1/2 oscillator frequency (6-clock mode); 1/4 oscillator frequency (12-clock mode)
- 1 0 Timer 0 overflow
- 1 1 External Input at ECI pin

In the CMOD SFR are three additional bits associated with the PCA. They are CIDL which allows the PCA to stop during idle mode, WDTE which enables or disables the watchdog function on module 4, and ECF which when set causes an interrupt and the PCA overflow flag CF (in the CCON SFR) to be set when the PCA timer overflows. These functions are shown in Figure 36.

The watchdog timer function is implemented in module 4 (see Figure 45).

The CCON SFR contains the run control bit for the PCA and the flags for the PCA timer (CF) and each module (refer to Figure 39). To run the PCA the CR bit (CCON.6) must be set by software. The PCA is shut off by clearing this bit. The CF bit (CCON.7) is set when

the PCA counter overflows and an interrupt will be generated if the ECF bit in the CMOD register is set, The CF bit can only be cleared by software. Bits 0 through 4 of the CCON register are the flags for the modules (bit 0 for module 0, bit 1 for module 1, etc.) and are set by hardware when either a match or a capture occurs. These flags also can only be cleared by software. The PCA interrupt system shown in Figure 37.

Each module in the PCA has a special function register associated with it. These registers are: CCAPM0 for module 0, CCAPM1 for module 1, etc. (see Figure 40). The registers contain the bits that control the mode that each module will operate in. The ECCF bit (CCAPMn.0 where n=0, 1, 2, 3, or 4 depending on the module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated module. PWM (CCAPMn.1) enables the pulse width modulation mode. The TOG bit (CCAPMn.2) when set causes the CEX output associated with the module to toggle when there is a match between the PCA counter and the module's capture/compare register. The match bit MAT (CCAPMn.3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the module's capture/compare register.

The next two bits CAPN (CCAPMn.4) and CAPP (CCAPMn.5) determine the edge that a capture input will be active on. The CAPN bit enables the negative edge, and the CAPP bit enables the positive edge. If both bits are set both edges will be enabled and a capture will occur for either transition. The last bit in the register ECOM (CCAPMn.6) when set enables the comparator function. Figure 41 shows the CCAPMn settings for the various PCA functions.

There are two additional registers associated with each of the PCA modules. They are CCAPnH and CCAPnL and these are the registers that store the 16-bit count when a capture occurs or a compare should occur. When a module is used in the PWM mode these registers are used to control the duty cycle of the output.

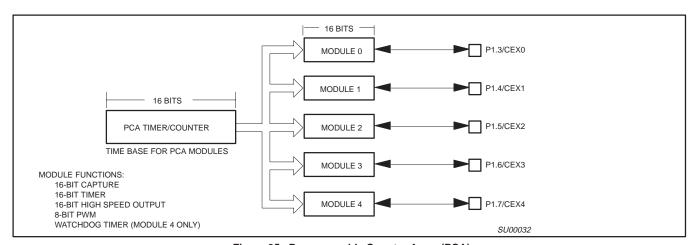


Figure 35. Programmable Counter Array (PCA)

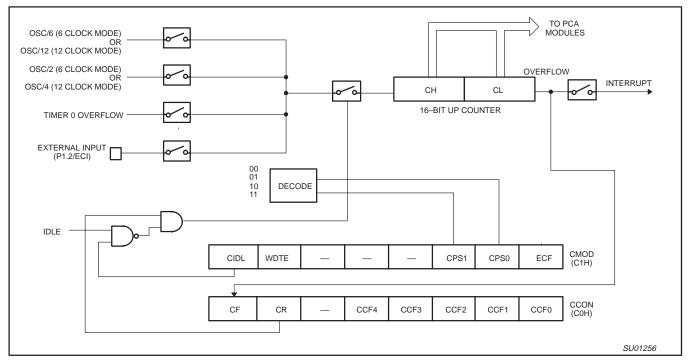


Figure 36. PCA Timer/Counter

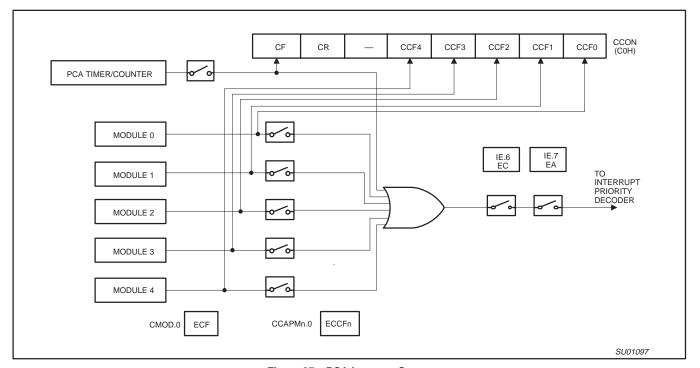


Figure 37. PCA Interrupt System

## P8xC660X2/661X2

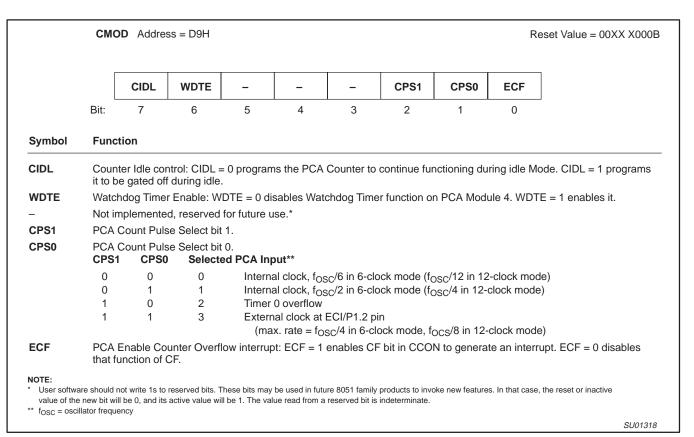


Figure 38. CMOD: PCA Counter Mode Register

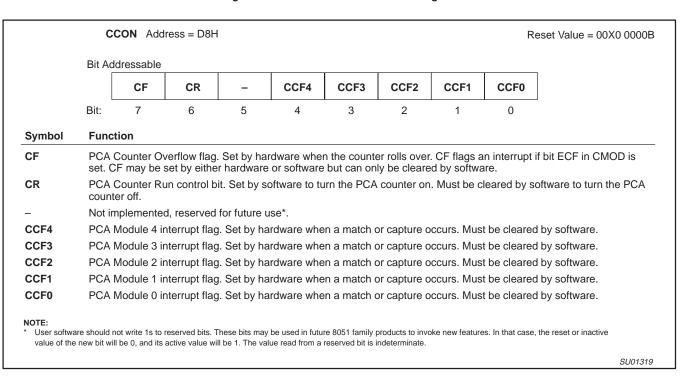


Figure 39. CCON: PCA Counter Control Register

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## P8xC660X2/661X2

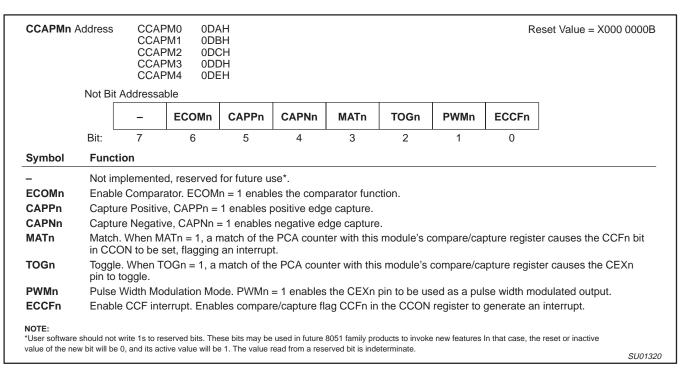


Figure 40. CCAPMn: PCA Modules Compare/Capture Registers

_	ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMn	ECCFn	MODULE FUNCTION
Х	0	0	0	0	0	0	0	No operation
Х	Х	1	0	0	0	0	Х	16-bit capture by a positive-edge trigger on CEXn
Х	Х	0	1	0	0	0	Х	16-bit capture by a negative trigger on CEXn
Х	Х	1	1	0	0	0	Х	16-bit capture by a transition on CEXn
Х	1	0	0	1	0	0	Х	16-bit Software Timer
Х	1	0	0	1	1	0	Х	16-bit High Speed Output
Х	1	0	0	0	0	1	0	8-bit PWM
Х	1	0	0	1	Х	0	Х	Watchdog Timer

Figure 41. PCA Module Modes (CCAPMn Register)

### **PCA Capture Mode**

To use one of the PCA modules in the capture mode either one or both of the CCAPM bits CAPN and CAPP for that module must be set. The external CEX input for the module (on port 1) is sampled for a transition. When a valid transition occurs the PCA hardware loads the value of the PCA counter registers (CH and CL) into the module's capture registers (CCAPnL and CCAPnH). If the CCFn bit for the module in the CCON SFR and the ECCFn bit in the CCAPMn SFR are set then an interrupt will be generated. Refer to Figure 42.

### 16-bit Software Timer Mode

The PCA modules can be used as software timers by setting both the ECOM and MAT bits in the modules CCAPMn register. The PCA timer will be compared to the module's capture registers and when a match occurs an interrupt will occur if the CCFn (CCON SFR) and the ECCFn (CCAPMn SFR) bits for the module are both set (see Figure 43).

### **High Speed Output Mode**

In this mode the CEX output (on port 1) associated with the PCA module will toggle each time a match occurs between the PCA

counter and the module's capture registers. To activate this mode the TOG, MAT, and ECOM bits in the module's CCAPMn SFR must be set (see Figure 44).

### **Pulse Width Modulator Mode**

All of the PCA modules can be used as PWM outputs. Figure 45 shows the PWM function. The frequency of the output depends on the source for the PCA timer. All of the modules will have the same frequency of output because they all share the PCA timer. The duty cycle of each module is independently variable using the module's capture register CCAPLn. When the value of the PCA CL SFR is less than the value in the module's CCAPLn SFR the output will be LOW, when it is equal to or greater than the output will be HIGH. When CL overflows from FF to 00, CCAPLn is reloaded with the value in CCAPHn. the allows updating the PWM without glitches. The PWM and ECOM bits in the module's CCAPMn register must be set to enable the PWM mode.

## P8xC660X2/661X2

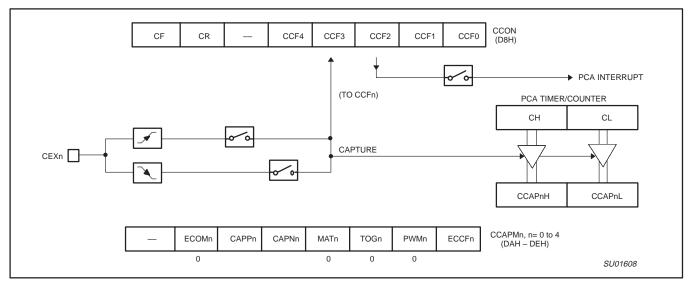


Figure 42. PCA Capture Mode

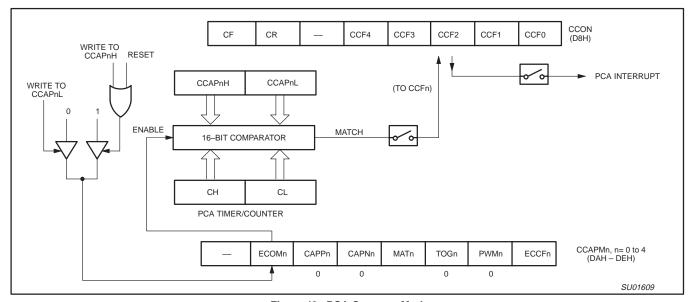


Figure 43. PCA Compare Mode

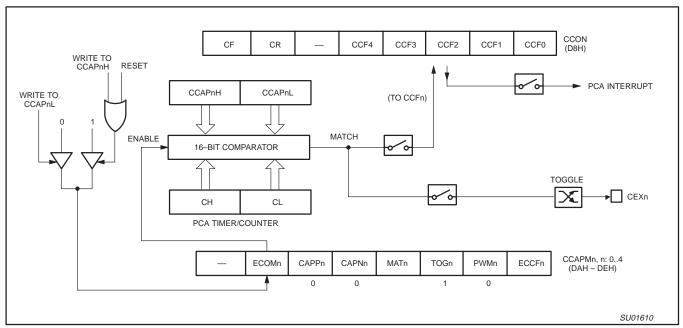


Figure 44. PCA High Speed Output Mode

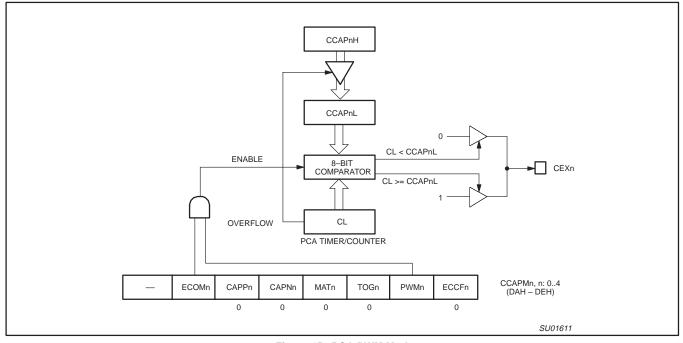


Figure 45. PCA PWM Mode

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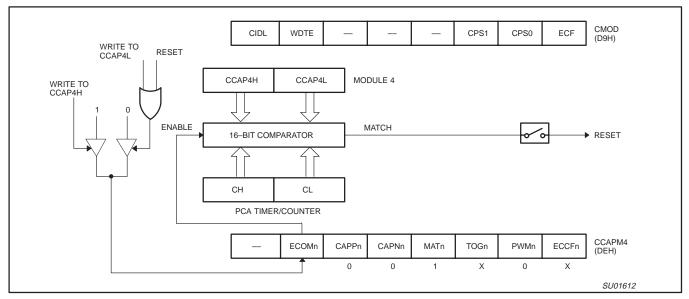


Figure 46. PCA Watchdog Timer mode (Module 4 only)

#### **PCA Watchdog Timer**

An on-board watchdog timer is available with the PCA to improve the reliability of the system without increasing chip count. Watchdog timers are useful for systems that are susceptible to noise, power glitches, or electrostatic discharge. Module 4 is the only PCA module that can be programmed as a watchdog. However, this module can still be used for other modes if the watchdog is not needed.

Figure 46 shows a diagram of how the watchdog works. The user pre-loads a 16-bit value in the compare registers. Just like the other compare modes, this 16-bit value is compared to the PCA timer value. If a match is allowed to occur, an internal reset will be generated. This will not cause the RST pin to be driven HIGH.

In order to hold off the reset, the user has three options:

- periodically change the compare value so it will never match the PCA timer.
- 2. periodically change the PCA timer value so it will never match the compare values, or
- disable the watchdog by clearing the WDTE bit before a match occurs and then re-enable it.

The first two options are more reliable because the watchdog timer is never disabled as in option #3. If the program counter ever goes astray, a match will eventually occur and cause an internal reset. The second option is also not recommended if other PCA modules are being used. Remember, the PCA timer is the time base for all modules; changing the time base for other modules would not be a good idea. Thus, in most applications the first solution is the best option.

Figure 47 shows the code for initializing the watchdog timer. Module 4 can be configured in either compare mode, and the WDTE bit in CMOD must also be set. The user's software then must periodically change (CCAP4H,CCAP4L) to keep a match from occurring with the PCA timer (CH,CL). This code is given in the WATCHDOG routine in Figure 47.

This routine should not be part of an interrupt service routine, because if the program counter goes astray and gets stuck in an infinite loop, interrupts will still be serviced and the watchdog will keep getting reset. Thus, the purpose of the watchdog would be defeated. Instead, call this subroutine from the main program within  $2^{16}$  count of the PCA timer.

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## P8xC660X2/661X2

```
INIT_WATCHDOG:
  MOV CCAPM4, #4CH ; Module 4 in compare mode MOV CCAP4L, #0FFH ; Write to LOW byte first
  MOV CCAP4H, #0FFH
                         ; Before PCA timer counts up to
                         ; FFFF Hex, these compare values
                          ; must be changed
  ORL CMOD, #40H
                          ; Set the WDTE bit to enable the
                          ; watchdog timer without changing
                          ; the other bits in CMOD
; Main program goes here, but CALL WATCHDOG periodically.
; ***********************
WATCHDOG:
                         ; Hold off interrupts
  CLR EA
  MOV CCAP4L, #00
                        ; Next compare value is within
  MOV CCAP4H, CH
                          ; 255 counts of the current PCA
  SETB EA
                          ; timer value
  RET
```

Figure 47. PCA Watchdog Timer Initialization Code

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## P8xC660X2/661X2

#### **Expanded Data RAM Addressing**

The P8xC660X2/661X2 has internal data memory that is mapped into four separate segments: the lower 128 bytes of RAM, upper 128 bytes of RAM, 128 bytes Special Function Register (SFR), and 256 bytes expanded RAM (ERAM) (768 bytes for the RD2).

The four segments are:

- The Lower 128 bytes of RAM (addresses 00H to 7FH) are directly and indirectly addressable.
- The Upper 128 bytes of RAM (addresses 80H to FFH) are indirectly addressable only.
- 3. The Special Function Registers, SFRs, (addresses 80H to FFH) are directly addressable only.
- The 256/768-bytes expanded RAM (ERAM, 00H 1FFH/2FFH) are indirectly accessed by move external instruction, MOVX, and with the EXTRAM bit cleared, see Figure 48.

The Lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFR. That means they have the same address, but are physically separate from SFR space.

When an instruction accesses an internal location above address 7FH, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction. Instructions that use direct addressing access SFR space. For example:

MOV 0A0H,#data

accesses the SFR at location 0A0H (which is P2). Instructions that use indirect addressing access the Upper 128 bytes of data RAM.

For example:

MOV @R0,acc

where R0 contains 0A0H, accesses the data byte at address 0A0H, rather than P2 (whose address is 0A0H).

The ERAM can be accessed by indirect addressing, with EXTRAM bit cleared and MOVX instructions. This part of memory is physically located on-chip, logically occupies the first 256/768 bytes of external data memory in the P8xC660X2/661X2.

With EXTRAM = 0, the ERAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to ERAM will not affect ports P0, P3.6 (WR#) and P3.7 (RD#). P2 SFR is output during external addressing. For example, with EXTRAM = 0,

MOVX @R0,acc

where R0 contains 0A0H, accesses the ERAM at address 0A0H rather than external memory. An access to external data memory locations higher than the ERAM will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, so with P0 and P2 as data/address bus, and P3.6 and P3.7 as write and read timing signals. Refer to Figure 49.

With EXTRAM = 1, MOVX @Ri and MOVX @DPTR will be similar to the standard 80C51. MOVX @ Ri will provide an 8-bit address multiplexed with data on Port 0 and any output port pins can be used to output higher order address bits. This is to provide the external paging capability. MOVX @DPTR will generate a 16-bit address. Port 2 outputs the high-order eight address bits (the contents of DPH) while Port 0 multiplexes the low-order eight address bits (DPL) with data. MOVX @Ri and MOVX @DPTR will generate either read or write signals on P3.6 ( $\overline{WR}$ ) and P3.7 ( $\overline{RD}$ ).

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may not be located in the ERAM.

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## P8xC660X2/661X2

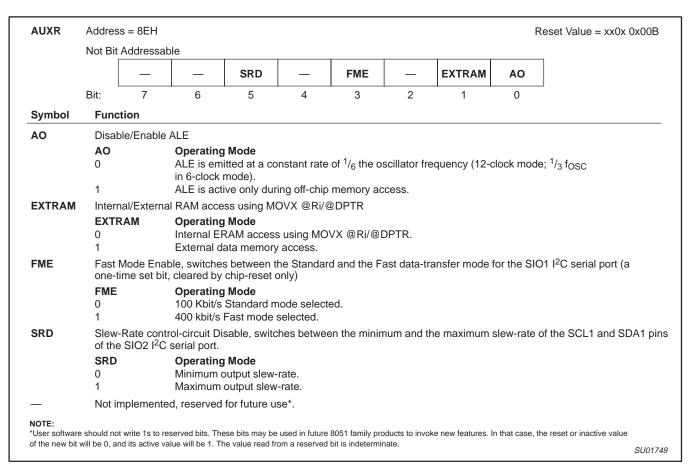


Figure 48. AUXR: Auxiliary Register

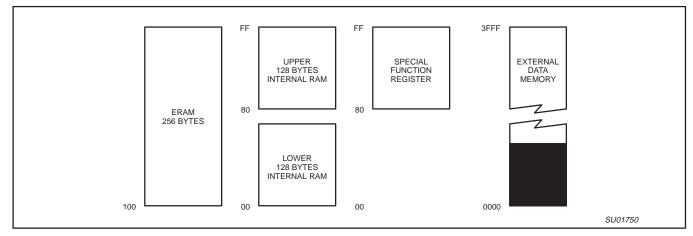


Figure 49. Internal and External Data Memory Address Space with EXTRAM = 0

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## P8xC660X2/661X2

# HARDWARE WATCHDOG TIMER (ONE-TIME ENABLED WITH RESET-OUT FOR P8XC66xX2)

The WDT is intended as a recovery method in situations where the CPU may be subjected to software upset. The WDT consists of a 14-bit counter and the WatchDog Timer reset (WDTRST) SFR. The WDT is disabled at reset. To enable the WDT, the user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When the WDT is enabled, it will increment every machine cycle while the oscillator is running and there is no way to disable the WDT except through reset (either hardware reset or WDT overflow reset). When the WDT overflows, it will drive an output reset HIGH pulse at the RST-pin (see the note below).

#### Using the WDT

To enable the WDT, the user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When the WDT is enabled, the user needs to service it by writing 01EH and 0E1H to WDTRST to avoid a WDT overflow. The 14-bit counter overflows when it reaches 16383 (3FFFH) and this will reset the device. When the WDT is enabled, it will increment every machine cycle while the oscillator is running. This means the user must reset the WDT at least every 16383 machine cycles. To reset the WDT, the user must write 01EH and 0E1H to WDTRST. WDTRST is a write only register. The WDT counter cannot be read or written. When the WDT overflows, it will generate an output RESET pulse at the reset pin (see note below). The RESET pulse duration is  $98 \times T_{\rm OSC}$  (6-clock mode; 196 in 12-clock mode), where  $T_{\rm OSC} = 1/f_{\rm OSC}$ . To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset

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## P8xC660X2/661X2

## **ABSOLUTE MAXIMUM RATINGS**1, 2, 3

PARAMETER	RATING	UNIT
Operating temperature under bias	0 to +70 or -40 to +85	°C
Storage temperature range	-65 to +150	°C
Voltage on EA/V <sub>PP</sub> pin to V <sub>SS</sub>	0 to +13.0	V
Voltage on any other pin to V <sub>SS</sub> <sup>4</sup>	-0.5 to +6.0	V
Maximum I <sub>OL</sub> per I/O pin	15	mA
Power dissipation (based on package heat transfer limitations, not device power consumption)	1.5	W

#### NOTES:

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any conditions other than those described in the AC and DC Electrical Characteristics section of this specification is not implied.

This product includes circuitry specifically designed for the protection of its internal devices from the damaging effects of excessive static charge. Nonetheless, it is suggested that conventional precautions be taken to avoid applying greater than the rated maximum.
 Parameters are valid over operating temperature range unless otherwise specified. All voltages are with respect to V<sub>SS</sub> unless otherwise

4. Transient voltage only.

#### **AC ELECTRICAL CHARACTERISTICS**

 $T_{amb} = 0$ °C to +70°C or -40°C to +85°C

					CLOCK FREQUENCY RANGE		
SYMBOL	FIGURE	PARAMETER	OPERATING MODE	POWER SUPPLY VOLTAGE	MIN	MAX	UNIT
1/t <sub>CLCL</sub>	55	Oscillator frequency	6-clock	5 V ± 10%	0	30	MHz
			6-clock	2.7 V to 5.5 V	0	16	MHz
			12-clock	5 V ± 10%	0	33	MHz
			12-clock	2.7 V to 5.5 V	0	16	MHz

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#### DC ELECTRICAL CHARACTERISTICS

 $T_{amb} = 0$  °C to +70 °C or -40 °C to +85 °C;  $V_{CC} = 2.7$  V to 5.5 V;  $V_{SS} = 0$  V (16 MHz max. CPU clock)

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP <sup>1</sup>	MAX	1
V <sub>IL</sub>	Input LOW voltage <sup>11</sup> (except EA, SCL, SDA)	4.0 V < V <sub>CC</sub> < 5.5 V	-0.5	1	0.2 V <sub>CC</sub> -0.1	V
		2.7 V < V <sub>CC</sub> < 4.0 V	-0.5	1	0.7 V <sub>CC</sub>	V
V <sub>IL1</sub>	LOW level input voltage EA		-0.5		0.2 V <sub>DD</sub> -0.35	V
V <sub>IH</sub>	Input HIGH voltage (ports 0, 1, 2, 3, EA)		0.2 V <sub>CC</sub> +0.9		V <sub>CC</sub> +0.5	V
V <sub>IH1</sub>	Input HIGH voltage, XTAL1, RST <sup>11</sup>		0.7 V <sub>CC</sub>		V <sub>CC</sub> +0.5	V
V <sub>IH2</sub>	Input HIGH voltage, SDL and SDA <sup>13</sup>		0.7 V <sub>DD</sub>		5.5	V
V <sub>OL</sub>	Output LOW voltage, ports 1, 28	$V_{CC} = 2.7 \text{ V; } I_{OL} = 1.6 \text{ mA}^2$	_		0.4	V
V <sub>OL1</sub>	Output LOW voltage, port 0, ALE, PSEN <sup>7,8</sup>	$V_{CC} = 2.7 \text{ V}; I_{OL} = 3.2 \text{ mA}^2$	_		0.4	V
V <sub>OH</sub>	Output HIGH voltage, ports 1, 2, 3 <sup>3</sup>	$V_{CC} = 2.7 \text{ V; } I_{OH} = -20 \mu\text{A}$	V <sub>CC</sub> - 0.7		_	V
		$V_{CC} = 4.5 \text{ V; } I_{OH} = -30 \mu\text{A}$	V <sub>CC</sub> - 0.7		_	V
V <sub>OH1</sub>	Output HIGH voltage (port 0 in external bus mode), ALE <sup>9</sup> , PSEN <sup>3</sup>	$V_{CC} = 2.7 \text{ V}; I_{OH} = -3.2 \text{ mA}$	V <sub>CC</sub> - 0.7		-	V
V <sub>HYS</sub>	Hysteresis of Schmitt Trigger inputs SCL and SDA (Fast Mode)		0.5V <sub>DD</sub> <sup>14</sup>		-	V
I <sub>IL</sub>	Logical 0 input current, ports 1, 2, 3	V <sub>IN</sub> = 0.4 V	-1		-50	μΑ
I <sub>TL</sub>	Logical 1-to-0 transition current, ports 1, 2, 36	V <sub>IN</sub> = 2.0 V; See note 4	-		-650	μΑ
ILI	Input leakage current, port 0	$0.45 < V_{IN} < V_{CC} - 0.3$	_		±10	μΑ
I <sub>LI2</sub>	Input leakage current SCL and SDA	0 V < V <sub>IN</sub> < 5.5 V	_	1	±10	μΑ
		0 V < V <sub>DD</sub> < 5.5 V				
I <sub>CC</sub>	Power supply current (see Figure 58 and Source Code):			1		
	Active mode @ 16 MHz					μΑ
	Idle mode @ 16 MHz					μΑ
	Power-down mode or clock stopped (see Figure 54 for conditions) <sup>12</sup>	T <sub>amb</sub> = 0 °C to 70 °C		2	30	μΑ
	,	$T_{amb} = -40 ^{\circ}\text{C} \text{ to } +85 ^{\circ}\text{C}$		3	50	μΑ
V <sub>RAM</sub>	RAM keep-alive voltage		1.2	1		V
R <sub>RST</sub>	Internal reset pull-down resistor		40	1	225	kΩ
C <sub>IO</sub>	Pin capacitance <sup>10</sup> (except EA)		_	1	15	pF

#### NOTES:

- 1. Typical ratings are not guaranteed. Values listed are based on tests conducted on limited number of samples at room temperature.
- 2. Capacitive loading on ports 0 and 2 may cause spurious noise to be superimposed on the V<sub>OL</sub>s of ALE and ports 1 and 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operations. In the worst cases (capacitive loading > 100 pF), the noise pulse on the ALE pin may exceed 0.8 V. In such cases, it may be desirable to qualify ALE with a Schmitt Trigger, or use an address latch with a Schmitt Trigger STROBE input. IOL can exceed these conditions provided that no single output sinks more than 5 mA and no more than two outputs exceed the test conditions.
- 3. Capacitive loading on ports 0 and 2 may cause the V<sub>OH</sub> on ALE and PSEN to momentarily fall below the V<sub>CC</sub>-0.7 specification when the address bits are stabilizing.
- Pins of ports 1, 2 and 3 source a transition current when they are being externally driven from 1 to 0. The transition current reaches its maximum value when  $V_{\text{IN}}$  is approximately 2 V.
- See Figures 60 through 63 for I<sub>CC</sub> test conditions and Figure 58 for I<sub>CC</sub> vs. Frequency 12-clock mode characteristics:

Active mode (operating):  $I_{CC} = 1.0 \text{ mA} + 1.1 \text{ mA} \times \text{FREQ.[MHz]}$  $I_{CC}$  = 7.0 mA + 1.1 mA × FREQ.[MHz] Active mode (reset):

- Idle mode:  $I_{CC} = 1.0 \text{ mA} + 0.44 \text{ mA} \times \text{FREQ.[MHz]}$ 6. This value applies to  $I_{amb} = 0 ^{\circ}\text{C}$  to  $+70 ^{\circ}\text{C}$ . For  $I_{amb} = -40 ^{\circ}\text{C}$  to  $+85 ^{\circ}\text{C}$ ,  $I_{TL} = -750 \,\mu\text{A}$ .
  7. Load capacitance for port 0, ALE, and  $\overline{PSEN} = 100 \text{ pF}$ , load capacitance for all other outputs = 80 pF.
- Under steady state (non-transient) conditions, I<sub>OL</sub> must be externally limited as follows: Maximum IOL per port pin: 15 mA (\*NOTE: This is 85 °C specification.)

Maximum I<sub>OL</sub> per 8-bit port: 26 mA Maximum total I<sub>OL</sub> for all outputs: 71 mA

If I<sub>OL</sub> exceeds the test condition, V<sub>OL</sub> may exceed the related specification. Pins are not guaranteed to sink current greater than the listed

9. ALE is tested to  $V_{OH1}$ , except when ALE is off then  $V_{OH}$  is the voltage specification.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

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10. Pin capacitance is characterized but not tested. Pin capacitance is less than 25 pF. Pin capacitance of ceramic package is less than 15 pF (except EA is 25 pF).

- 11. To improve noise rejection a nominal 100 ns glitch rejection circuitry has been added to the RST pin, and a nominal 15 ns glitch rejection circuitry has been added to the INT0 and INT1 pins. Previous devices provided only an inherent 5 ns of glitch rejection.
- 12. Power down mode for 3 V range: Commercial Temperature Range typ: 0.5 μA, max. 20 μA; Industrial Temperature Range typ. 1.0 μA, max. 30 μA;
- 13. The input threshold voltage of SCL and SDA (SIO1) meets the  $I^2C$  specification, so an input voltage below 0.3  $V_{DD}$  will be recognized as a logic 0 while an input voltage above 0.7  $V_{DD}$  will be recognized as a logic 1.
- 14. Not 100% tested.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

## P8xC660X2/661X2

#### DC ELECTRICAL CHARACTERISTICS

 $T_{amb} = 0$  °C to +70 °C or -40 °C to +85 °C;  $V_{CC} = 5$  V ±10%;  $V_{SS} = 0$  V (30/33 MHz max. CPU clock)

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP <sup>1</sup>	MAX	1
V <sub>IL</sub>	Input LOW voltage <sup>11</sup> (except EA, SCL, SDA)	4.5 V < V <sub>CC</sub> < 5.5 V	-0.5	1	0.2 V <sub>CC</sub> -0.1	V
V <sub>IL1</sub>	LOW level input voltage EA		-0.5		0.2 V <sub>DD</sub> -0.35	V
V <sub>IH</sub>	Input HIGH voltage (ports 0, 1, 2, 3, EA)		0.2 V <sub>CC</sub> +0.9		V <sub>CC</sub> +0.5	V
V <sub>IH1</sub>	Input HIGH voltage, XTAL1, RST <sup>11</sup>		0.7 V <sub>CC</sub>		V <sub>CC</sub> +0.5	V
V <sub>IH2</sub>	Input HIGH voltage, SDL and SDA <sup>12</sup>		0.7 V <sub>DD</sub>		5.5	V
V <sub>OL</sub>	Output LOW voltage, ports 1, 2, 3 8	$V_{CC} = 4.5 \text{ V}; I_{OL} = 1.6 \text{ mA}^2$	-		0.4	V
V <sub>OL1</sub>	Output LOW voltage, port 0, ALE, PSEN 7, 8	$V_{CC} = 4.5 \text{ V}; I_{OL} = 3.2 \text{ mA}^2$	-		0.4	V
V <sub>OH</sub>	Output HIGH voltage, ports 1, 2, 3 3	$V_{CC} = 4.5 \text{ V}; I_{OH} = -30 \mu\text{A}$	V <sub>CC</sub> - 0.7		-	V
V <sub>OH1</sub>	Output HIGH voltage (port 0 in external bus mode), ALE <sup>9</sup> , PSEN <sup>3</sup>	$V_{CC} = 4.5 \text{ V}; I_{OH} = -3.2 \text{ mA}$	V <sub>CC</sub> - 0.7		-	V
V <sub>HYS</sub>	Hysteresis of Schmitt Trigger inputs SCL and SDA (Fast Mode) <sup>13</sup>		0.5V <sub>DD</sub>		-	V
I <sub>IL</sub>	Logical 0 input current, ports 1, 2, 3	V <sub>IN</sub> = 0.4 V	<b>–</b> 1		-50	μΑ
I <sub>TL</sub>	Logical 1-to-0 transition current, ports 1, 2, 3 6	V <sub>IN</sub> = 2.0 V; See note 4	-		-650	μΑ
I <sub>LI</sub>	Input leakage current, port 0	$0.45 < V_{IN} < V_{CC} - 0.3$	-		±10	μΑ
I <sub>LI2</sub>	Input leakage current SCL and SDA	0 V < V <sub>IN</sub> < 5.5 V	_		±10	μΑ
		0 V < V <sub>DD</sub> < 5.5 V				
I <sub>CC</sub>	Power supply current					
	Active mode (see Note 5)					
	Idle mode (see Note 5)					
	Power-down mode or clock stopped (see Figure 63 for conditions)	T <sub>amb</sub> = 0 °C to 70 °C		2	30	μΑ
		$T_{amb} = -40 ^{\circ}\text{C} \text{ to } +85 ^{\circ}\text{C}$		3	50	μΑ
$V_{RAM}$	RAM keep-alive voltage		1.2			V
R <sub>RST</sub>	Internal reset pull-down resistor		40		225	kΩ
C <sub>IO</sub>	Pin capacitance <sup>10</sup> (except <del>EA</del> )		<b> </b>		15	pF

## NOTES:

- 1. Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
- 2. Capacitive loading on ports 0 and 2 may cause spurious noise to be superimposed on the V<sub>OL</sub>s of ALE and ports 1 and 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operations. In the worst cases (capacitive loading > 100 pF), the noise pulse on the ALE pin may exceed 0.8 V. In such cases, it may be desirable to qualify ALE with a Schmitt Trigger, or use an address latch with a Schmitt Trigger STROBE input. I<sub>OL</sub> can exceed these conditions provided that no single output sinks more than 5 mA and no more than two outputs exceed the test conditions.
- 3. Capacitive loading on ports 0 and 2 may cause the V<sub>OH</sub> on ALE and PSEN to momentarily fall below the V<sub>CC</sub>-0.7 specification when the address bits are stabilizing.
- Pins of ports 1, 2 and 3 source a transition current when they are being externally driven from 1 to 0. The transition current reaches its maximum value when V<sub>IN</sub> is approximately 2 V.
- 5. See Figures 60 through 63 for  $I_{CC}$  test conditions and Figure 58 for  $I_{CC}$  vs. Frequency.

12-clock mode characteristics:

Active mode (operating):  $I_{CC} = 1.0 \text{ mA} + 1.1 \text{ mA} \times \text{FREQ.[MHz]}$ Active mode (reset):  $I_{CC} = 7.0 \text{ mA} + 1.1 \text{ mA} \times \text{FREQ.[MHz]}$  $I_{CC} = 1.0 \text{ mA} + 0.44 \text{ mA} \times \text{FREQ.[MHz]}$ 

- 6. This value applies to  $T_{amb} = 0^{\circ}C$  to  $+70^{\circ}C$ . For  $T_{amb} = -40^{\circ}C$  to  $+85^{\circ}C$ ,  $I_{TL} = -750 \ \mu A$ .
- 7. Load capacitance for port 0, ALE, and  $\overline{\text{PSEN}}$  = 100 pF, load capacitance for all other outputs = 80 pF.
- Under steady state (non-transient) conditions, I<sub>OL</sub> must be externally limited as follows: Maximum I<sub>OL</sub> per port pin: 15 mA (\*NOTE: This is 85 °C specification.)

Maximum I<sub>OL</sub> per 8-bit port: 26 mA Maximum total I<sub>OL</sub> for all outputs: 71 mA

If  $I_{OL}$  exceeds the test condition,  $V_{OL}$  may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.

- ALE is tested to V<sub>OH1</sub>, except when ALE is off then V<sub>OH</sub> is the voltage specification.
- 10. Pin capacitance is characterized but not tested. Pin capacitance is less than 25 pF. Pin capacitance of ceramic package is less than 15 pF (except EA is 25 pF).
- 11. To improve noise rejection a nominal 100 ns glitch rejection circuitry has been added to the RST pin, and a nominal 15 ns glitch rejection circuitry has been added to the INTO and INTO pins. Previous devices provided only an inherent 5 ns of glitch rejection.

 $80C51\ 8\text{-bit}$  microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB  $I^2C$  interfaces

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<sup>12.</sup> The input threshold voltage of SCL and SDA (SIO1) meets the I<sup>2</sup>C specification, so an input voltage below 0.3 V<sub>DD</sub> will be recognized as a logic 0 while an input voltage above 0.7 V<sub>DD</sub> will be recognized as a logic 1.
13. Not 100% tested.

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Product data

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I2C interfaces

## AC ELECTRICAL CHARACTERISTICS (12-CLOCK MODE, 5 V ±10% OPERATION)

 $T_{amb} = 0 \, ^{\circ}C \text{ to } +70 \, ^{\circ}C \text{ or } -40 \, ^{\circ}C \text{ to } +85 \, ^{\circ}C \text{ ; } V_{CC} = 5 \, ^{\circ}V \pm 10\%, \, V_{SS} = 0 \, V^{1,2,3,4}$ 

Symbol	Figure	Parameter	Limits		16 MHz	Clock	Unit
			MIN	MAX	MIN	MAX	٦
1/t <sub>CLCL</sub>	55	Oscillator frequency	0	33			MHz
LHLL	50	ALE pulse width	2 t <sub>CLCL</sub> -8		117		ns
AVLL	50	Address valid to ALE LOW	t <sub>CLCL</sub> -13		49.5		ns
tLLAX	50	Address hold after ALE LOW	t <sub>CLCL</sub> -20		42.5		ns
t <sub>LLIV</sub>	50	ALE LOW to valid instruction in		4 t <sub>CLCL</sub> -35		215	ns
LLPL	50	ALE LOW to PSEN LOW	t <sub>CLCL</sub> -10		52.5		ns
t <sub>PLPH</sub>	50	PSEN pulse width	3 t <sub>CLCL</sub> -10		177.5		ns
t <sub>PLIV</sub>	50	PSEN LOW to valid instruction in		3 t <sub>CLCL</sub> -35		152.5	ns
t <sub>PXIX</sub>	50	Input instruction hold after PSEN	0		0		ns
t <sub>PXIZ</sub>	50	Input instruction float after PSEN		t <sub>CLCL</sub> -10		52.5	ns
t <sub>AVIV</sub>	50	Address to valid instruction in		5 t <sub>CLCL</sub> -35		277.5	ns
t <sub>PLAZ</sub>	50	PSEN LOW to address float		10		10	ns
Data Men	nory		1				
t <sub>RLRH</sub>	51	RD pulse width	6 t <sub>CLCL</sub> -20		355		ns
twlwh	52	WR pulse width	6 t <sub>CLCL</sub> –20		355		ns
RLDV	51	RD LOW to valid data in	5 t <sub>CLCL</sub> –35			277.5	ns
RHDX	51	Data hold after RD	0	0202	0		ns
t <sub>RHDZ</sub>	51	Data float after RD		2 t <sub>CLCL</sub> -10		115	ns
t <sub>LLDV</sub>	51	ALE LOW to valid data in		8 t <sub>CLCL</sub> -35		465	ns
t <sub>AVDV</sub>	51	Address to valid data in		9 t <sub>CLCL</sub> -35		527.5	ns
t <sub>LLWL</sub>	51, 52	ALE LOW to RD or WR LOW	3 t <sub>CLCL</sub> -15	3 t <sub>CLCL</sub> +15	172.5	202.5	ns
t <sub>AVWL</sub>	51, 52	Address valid to WR LOW or RD LOW	4 t <sub>CLCL</sub> -15	0202	235		ns
t <sub>QVWX</sub>	52	Data valid to WR transition	t <sub>CLCL</sub> –25		37.5		ns
t <sub>WHQX</sub>	52	Data hold after WR	t <sub>CLCL</sub> –15		47.5		ns
t <sub>QVWH</sub>	52	Data valid to WR HIGH	7 t <sub>CLCL</sub> –5		432.5		ns
t <sub>RLAZ</sub>	51	RD LOW to address float	0101	0		0	ns
twhlh	51, 52	RD or WR HIGH to ALE HIGH	t <sub>CLCL</sub> -10	t <sub>CLCL</sub> +10	52.5	72.5	ns
External (			0202	0101			
t <sub>CHCX</sub>	55	High time	0.32 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CLCX</sub>			ns
tclcx	55	Low time	0.32 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CHCX</sub>			ns
CLCH	55	Rise time	0202	5			ns
CHCL	55	Fall time		5			ns
Shift regi	ster		1	-1	1		
t <sub>XLXL</sub>	54	Serial port clock cycle time	12 t <sub>CLCL</sub>		750		ns
t <sub>QVXH</sub>	54	Output data setup to clock rising edge	10 t <sub>CLCL</sub> –25		600		ns
t <sub>XHQX</sub>	54	Output data hold after clock rising edge	2 t <sub>CLCL</sub> -15		110		ns
t <sub>XHDX</sub>	54	Input data hold after clock rising edge	0		0		ns
t <sub>XHDV</sub>	54	Clock rising edge to input data valid <sup>5</sup>		10 t <sub>CLCL</sub> -133		492	ns

- Parameters are valid over operating temperature range unless otherwise specified.
   Load capacitance for port 0, ALE, and PSEN = 100 pF, load capacitance for all outputs = 80 pF
- 3. Interfacing the microcontroller to devices with float time up to 45 ns is permitted. This limited bus contention will not cause damage to port 0
- 4. Parts are guaranteed by design to operate down to 0 Hz.
  5. Below 16 MHz this parameter is 8 t<sub>CLCL</sub> 133.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I2C interfaces

## P8xC660X2/661X2

## AC ELECTRICAL CHARACTERISTICS (12-CLOCK MODE, 2.7 V TO 5.5 V OPERATION)

 $T_{amb} = 0$  °C to +70 °C or -40 °C to +85 °C;  $V_{CC} = 2.7$  V to 5.5 V,  $V_{SS} = 0$  V<sup>1,2,3,4</sup>

Symbol	Figure	Parameter	Limits		16 MHz	Clock	Unit
			MIN	MAX	MIN	MAX	
1/t <sub>CLCL</sub>	55	Oscillator frequency	0	16			MHz
LHLL	50	ALE pulse width	2t <sub>CLCL</sub> -10		115		ns
AVLL	50	Address valid to ALE LOW	t <sub>CLCL</sub> -15		47.5		ns
LLAX	50	Address hold after ALE LOW	t <sub>CLCL</sub> -25		37.5		ns
t <sub>LLIV</sub>	50	ALE LOW to valid instruction in		4 t <sub>CLCL</sub> -55		195	ns
LLPL	50	ALE LOW to PSEN LOW	t <sub>CLCL</sub> -15		47.5		ns
PLPH	50	PSEN pulse width	3 t <sub>CLCL</sub> -15		172.5		ns
t <sub>PLIV</sub>	50	PSEN LOW to valid instruction in		3 t <sub>CLCL</sub> -55		132.5	ns
t <sub>PXIX</sub>	50	Input instruction hold after PSEN	0	0202	0		ns
t <sub>PXIZ</sub>	50	Input instruction float after PSEN		t <sub>CLCL</sub> -10		52.5	ns
t <sub>AVIV</sub>	50	Address to valid instruction in		5 t <sub>CLCL</sub> –50		262.5	ns
t <sub>PLAZ</sub>	50	PSEN LOW to address float	3232			10	ns
Data Men	nory				-	<b>I</b>	
t <sub>RLRH</sub>	51	RD pulse width	6 t <sub>CLCL</sub> -25		350		ns
twlwh	52	WR pulse width	6 t <sub>CLCL</sub> –25		350		ns
t <sub>RLDV</sub>	51	RD LOW to valid data in	- 0000	5 t <sub>CLCL</sub> -50		262.5	ns
RHDX	51	Data hold after RD	0		0		ns
RHDZ	51	Data float after RD		2 t <sub>CLCL</sub> -20	1	105	ns
t <sub>LLDV</sub>	51	ALE LOW to valid data in		8 t <sub>CLCL</sub> -55		445	ns
t <sub>AVDV</sub>	51	Address to valid data in		9 t <sub>CLCL</sub> –50		512.5	ns
t <sub>LLWL</sub>	51, 52	ALE LOW to RD or WR LOW	3 t <sub>CLCL</sub> -20	3 t <sub>CLCL</sub> +20	167.5	207.5	ns
tavwl	51, 52	Address valid to WR LOW or RD LOW	4 t <sub>CLCL</sub> -20	- CLOL	230		ns
tqvwx	52	Data valid to WR transition	t <sub>CLCL</sub> –30		32.5		ns
twhqx	52	Data hold after WR	t <sub>CLCL</sub> –20		42.5		ns
t <sub>QVWH</sub>	52	Data valid to WR HIGH	7 t <sub>CLCL</sub> -10		427.5		ns
t <sub>RLAZ</sub>	51	RD LOW to address float	I JOLOL 15	0	1	0	ns
twhlh	51, 52	RD or WR HIGH to ALE HIGH	t <sub>CLCL</sub> -15	t <sub>CLCL</sub> +15	47.5	77.5	ns
External (			OLOL	TOLOL			
tchcx	55	High time	0.32 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CLCX</sub>	Τ		ns
CLCX	55	Low time	0.32 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CHCX</sub>			ns
CLCH	55	Rise time	GLOL	5			ns
CHCL	55	Fall time		5			ns
Shift regi				1 -	1		1
XLXL	54	Serial port clock cycle time	12 t <sub>CLCL</sub>		750		ns
QVXH	54	Output data setup to clock rising edge	10 t <sub>CLCL</sub> –25		600		ns
t <sub>XHQX</sub>	54	Output data hold after clock rising edge	2 t <sub>CLCL</sub> -15		110		ns
XHQX	54	Input data hold after clock rising edge	0		0		ns
tXHDV	54	Clock rising edge to input data valid <sup>5</sup>		10 t <sub>CLCL</sub> -133	+ -	492	ns

- Parameters are valid over operating temperature range unless otherwise specified.
   Load capacitance for port 0, ALE, and PSEN = 100 pF, load capacitance for all outputs = 80 pF
- 3. Interfacing the microcontroller to devices with float time up to 45 ns is permitted. This limited bus contention will not cause damage to port 0
- 4. Parts are guaranteed by design to operate down to 0 Hz.
  5. Below 16 MHz this parameter is 8 t<sub>CLCL</sub> 133.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I2C interfaces

## P8xC660X2/661X2

## AC ELECTRICAL CHARACTERISTICS (6-CLOCK MODE, 5 V ±10% OPERATION)

 $T_{amb} = 0 \, ^{\circ}\text{C}$  to +70  $^{\circ}\text{C}$  or -40  $^{\circ}\text{C}$  to +85  $^{\circ}\text{C}$  ;  $V_{CC} = 5 \, \text{V} \pm 10\%$ ,  $V_{SS} = 0 \, \text{V}^{1,2,3,4,5}$ 

Symbol	Figure	Parameter	Limits		16 MHz Clock		Unit
			MIN	MAX	MIN	MAX	7
I/t <sub>CLCL</sub>	55	Oscillator frequency	0	30			MHz
LHLL	50	ALE pulse width	t <sub>CLCL</sub> -8		54.5		ns
AVLL	50	Address valid to ALE LOW	0.5 t <sub>CLCL</sub> -13		18.25		ns
LLAX	50	Address hold after ALE LOW	0.5 t <sub>CLCL</sub> -20		11.25		ns
LLIV	50	ALE LOW to valid instruction in		2 t <sub>CLCL</sub> -35		90	ns
LLPL	50	ALE LOW to PSEN LOW	0.5 t <sub>CLCL</sub> -10		21.25		ns
PLPH	50	PSEN pulse width	1.5 t <sub>CLCL</sub> -10		83.75		ns
PLIV	50	PSEN LOW to valid instruction in		1.5 t <sub>CLCL</sub> -35		58.75	ns
PXIX	50	Input instruction hold after PSEN	0		0		ns
PXIZ	50	Input instruction float after PSEN		0.5 t <sub>CLCL</sub> -10		21.25	ns
AVIV	50	Address to valid instruction in		2.5 t <sub>CLCL</sub> -35		121.25	ns
PLAZ	50	PSEN LOW to address float		10		10	ns
Data Men	nory						
RLRH	51	RD pulse width 3 t <sub>CLCL</sub> -20			167.5		ns
WLWH	52	WR pulse width	3 t <sub>CLCL</sub> -20		167.5		ns
RLDV	51	RD LOW to valid data in				121.25	ns
RHDX	51	Data hold after RD	0		0		ns
RHDZ	51	Data float after RD		t <sub>CLCL</sub> -10		52.5	ns
LLDV	51	ALE LOW to valid data in		4 t <sub>CLCL</sub> -35		215	ns
AVDV	51	Address to valid data in		4.5 t <sub>CLCL</sub> -35		246.25	ns
LLWL	51, 52	ALE LOW to RD or WR LOW	1.5 t <sub>CLCL</sub> -15	1.5 t <sub>CLCL</sub> +15	78.75	108.75	ns
AVWL	51, 52	Address valid to WR LOW or RD LOW	2 t <sub>CLCL</sub> -15		110		ns
QVWX	52	Data valid to WR transition	0.5 t <sub>CLCL</sub> -25		6.25		ns
WHQX	52	Data hold after WR	0.5 t <sub>CLCL</sub> -15		16.25		ns
QVWH	52	Data valid to WR HIGH	3.5 t <sub>CLCL</sub> -5		213.75		ns
RLAZ	51	RD LOW to address float		0		0	ns
WHLH	51, 52	RD or WR HIGH to ALE HIGH	0.5 t <sub>CLCL</sub> -10	0.5 t <sub>CLCL</sub> +10	21.25	41.25	ns
External							
СНСХ	55	High time	0.4 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CLCX</sub>			ns
CLCX	55	Low time	0.4 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CHCX</sub>			ns
CLCH	55	Rise time		5			ns
CHCL	55	Fall time		5			ns
Shift regi	ster						
XLXL	54	Serial port clock cycle time	6 t <sub>CLCL</sub>		375		ns
QVXH	54	Output data setup to clock rising edge	5 t <sub>CLCL</sub> –25		287.5		ns
XHQX	54	Output data hold after clock rising edge	t <sub>CLCL</sub> -15		47.5		ns
XHDX	54	Input data hold after clock rising edge	0		0		ns
XHDV	54	Clock rising edge to input data valid <sup>6</sup>		5 t <sub>CLCL</sub> -133		179.5	ns

- 1. Parameters are valid over operating temperature range unless otherwise specified.
- Load capacitance for port 0, ALE, and PSEN=100 pF, load capacitance for all outputs = 80 pF
   Interfacing the microcontroller to devices with float time up to 45ns is permitted. This limited bus contention will not cause damage to port 0
- 4. Parts are guaranteed by design to operate down to 0 Hz.
- 5. Data shown in the table are the best mathematical models for the set of measured values obtained in tests. If a particular parameter calculated at a customer specified frequency has a negative value, it should be considered equal to zero.
- 6. Below 16 MHz this parameter is 4 t<sub>CLCL</sub> 133

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## P8xC660X2/661X2

### AC ELECTRICAL CHARACTERISTICS (6-CLOCK MODE, 2.7 V TO 5.5 V OPERATION)

 $T_{amb} = 0 \, ^{\circ}\text{C}$  to +70  $^{\circ}\text{C}$  or -40  $^{\circ}\text{C}$  to +85  $^{\circ}\text{C}$  :  $V_{CC} = 2.7 \, \text{V}$  to 5.5 V.  $V_{SS} = 0 \, \text{V}^{1,2,3,4,5}$ 

Symbol	Figure	Parameter	Limits		16 MHz (	Clock	Unit
			MIN	MAX	MIN	MAX	
1/t <sub>CLCL</sub>	55	Oscillator frequency	0	16			MHz
LHLL	50	ALE pulse width	t <sub>CLCL</sub> -10		52.5		ns
AVLL	50	Address valid to ALE LOW	0.5 t <sub>CLCL</sub> -15		16.25		ns
tLLAX	50	Address hold after ALE LOW	0.5 t <sub>CLCL</sub> -25		6.25		ns
t <sub>LLIV</sub>	50	ALE LOW to valid instruction in		2 t <sub>CLCL</sub> -55		70	ns
t <sub>LLPL</sub>	50	ALE LOW to PSEN LOW	0.5 t <sub>CLCL</sub> -15		16.25		ns
t <sub>PLPH</sub>	50	PSEN pulse width	1.5 t <sub>CLCL</sub> -15		78.75		ns
t <sub>PLIV</sub>	50	PSEN LOW to valid instruction in		1.5 t <sub>CLCL</sub> -55		38.75	ns
t <sub>PXIX</sub>	50	Input instruction hold after PSEN	0		0		ns
t <sub>PXIZ</sub>	50	Input instruction float after PSEN		0.5 t <sub>CLCL</sub> -10		21.25	ns
t <sub>AVIV</sub>	50	Address to valid instruction in		2.5 t <sub>CLCL</sub> -50		101.25	ns
t <sub>PLAZ</sub>	50	PSEN LOW to address float				10	ns
Data Men	nory		<u> </u>				
t <sub>RLRH</sub>	51	RD pulse width	3 t <sub>CLCL</sub> -25		162.5		ns
t <sub>WLWH</sub>	52	WR pulse width	3 t <sub>CLCL</sub> -25		162.5		ns
t <sub>RLDV</sub>	51	RD LOW to valid data in	2.5 t <sub>CLCL</sub> -50			106.25	ns
tRHDX	51	Data hold after RD	0		0		ns
t <sub>RHDZ</sub>	51	Data float after RD	t <sub>CLCL</sub> -20			42.5	ns
LLDV	51	ALE LOW to valid data in		4 t <sub>CLCL</sub> -55		195	ns
AVDV	51	Address to valid data in		4.5 t <sub>CLCL</sub> -50		231.25	ns
tllwl	51, 52	ALE LOW to RD or WR LOW	1.5 t <sub>CLCL</sub> -20	1.5 t <sub>CLCL</sub> +20	73.75	113.75	ns
t <sub>AVWL</sub>	51, 52	Address valid to WR LOW or RD LOW	2 t <sub>CLCL</sub> -20		105		ns
tqvwx	52	Data valid to WR transition	0.5 t <sub>CLCL</sub> -30		1.25		ns
twhqx	52	Data hold after WR	0.5 t <sub>CLCL</sub> -20		11.25		ns
t <sub>QVWH</sub>	52	Data valid to WR HIGH	3.5 t <sub>CLCL</sub> -10		208.75		ns
t <sub>RLAZ</sub>	51	RD LOW to address float		0		0	ns
t <sub>WHLH</sub>	51, 52	RD or WR HIGH to ALE HIGH	0.5 t <sub>CLCL</sub> -15	0.5 t <sub>CLCL</sub> +15	16.25	46.25	ns
External	Clock						
tснсх	55	High time	0.4 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CLCX</sub>			ns
tclcx	55	Low time	0.4 t <sub>CLCL</sub>	t <sub>CLCL</sub> - t <sub>CHCX</sub>			ns
tclch	55	Rise time		5			ns
tchcl	55	Fall time		5			ns
Shift regi	ster						
t <sub>XLXL</sub>	54	Serial port clock cycle time	6 t <sub>CLCL</sub>		375		ns
t <sub>QVXH</sub>	54	Output data setup to clock rising edge	5 t <sub>CLCL</sub> -25		287.5		ns
t <sub>XHQX</sub>	54	Output data hold after clock rising edge	t <sub>CLCL</sub> -15		47.5		ns
t <sub>XHDX</sub>	54	Input data hold after clock rising edge	0		0		ns
t <sub>XHDV</sub>	54	Clock rising edge to input data valid <sup>6</sup>		5 t <sub>CLCL</sub> -133		179.5	ns

- 1. Parameters are valid over operating temperature range unless otherwise specified.
- Load capacitance for port 0, ALE, and PSEN=100 pF, load capacitance for all outputs = 80 pF
   Interfacing the microcontroller to devices with float time up to 45ns is permitted. This limited bus contention will not cause damage to port 0
- 4. Parts are guaranteed by design to operate down to 0 Hz.
- 5. Data shown in the table are the best mathematical models for the set of measured values obtained in tests. If a particular parameter calculated at a customer specified frequency has a negative value, it should be considered equal to zero.
- 6. Below 16 MHz this parameter is 4 t<sub>CLCL</sub> 133

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### I<sup>2</sup>C-BUS INTERFACE TIMING (5 V, 3.5 MHZ TO 16 MHZ) NOT TESTED, GUARANTEED BY DESIGN

All values referred to  $V_{\text{IH}(\text{min})}$  and  $V_{\text{IL}(\text{max})}$  levels; see Figure TBD

Symbol	Figure	Parameter	I <sup>2</sup> C-BUS				Unit	
			STANDA	RD MODE	FAST MODE			
			MIN	MAX	MIN	MAX	7	
f <sub>SCL</sub>		SCL clock frequency	0	100	0	400	kHz	
t <sub>BUF</sub>		Bus free time between a STOP and START condition	4.7	-	1.3	-	μs	
t <sub>HD;</sub> STA		Hold time (repeated) START condition. After this period, the first clock pulse is generated	4.0	_	0.6	_	μs	
$t_{LOW}$		LOW period of the SCL clock	4.7	_	1.3	-	μs	
t <sub>HIGH</sub>		High period of the SCL clock	4.0	_	0.6	_	μs	
t <sub>SU; STA</sub>		Set-up time for a repeated START condition	4.7	_	0.6	-	μs	
t <sub>HD;DAT</sub>		Data hold time:  – for CBUS compatible masters (notes 1, 3)  – for I <sup>2</sup> C–bus devices (notes 1, 2)	5.0	_ _		_ 0.9	μs	
t <sub>SU;DAT</sub>		Data set-up time	250	-	100 <sup>3</sup>	-	ns	
t <sub>FD</sub> , t <sub>FC</sub>		Rise time of both SDA and SCL signals	-	1000	20 + 0.1 c <sub>b</sub> <sup>4</sup>	300	ns	
t <sub>FD</sub> , t <sub>FC</sub>		Fall time of both SDA and SCL signals	_	300				
t <sub>SU; STO</sub>		Set-up time for STOP condition	4.0	_	0.6	_	μs	
C <sub>b</sub>		Capacitive load for each bus line	_	400	_	400	pF	
t <sub>SP</sub>		Pulse width of spikes which must be suppressed by the input filter	_	_	0	50	ns	

#### NOTES:

- A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V<sub>IHmin</sub> of the SCL signal) in order to bridge the undefined region of the falling edge of SCL.
- 2. The maximum t<sub>HD:DAT</sub> has only to be met if the device does not stretch the LOW period (t<sub>LOW)</sub> of the SCL signal.
- 3. A fast mode l<sup>2</sup>C-bus device can be used in a standard mode l<sup>2</sup>C-bus system, but the requirement t<sub>SU, DAT</sub> > 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line t<sub>R(max)</sub> + t<sub>SU<DAT</sub> = 1000 + 250 = 1250 ns (according to the standard-mode l<sup>2</sup>C-bus specification) before the SCL line is released.
- 4.  $C_b = total$  capacitance of one bus line in pF.

#### **EXPLANATION OF THE AC SYMBOLS**

Each timing symbol has five characters. The first character is always 't' (= time). The other characters, depending on their positions, indicate the name of a signal or the logical status of that signal. The designations are:

A - Address

C - Clock

D - Input data

H - Logic level HIGH

I - Instruction (program memory contents)

L – Logic level LOW, or ALE

P - PSEN

Q - Output data

 $R - \overline{RD}$  signal

t - Time

V - Valid

W- WR signal

X – No longer a valid logic level

Z - Float

**Examples:** t<sub>AVLL</sub> = Time for address valid to ALE LOW.

 $t_{LLPL}$  =Time for ALE LOW to  $\overline{PSEN}$  LOW.

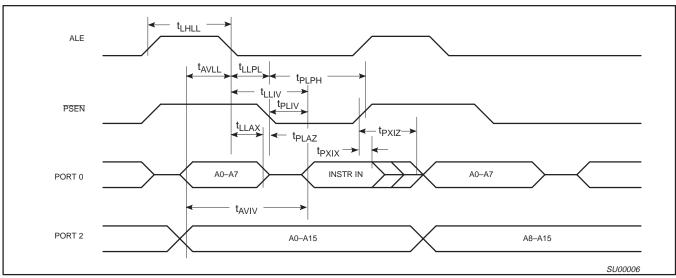


Figure 50. External Program Memory Read Cycle

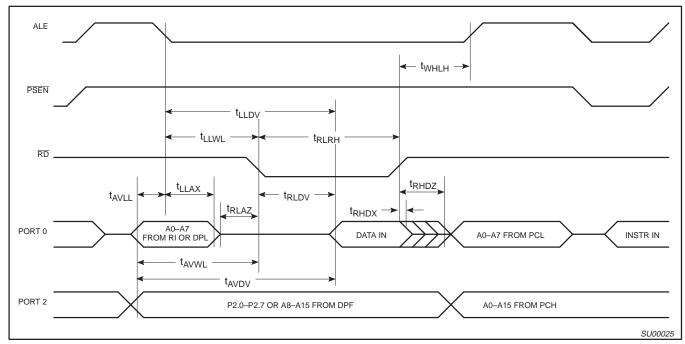


Figure 51. External Data Memory Read Cycle

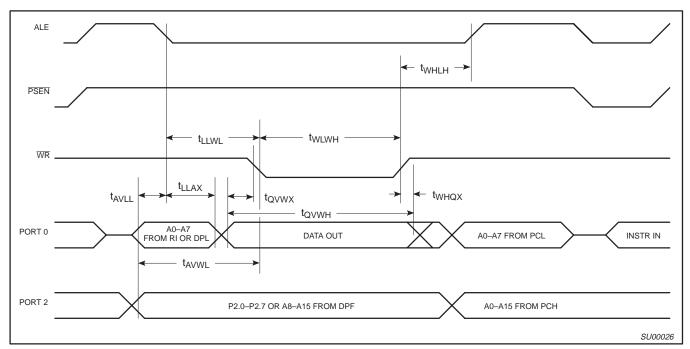


Figure 52. External Data Memory Write Cycle

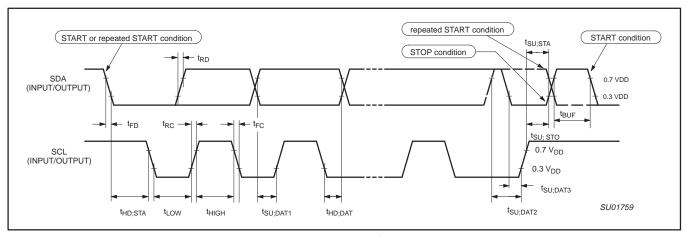


Figure 53. Timing SIO1 (I<sup>2</sup>C) interface

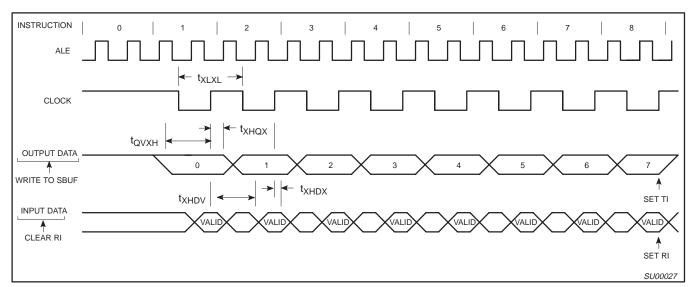


Figure 54. Shift Register Mode Timing

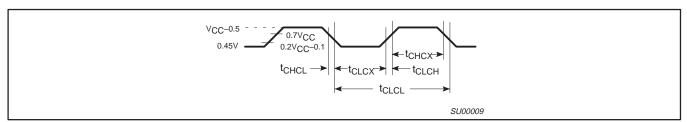


Figure 55. External Clock Drive

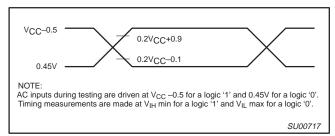


Figure 56. AC Testing Input/Output

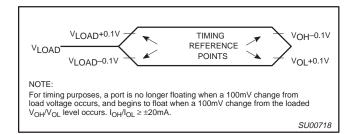


Figure 57. Float Waveform

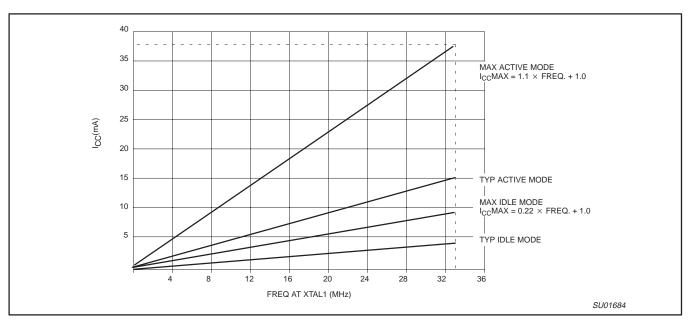


Figure 58. I<sub>CC</sub> vs. FREQ for 12-clock operation Valid only within frequency specifications of the specified operating voltage

```
##
                                  / *js* /
       as31 version V2.10
##
##
##
       source file: idd_ljmp1.asm
         list file: idd_ljmp1.lst
##
                                   created Fri Apr 20 15:51:40 2001
#0000
                   # AUXR equ 08Eh
#0000
                   # CKCON equ 08Fh
#0000
                   # org 0
                   # LJMP_LABEL:
0000 /75;/8E;/01;
                   #
                                     {\tt AUXR}, {\tt \#001h} ; turn off ALE
                             MOV
0003 /02;/FF;/FD;
                             LJMP
                                     LJMP_LABEL
                                                ; jump to end of address space
                   #
0005 /00;
                             NOP
#FFFD
                   # org Offfdh
                   # LJMP_LABEL:
                   #
FFFD /02;/FD;FF;
                   #
                             LJMP_LABEL
                   #
                             NOP
                   #
* / "
                                                                              SU01499
```

Figure 59. Source code used in measuring  $I_{\mbox{\scriptsize DD}}$  operational

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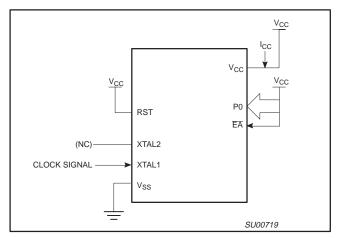


Figure 60. I<sub>CC</sub> Test Condition, Active Mode All other pins are disconnected

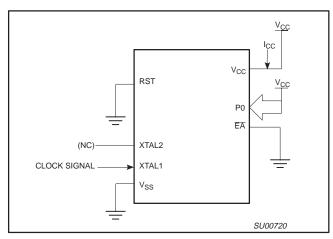


Figure 61. I<sub>CC</sub> Test Condition, Idle Mode All other pins are disconnected

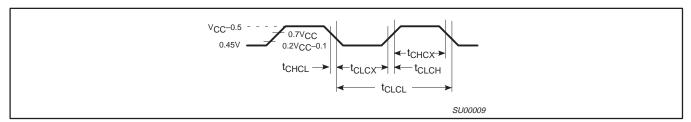


Figure 62. Clock Signal Waveform for  $I_{CC}$  Tests in Active and Idle Modes  $t_{CLCH} = t_{CHCL} = 5 ns$ 

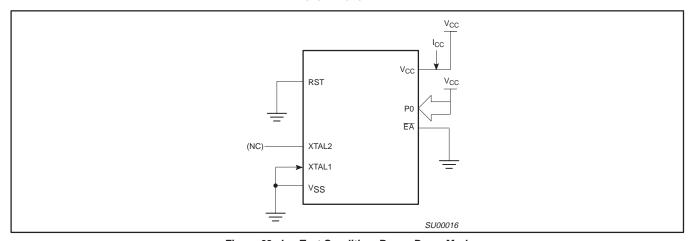


Figure 63.  $I_{CC}$  Test Condition, Power Down Mode All other pins are disconnected.  $V_{CC}$  = 2 V to 5.5 V

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## P8xC660X2/661X2

#### **EPROM CHARACTERISTICS**

All these devices can be programmed by using a modified Improved Quick-Pulse Programming algorithm. It differs from older methods in the value used for  $V_{PP}$  (programming supply voltage) and in the width and number of the ALE/PROG pulses.

The family contains two signature bytes that can be read and used by an EPROM programming system to identify the device. The signature bytes identify the device as being manufactured by Philips.

Table 17 shows the logic levels for reading the signature byte, and for programming the program memory, the encryption table, and the security bits. The circuit configuration and waveforms for quick-pulse programming are shown in Figures 64 and 65. Figure 66 shows the circuit configuration for normal program memory verification.

### **Quick-Pulse Programming**

The setup for microcontroller quick-pulse programming is shown in Figure 64. Note that the device is running with a 4 to 6MHz oscillator. The reason the oscillator needs to be running is that the device is executing internal address and program data transfers.

The address of the EPROM location to be programmed is applied to ports 1 and 2, as shown in Figure 64. The code byte to be programmed into that location is applied to port 0. RST, PSEN and pins of ports 2 and 3 specified in Table 17 are held at the 'Program Code Data' levels indicated in Table 17. The ALE/PROG is pulsed LOW 5 times as shown in Figure 65.

To program the encryption table, repeat the 5 pulse programming sequence for addresses 0 through 1FH, using the 'Pgm Encryption Table' levels. Do not forget that after the encryption table is programmed, verification cycles will produce only encrypted data.

To program the security bits, repeat the 5 pulse programming sequence using the 'Pgm Security Bit' levels. After one security bit is programmed, further programming of the code memory and encryption table is disabled. However, the other security bits can still be programmed.

Note that the  $\overline{\text{EA}}/\text{V}_{PP}$  pin must not be allowed to go above the maximum specified  $\text{V}_{PP}$  level for any amount of time. Even a narrow glitch above that voltage can cause permanent damage to the device. The  $\text{V}_{PP}$  source should be well regulated and free of glitches and overshoot.

#### **Program Verification**

If security bits 2 and 3 have not been programmed, the on-chip program memory can be read out for program verification. The address of the program memory locations to be read is applied to ports 1 and 2 as shown in Figure 66. The other pins are held at the 'Verify Code Data' levels indicated in Table 17. The contents of the address location will be emitted on port 0. External pull-ups are required on port 0 for this operation.

If the 64 byte encryption table has been programmed, the data presented at port 0 will be the exclusive NOR of the program byte with one of the encryption bytes. The user will have to know the encryption table contents in order to correctly decode the verification data. The encryption table itself cannot be read out.

#### **Reading the Signature Bytes**

The signature bytes are read by the same procedure as a normal verification of locations 030H and 031H, except that P3.6 and P3.7 need to be pulled to a logic LOW. The values are:

(030H) = 15H indicates manufactured by Philips

(031H) = C9H indicates P8xC66xX2

(060H) = 01H (660) 02H (661)

#### **Program/Verify Algorithms**

Any algorithm in agreement with the conditions listed in Table 17, and which satisfies the timing specifications, is suitable.

#### **Security Bits**

With none of the security bits programmed the code in the program memory can be verified. If the encryption table is programmed, the code will be encrypted when verified. When only security bit 1 (see Table 18) is programmed, MOVC instructions executed from external program memory are disabled from fetching code bytes from the internal memory, EA is latched on Reset and all further programming of the EPROM is disabled. When security bits 1 and 2 are programmed, in addition to the above, verify mode is disabled. When all three security bits are programmed, all of the conditions above apply and all external program memory execution is disabled.

#### **Encryption Array**

64 bytes of encryption array are initially unprogrammed (all 1s).

<sup>™</sup>Trademark phrase of Intel Corporation.

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**Table 17. EPROM Programming Modes** 

MODE	RST	PSEN	ALE/PROG	EA/V <sub>PP</sub>	P2.7	P2.6	P3.7	P3.6	P3.3
Read signature	1	0	1	1	0	0	0	0	Х
Program code data	1	0	0*	$V_{PP}$	1	0	1	1	Х
Verify code data	1	0	1	1	0	0	1	1	Х
Pgm encryption table	1	0	0*	$V_{PP}$	1	0	1	0	Х
Pgm security bit 1	1	0	0*	$V_{PP}$	1	1	1	1	Х
Pgm security bit 2	1	0	0*	$V_{PP}$	1	1	0	0	Х
Pgm security bit 3	1	0	0*	$V_{PP}$	0	1	0	1	Х
Program to 6-clock mode	1	0	0*	$V_{PP}$	0	0	1	0	0
Verify 6-clock <sup>4</sup>	1	0	1	1	е	0	0	1	1
Verify security bits <sup>5</sup>	1	0	1	1	е	0	1	0	Х

#### NOTES:

- 1. '0' = Valid LOW for that pin, '1' = Valid HIGH for that pin.
- 2.  $V_{PP} = 12.75 \text{ V} \pm 0.25 \text{ V}.$
- 3. V<sub>CC</sub> = 5 V±10% during programming and verification.
  4. Bit is output on P0.4 (1 = 12x, 0 = 6x).
- 5. Security bit one is output on P0.7. Security bit two is output on P0.6. Security bit three is output on P0.3.
- ALE/PROG receives 5 programming pulses for code data (also for user array; 5 pulses for encryption or security bits) while  $V_{PP}$  is held at 12.75 V. Each programming pulse is LOW for 100  $\mu$ s ( $\pm$ 10  $\mu$ s) and HIGH for a minimum of 10  $\mu$ s.

Table 18. Program Security Bits for EPROM Devices

PRO	OGRAM L	оск вітя	31, 2	
	SB1	SB2	SB3	PROTECTION DESCRIPTION
1	U	U	U	No Program Security features enabled. (Code verify will still be encrypted by the Encryption Array if programmed.)
2	Р	U	U	MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory, $\overline{EA}$ is sampled and latched on Reset, and further programming of the EPROM is disabled.
3	Р	Р	U	Same as 2, also verify is disabled.
4	4 P P P		Р	Same as 3, external execution is disabled. Internal data RAM is not accessible.

#### NOTES:

- 1. P programmed. U unprogrammed.
- 2. Any other combination of the security bits is not defined.

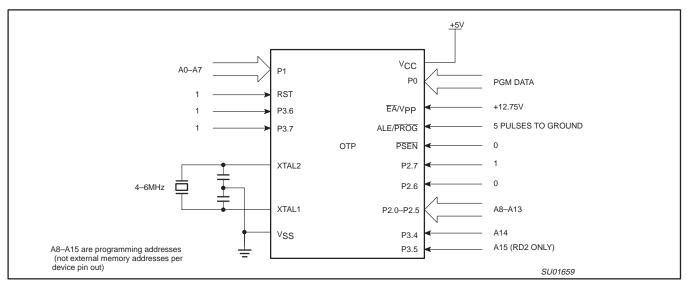


Figure 64. Programming Configuration

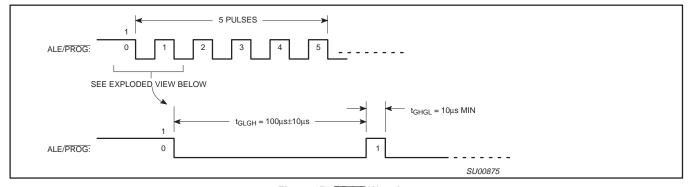


Figure 65. PROG Waveform

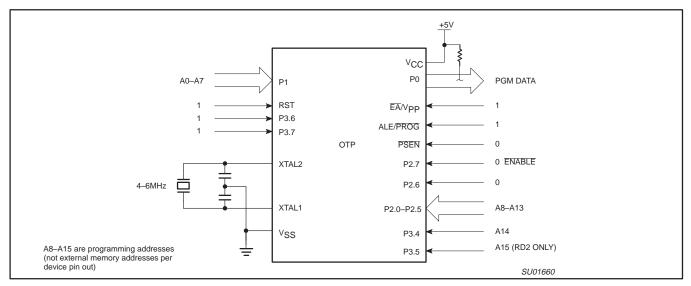


Figure 66. Program Verification

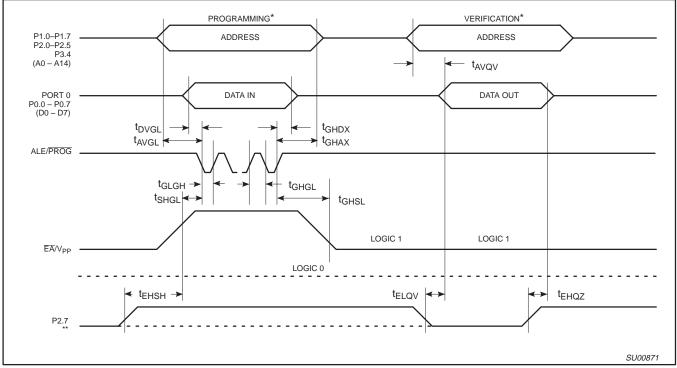
#### **EPROM PROGRAMMING AND VERIFICATION CHARACTERISTICS**

 $T_{amb}$  = 21°C to +27°C,  $V_{CC}$  = 5V±10%,  $V_{SS}$  = 0V (See Figure 67)

SYMBOL	PARAMETER	MIN	MAX	UNIT
V <sub>PP</sub>	Programming supply voltage	12.5	13.0	V
I <sub>PP</sub>	Programming supply current		50 <sup>1</sup>	mA
1/t <sub>CLCL</sub>	Oscillator frequency	4	6	MHz
t <sub>AVGL</sub>	Address setup to PROG LOW	48t <sub>CLCL</sub>		
t <sub>GHAX</sub>	Address hold after PROG	48t <sub>CLCL</sub>		
t <sub>DVGL</sub>	Data setup to PROG LOW	48t <sub>CLCL</sub>		
t <sub>GHDX</sub>	Data hold after PROG	48t <sub>CLCL</sub>		
t <sub>EHSH</sub>	P2.7 (ENABLE) HIGH to V <sub>PP</sub>	48t <sub>CLCL</sub>		
t <sub>SHGL</sub>	V <sub>PP</sub> setup to PROG LOW	10		μs
t <sub>GHSL</sub>	V <sub>PP</sub> hold after PROG	10		μs
t <sub>GLGH</sub>	PROG width	90	110	μs
t <sub>AVQV</sub>	Address to data valid		48t <sub>CLCL</sub>	
t <sub>ELQZ</sub>	ENABLE LOW to data valid		48t <sub>CLCL</sub>	
t <sub>EHQZ</sub>	Data float after ENABLE	0	48t <sub>CLCL</sub>	
t <sub>GHGL</sub>	PROG HIGH to PROG LOW	10		μs

#### NOTE:

1. Not tested.



#### NOTES:

- FOR PROGRAMMING CONFIGURATION SEE FIGURE 64. FOR VERIFICATION CONDITIONS SEE FIGURE 66.
- \*\* SEE TABLE 17.

Figure 67. EPROM Programming and Verification

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## P8xC660X2/661X2

#### MASK ROM DEVICES

### **Security Bits**

With none of the security bits programmed the code in the program memory can be verified. If the encryption table is programmed, the code will be encrypted when verified. When only security bit 1 (see Table 19) is programmed, MOVC instructions executed from external program memory are disabled from fetching code bytes

from the internal memory,  $\overline{\text{EA}}$  is latched on Reset and all further programming of the EPROM is disabled. When security bits 1 and 2 are programmed, in addition to the above, verify mode is disabled.

## **Encryption Array**

64 bytes of encryption array are initially unprogrammed (all 1s).

Table 19. Program Security Bits

PROGRA	AM LOCK	BITS <sup>1, 2</sup>	
	SB1	SB2	PROTECTION DESCRIPTION
1	U		No Program Security features enabled. (Code verify will still be encrypted by the Encryption Array if programmed.)
2	Р	U	MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory, $\overline{EA}$ is sampled and latched on Reset, and further programming of the EPROM is disabled.

#### NOTES:

- 1. P programmed. U unprogrammed.
- 2. Any other combination of the security bits is not defined.

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB  $I^2C$  interfaces

## P8xC660X2/661X2

## ROM CODE SUBMISSION FOR 16K ROM DEVICES (83C66x)

When submitting ROM code for the 16K ROM devices, the following must be specified:

- 1. 16k byte user ROM data
- 2. 64 byte ROM encryption key
- 3. ROM security bits.

ADDRESS	CONTENT	BIT(S)	COMMENT
0000H to 3FFFH	DATA	7:0	User ROM Data
4000H to 403FH	KEY	7:0	ROM Encryption Key FFH = no encryption
4040H	SEC	0	ROM Security Bit 1 0 = enable security 1 = disable security
4040H	SEC	1	ROM Security Bit 2 0 = enable security 1 = disable security

Security Bit 1: When programmed, this bit has two effects on masked ROM parts:

- 1. External MOVC is disabled, and
- 2. EA is latched on Reset.

Security Bit 2: When programmed, this bit inhibits Verify User ROM.

NOTE: Security Bit 2 cannot be enabled unless Security Bit 1 is enabled.

If the ROM Code file does not include the options, the following information must be included with the ROM code.

For each of the following, check the appropriate box, and send to Philips along with the code:

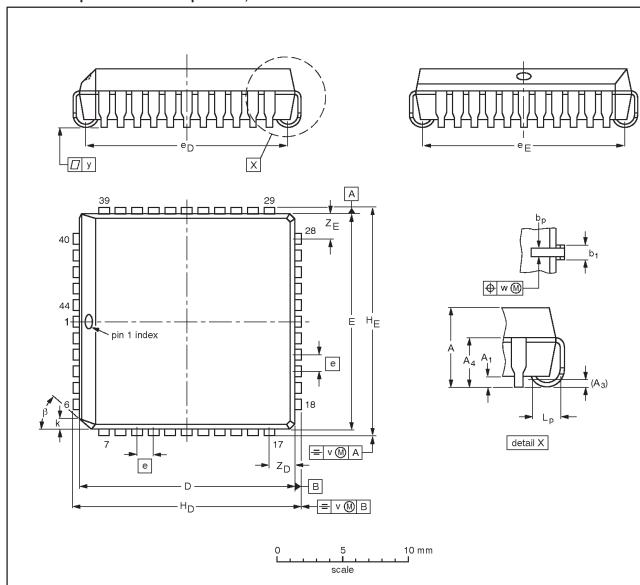
Security Bit #1:	☐ Enabled	☐ Disabled
Security Bit #2:	☐ Enabled	☐ Disabled
Encryption:	□ No	☐ Yes If Yes, must send key file.

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## P8xC660X2/661X2

## PLCC44: plastic leaded chip carrier; 44 leads

SOT187-2



## DIMENSIONS (mm dimensions are derived from the original inch dimensions)

UNIT	А	A <sub>1</sub> min.	A <sub>3</sub>	A <sub>4</sub> max.	bр	b <sub>1</sub>	D <sup>(1)</sup>	E <sup>(1)</sup>	е	еD	еE	Н <sub>D</sub>	HE	k	Lp	v	w	у	Z <sub>D</sub> <sup>(1)</sup> max.	Z <sub>E</sub> <sup>(1)</sup> max.	β
mm	4.57 4.19	0.51	0.25	3.05	0.53 0.33	0.81 0.66		16.66 16.51	1.27	16.00 14.99					1.44 1.02	0.18	0.18	0.1	2.16	2.16	45°
inches	0.180 0.165	0.02	0.01					0.656 0.650		0.63 0.59			0.695 0.685			0.007	0.007	0.004	0.085		

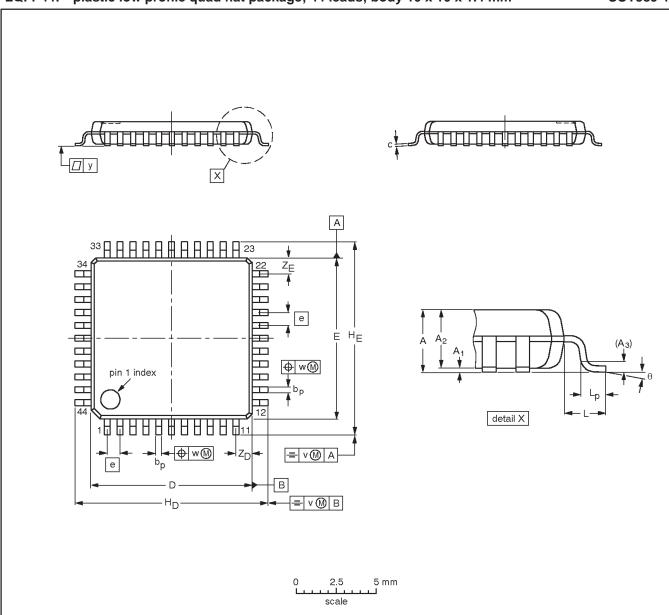
#### Note

1. Plastic or metal protrusions of 0.25 mm (0.01 inch) maximum per side are not included.

OUTLINE		REFER	EUROPEAN	ISSUE DATE		
VERSION	RSION IEC JE		JEITA	PROJECTION	ISSUE DATE	
SOT187-2	112E10	MS-018	EDR-7319		<del>-99-12-27-</del> 01-11-14	

LQFP44: plastic low profile quad flat package; 44 leads; body 10 x 10 x 1.4 mm

SOT389-1



#### **DIMENSIONS (mm are the original dimensions)**

UNIT	A max.	A <sub>1</sub>	A <sub>2</sub>	A <sub>3</sub>	Ьp	С	D <sup>(1)</sup>	E <sup>(1)</sup>	е	НD	HE	L	Lp	v	w	у	Z <sub>D</sub> <sup>(1)</sup>	Z <sub>E</sub> <sup>(1)</sup>	θ
mm	1.6	0.15 0.05	1.45 1.35	0.25	0.45 0.30	0.20 0.12	10.1 9.9	10.1 9.9	0.8	12.15 11.85	12.15 11.85	1	0.75 0.45	0.2	0.2	0.1	1.14 0.85	1.14 0.85	7° 0°

#### Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE		REFER	EUROPEAN	ISSUE DATE	
VERSION	IEC	JEDEC	JEITA	PROJECTION	ISSUE DATE
SOT389-1	136E08	MS-026			<del>-00-01-19-</del> 02-06-07

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB  $\rm I^2C$  interfaces

## P8xC660X2/661X2

## **REVISION HISTORY**

Rev	Date	Description
_3	20031002	Product data (9397 750 12144); ECN 853-2416 30396 dated 2003 September 30
		Modifications:
		Corrected pin description for V <sub>SS</sub>
		Corrected AUXR (Figure 48).
_2	20030619	Product data (9397 750 11439); ECN 853-2416 29870 dated 2003 Apr 28
_1	20030312	Product data (9397 750 11126); ECN 853-2416 29538 dated 2003 Feb 13

80C51 8-bit microcontroller family 16 KB OTP/ROM, 512B RAM, low voltage (2.7 to 5.5 V), low power, high speed (30/33 MHz), two 400KB I<sup>2</sup>C interfaces

P8xC660X2/661X2



Purchase of Philips I<sup>2</sup>C components conveys a license under the Philips' I<sup>2</sup>C patent to use the components in the I<sup>2</sup>C system provided the system conforms to the I<sup>2</sup>C specifications defined by Philips. This specification can be ordered using the code 9398 393 40011.

#### **Data sheet status**

Level	Data sheet status [1]	Product status <sup>[2] [3]</sup>	Definitions
I	Objective data	Development	This data sheet contains data from the objective specification for product development.  Philips Semiconductors reserves the right to change the specification in any manner without notice.
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- [2] The product status of the device(s) described in this data sheet may have changed since this data sheet was published. The latest information is available on the Internet at URL http://www.semiconductors.philips.com.
- [3] For data sheets describing multiple type numbers, the highest-level product status determines the data sheet status.

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